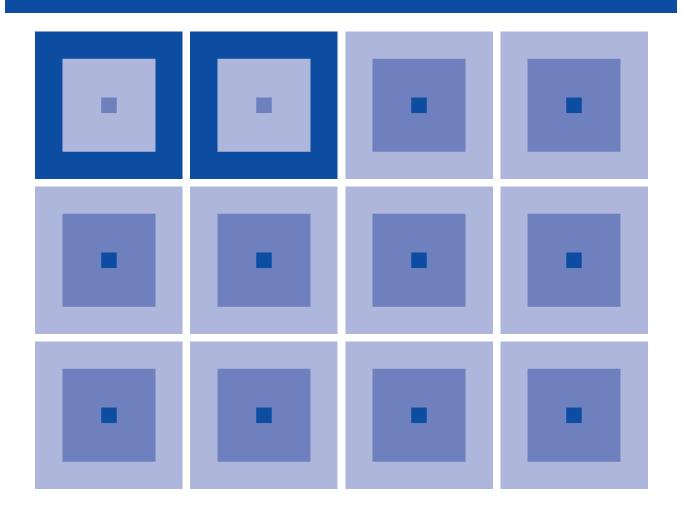
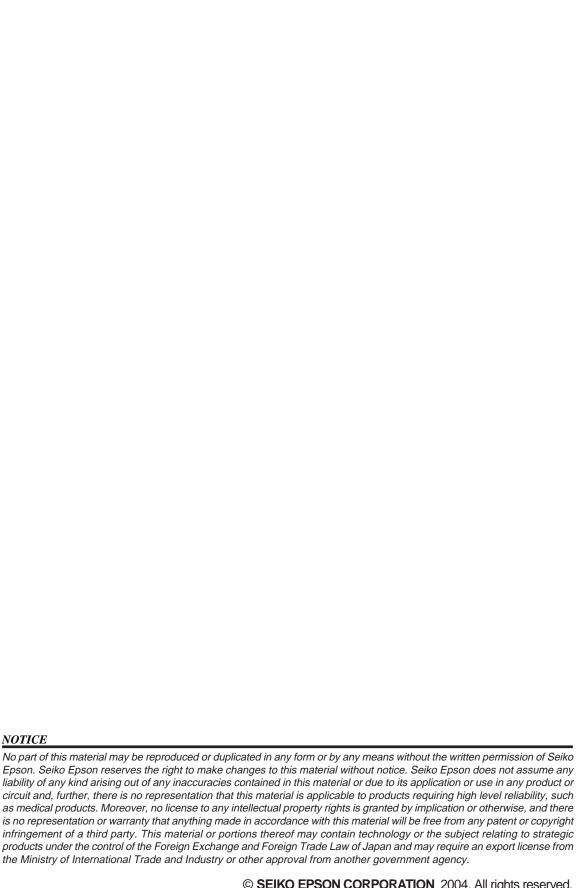


# CMOS 8-BIT SINGLE CHIP MICROCOMPUTER S1C88349 Technical Manual S1C88349 Technical Hardware

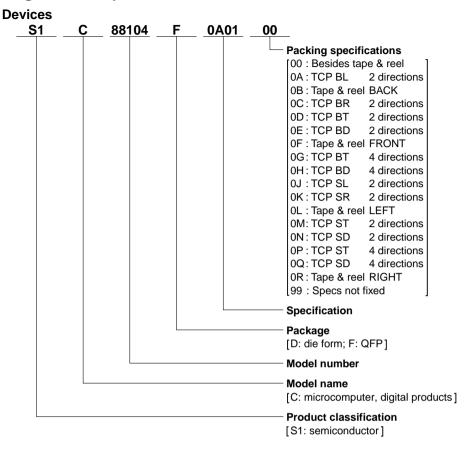


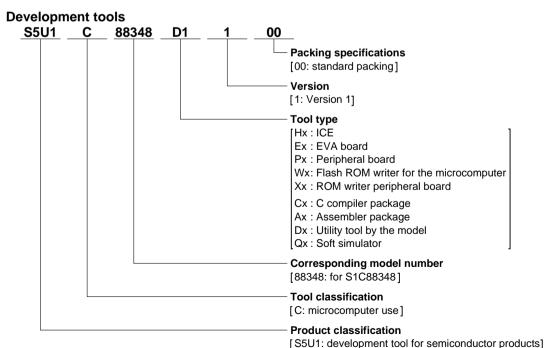


# Revisions and Additions for this manual

Chapter	Section	Page	Item	Contents
1	1.2	2	Fig. 1.2.1 S1C88349 block diagram	The figure was revised.
	1.5	5	Mask Option	The manual name was revised.
		7	Peripheral circuit board option	Explanation was added.
5	5.11.11	105	Programming notes	Item (6) was added.
	5.15.1	127	Conversion time	The values were revised.
8	8.2	150	Recommended Operating Conditions	The table was revised.
	8.4	154	Analog Circuit Characteristics	The table was revised.
			■ A/D converter circuit	
	8.6	157	AC Characteristics	A note was added.
			■ External memory access • Read cycle	
	8.8	168	Characteristics Curves (reference value)	The figures were revised.
			■ Power current (During operation with OSC3)	
			<crystal ceramic="" oscillation=""></crystal>	
			and <cr oscillation=""></cr>	
10	10.2	176	Table 10.2.1 Pad coordinates	The table was revised.
Appendix		177-	S5U1C88000P1&S5U1C88816P2 Manual	A new section was added.
			(Peripheral Circuit Board for S1C88349)	

# Configuration of product number





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# 1 INTRODUCTION

The S1C88349 microcomputer features the S1C88 (Model 3) CMOS 8-bit core CPU along with 48K bytes of ROM, 2K bytes of RAM, three different timers, a serial interface with optional asynchronization or clock synchronization, and an A/D converter.

The S1C88349 is fully operable over a wide range of voltages, and can perform high speed operations

even at low voltage. Like all the equipment in the S1C Family, these microcomputers have low power consumption.

A 19-bit external address bus and 4 bits chip enable signals make it possible for this microcomputer to control up to  $512K\times4$  bytes of memory, making them ideal for high performance data bank systems.

# 1.1 Features

Table 1.1.1 lists the features of the S1C88349.

Table 1.1.1 Main features

Core CPU	S1C88 (MODEL3)	) CMOS 8-bit core CPU				
OSC1 oscillation circuit	Crystal oscillation circuit/CR oscillation circuit/external clock input 32.768 kHz (Typ.)					
OSC3 oscillation circuit	Crystal oscillation circuit/ceramic oscillation circuit/CR oscillation circuit/external clock input 8.2 MHz (Max.)					
Instruction set	608 types (usable f	or multiplication and division in	nstructions)			
Min. instruction execution time	0.244 µsec/8.2 MF	Iz (2 clock)				
Internal ROM capacity	48K bytes					
Internal RAM capacity	2K bytes/RAM	3,216 bits/display memory				
Bus line	Address bus: 19 b	its (also usable as a general ou	itput port when not used as a bus)			
	Data bus: 8 bit	s (also usable as a general I/0	O port when not used as a bus)			
	CE signal: 4 bit	s ¬				
	WR signal: 1 bit	(also usable as a general ou	tput port when not used as a bus)			
	RD signal: 1 bit					
Input port	-	be set for event counter external	clock input and bus request signal input terminal)			
Output port			ntrol, FOUT, TOUT and bus acknowledge signal output terminal)			
I/O port	8 bits (4 bits each of	can be set for serial interface inp	out/output and analog comparator/AD input)			
Serial interface	1ch (optional clock	synchronous system or asynch	ronous system)			
Timer			s a an event counter or 2ch as a 16 bits programmable timer for 1ch)			
	Clock timer (8 bits		,			
	Stopwatch timer (8	,				
LCD driver		upports $5 \times 8$ or $5 \times 5$ fonts)				
		commons (1/5 bias)				
	_	or 8 commons (1/5 bias)				
Built-in LCD power supply circuit (booster type, 5 potentials)						
Sound generator	Envelope function, equipped with volume control					
Watchdog timer	Built-in	, -1- 11				
Analog comparator		vailable if A/D converter is used	)			
A/D converter	`		5 LSB (not available if analog comparator is used)			
Supply voltage detection		6 different voltage levels				
(SVD) circuit	r					
Interrupt	External interrupt:	Input interrupt	2 systems (3 types)			
	Internal interrupt:	•	3 systems (9 types)			
		Serial interface interrupt	1 system (3 types)			
		A/D converter interrupt	1 system (1 type)			
Supply voltage	Normal mode:	2.4 V-5.5 V (Max. 4.2 MHz)	• • • • • • • • • • • • • • • • • • • •			
11 7		1.8 V-3.5 V (Max. 80 kHz)	V <sub>D1</sub> = 1.2 V			
	^	3.5 V–5.5 V (Max. 8.2 MHz)				
Current consumption	SLEEP mode:	0.3 μΑ				
	HALT mode:	1.5 μA (Typ./normal mode)				
	Run (32 kHz):	9 μA (Typ./normal mode)				
	Run (4 MHz):	1.1 mA (Typ./normal mode)				
Supply form	, ,	FP21-176pin or chip				
Sabbil torm	2.110 17 opin, Q1	121 1. opin of emp				

<sup>\*</sup> The number of bits cited for output ports and I/O ports does not include those shared with the bus.

# 1.2 Block Diagram

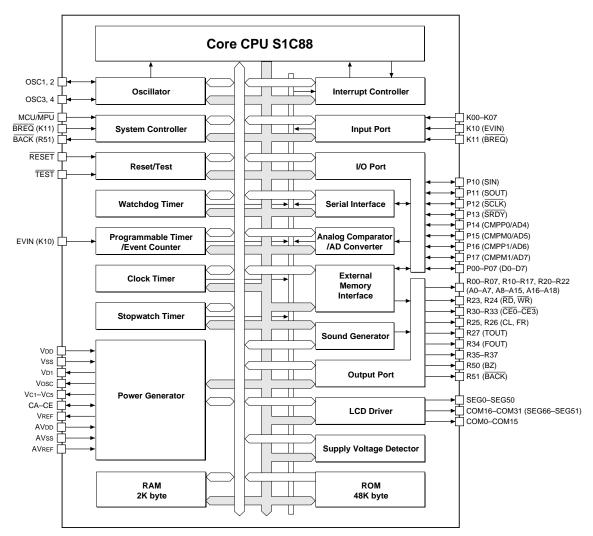
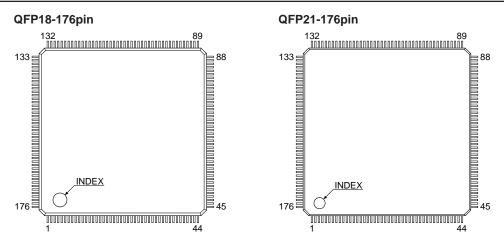


Fig. 1.2.1 S1C88349 block diagram

# 1.3 Pin Layout Diagram



Pin No.	Pin name	Pin No.	Pin name	Pin No.	Pin name	Pin No.	Pin name
1	SEG2	45	SEG46	89	OSC1	133	R11/A9
2	SEG3	46	SEG47	90	OSC2	134	R12/A10
3	SEG4	47	SEG48	91	TEST	135	R13/A11
4	SEG5	48	SEG49	92	RESET	136	R14/A12
5	SEG6	49	SEG50	93	MCU/MPU	137	R15/A13
6	SEG7	50	COM31/SEG51	94	K11/BREQ	138	R16/A14
7	SEG8	51	COM30/SEG52	95	K10/EVIN	139	R17/A15
8	SEG9	52	COM29/SEG53	96	K07	140	R20/A16
9	SEG10	53	COM28/SEG54	97	K06	141	R21/A17
10	SEG11	54	COM27/SEG55	98	K05	142	R22/A18
11	SEG12	55	COM26/SEG56	99	K04	143	R23/RD
12	SEG13	56	COM25/SEG57	100	K03	144	R24/WR
13	SEG14	57	COM24/SEG58	101	K02	145	R25/CL
14	SEG15	58	COM23/SEG59	102	K01	146	R26/FR
15	SEG16	59	COM22/SEG60	103	K00	147	R27/TOUT
16	SEG17	60	COM21/SEG61	104	P17/CMPM1/AD7	148	R30/CE0
17	SEG18	61	COM20/SEG62	105	P16/CMPP1/AD6	149	R31/CE1
18	SEG19	62	COM19/SEG63	106	P15/CMPM0/AD5	150	R32/CE2
19	SEG20	63	COM18/SEG64	107	P14/CMPP0/AD4	151	R33/CE3
20	SEG21	64	COM17/SEG65	108	P13/SRDY	152	R34/FOUT
21	SEG22	65	COM16/SEG66	109	P12/SCLK	153	R35
22	SEG23	66	N.C.	110	P11/SOUT	154	R36
23	SEG24	67	N.C.	111	P10/SIN	155	R37
24	SEG25	68	N.C.	112	AVdd	156	Vss
25	SEG26	69	N.C.	113	AVss	157	R50/BZ
26	SEG27	70	N.C.	114	AVREF	158	R51/BACK
27	SEG28	71	N.C.	115	Vdd	159	COM0
28	SEG29	72	Vref	116	P07/D7	160	COM1
29	SEG30	73	CE	117	P06/D6	161	COM2
30	SEG31	74	CD	118	P05/D5	162	COM3
31	SEG32	75	CC	119	P04/D4	163	COM4
32	SEG33	76	СВ	120	P03/D3	164	COM5
33	SEG34	77	CA	121	P02/D2	165	COM6
34	SEG35	78	Vc5	122	P01/D1	166	COM7
35	SEG36	79	VC4	123	P00/D0	167	COM8
36	SEG37	80	V <sub>C3</sub>	124	R00/A0	168	COM9
37	SEG38	81	Vc2	125	R01/A1	169	COM10
38	SEG39	82	Vcı	126	R02/A2	170	COM11
39	SEG40	83	OSC3	127	R03/A3	171	COM12
40	SEG41	84	OSC4	128	R04/A4	172	COM13
41	SEG42	85	VDI	129	R05/A5	173	COM14
42	SEG43	86	Vdd	130	R06/A6	174	COM15
43	SEG44	87	Vss	131	R07/A7	175	SEG0
44	SEG45	88	Vosc	132	R10/A8	176	SEG1

Fig. 1.3.1 S1C88349 pin layout

# 1.4 Pin Description

Table 1.4.1 S1C88349 pin description

Pin name	Pin No.	In/Out	Function
V <sub>DD</sub>	86, 115	- III/Out	Power supply (+) terminal
Vss	87, 156		Power supply (GND) terminal
V <sub>D1</sub>	85		Internal logic system voltage regulator output terminal
Vosc	88		Oscillation system voltage regulator output terminal
VC1-VC5	82–78		LCD drive voltage output terminals
CA-CE	77–73		Booster capacitor connection terminals for LCD
			*
VREF OSC1	72 89	I	LCD system power testing terminal
OSCI	89	1	OSC1 oscillation input terminal
0000	00		(select crystal oscillation/CR oscillation/external clock input by mask option)
OSC2	90	0	OSC1 oscillation output terminal
OSC3	83	I	OSC3 oscillation input terminal
0001	0.4		(select crystal/ceramic/CR oscillation/external clock input by mask option)
OSC4	84	0	OSC3 oscillation output terminal
MCU/MPU	93	I	Terminal for setting MCU or MPU modes
K00-K07	103–96	I	Input terminals (K00–K07)
K10/EVIN	95	I	Input terminal (K10) or event counter external clock input terminal (EVIN)
K11/BREQ	94	I	Input terminal (K11) or bus request signal input terminal (BREQ)
R00-R07/A0-A7	124–131	O	Output terminals (R00–R07) or address bus (A0–A7)
R10-R17/A8-A15	132–139	О	Output terminals (R10–R17) or address bus (A8–A15)
R20-R22/A16-A18	140–142	О	Output terminals (R20–R22) or address bus (A16–A18)
R23/RD	143	О	Output terminal (R23) or read signal output terminal ( $\overline{RD}$ )
R24/WR	144	O	Output terminal (R24) or write signal output terminal (WR)
R25/CL	145	O	Output terminal (R25) or LCD synchronous signal output terminal (CL)
R26/FR	146	О	Output terminal (R26) or LCD frame signal output terminal (FR)
R27/TOUT	147	О	Output terminal (R27)
			or programmable timer underflow signal output terminal (TOUT)
R30-R33/ <del>CE0</del> - <del>CE3</del>	148-151	О	Output terminals (R30–R33) or chip enable output terminals (\(\overline{CE0} - \overline{CE3}\))
R34/FOUT	152	О	Output terminal (R34) or clock output terminal (FOUT)
R35-R37	153-155	О	Output terminals (R35–R37)
R50/BZ	157	О	Output terminal (R50) or buzzer output terminal (BZ)
R51/BACK	158	О	Output terminal (R51) or bus acknowledge signal output terminal (BACK)
P00-P07/D0-D7	123-116	I/O	I/O terminals (P00–P07) or data bus (D0–D7)
P10/SIN	111	I/O	I/O terminal (P10) or serial I/F data input terminal (SIN)
P11/SOUT	110	I/O	I/O terminal (P11) or serial I/F data output terminal (SOUT)
P12/SCLK	109	I/O	I/O terminal (P12) or serial I/F clock I/O terminal (SCLK)
P13/SRDY	108	I/O	I/O terminal (P13) or serial I/F ready signal output terminal (SRDY)
P14/CMPP0/AD4	107	I/O	I/O terminal (P14), analog comparator 0 non-inverted input terminal
			or A/D converter input terminal
P15/CMPM0/AD5	106	I/O	I/O terminal (P15), analog comparator 0 inverted input terminal
			or A/D converter input terminal
P16/CMPP1/AD6	105	I/O	I/O terminal (P16), analog comparator 1 non-inverted input terminal
	100	2.0	or A/D converter input terminal
P17/CMPM1/AD7	104	I/O	I/O terminal (P17), analog comparator 1 inverted input terminal
// (2.22 1/11/11/27	101	2.0	or A/D converter input terminal
COM0-COM15	159–174	0	LCD common output terminals
COM16-COM31	65–50	0	LCD common output terminals (when 1/32 duty is selected)
/SEG66–SEG51	05 50	9	or LCD segment output terminal (when 1/32 duty is selected)
SEG0-SEG50	175–176, 1–49	0	LCD segment output terminals
RESET	92	I	Initial reset input terminal
TEST	91	I	Test input terminal
AVDD	112		Analog system power supply (+) terminal
			Analog system power supply (+) terminal  Analog system power supply (–) terminal
AVDEE	113	_	
AVREF	114		Analog system reference voltage terminal

# 1.5 Mask Option

Mask options shown below are provided for the S1C88349. Several hardware specifications are prepared in each mask option, and one of them can be selected according to the application. The function option generator winfog, that has been prepared as the development software tool of the S1C88349, is used for this selection. Mask pattern of the IC is finally generated based on the data created by the winfog. Refer to the "S5U1C88000C Manual II" for details on the winfog.

## **Option list**

The following options can be set for the S1C88349 and ICE (S5U1C88000H5). Multiple specifications are available in each option item as indicated in the Option List. Select the specifications that meet the target system and check the appropriate box. The option selection is done interactively on the screen during winfog execution, using this option list as reference.

## PERIPHERAL CIRCUIT BOARD option list

A OSC1 SYSTEM CLOCK  ☐ 1. Internal Clock ☐ 2. User Clock				
B OSC3 SYSTEM CLOCK				
☐ 1. Internal Clock				
☐ 2. User Clock				

S1C88349 mask op	otion list
[	OCK  □ 1. Crystal □ 2. External Clock □ 3. CR □ 4. Crystal (with Gate Capacity)
	OCK  □ 1. Crystal □ 2. Ceramic □ 3. CR □ 4. External Clock
	AL SET  □ 1. 512K (MAX) □ 2. 512K (MIN) □ 3. 64K
5 SVD RESET	□ 1. Not Use □ 2. Use

6 INPUT PORT PULL UP RESISTOR	
• K00	☐ 2. Gate Direct
• K01  1. With Resistor	☐ 2. Gate Direct
• K02 1. With Resistor	☐ 2. Gate Direct
• K03 1. With Resistor	□ 2. Gate Direct
• K04 1. With Resistor	□ 2. Gate Direct
• K05 1. With Resistor	□ 2. Gate Direct
• K06   1. With Resistor	□ 2. Gate Direct
• K07 ☐ 1. With Resistor	2. Gate Direct
• K10 ☐ 1. With Resistor	☐ 2. Gate Direct
• K11 ☐ 1. With Resistor	☐ 2. Gate Direct
• MCU/MPU □ 1. With Resistor	☐ 2. Gate Direct
• RESET □ 1. With Resistor	☐ 2. Gate Direct
7 I/O PORT PULL UP RESISTOR	
• P00 ☐ 1. With Resistor	☐ 2. Gate Direct
• P01 🗆 1. With Resistor	☐ 2. Gate Direct
• P02 🗆 1. With Resistor	☐ 2. Gate Direct
• P03 🗆 1. With Resistor	☐ 2. Gate Direct
• P04 □ 1. With Resistor	☐ 2. Gate Direct
• P05 □ 1. With Resistor	☐ 2. Gate Direct
• P06 ☐ 1. With Resistor	☐ 2. Gate Direct
• P07 1. With Resistor	☐ 2. Gate Direct
• P10 1. With Resistor	☐ 2. Gate Direct
• P11 1. With Resistor	☐ 2. Gate Direct
• P12 1. With Resistor	☐ 2. Gate Direct
• P13 1. With Resistor	☐ 2. Gate Direct
• P14 1. With Resistor	☐ 2. Gate Direct
• P15 1. With Resistor	2. Gate Direct
• P16 1. With Resistor	2. Gate Direct
• P17 □ 1. With Resistor	☐ 2. Gate Direct
8 OUTPUT PORT SPECIFICATION	
• R00	☐ 2. Nch Open Drain
• R01	□ 2. Nch Open Drain
• R02 1. Complementary	□ 2. Nch Open Drain
• R03 1. Complementary	☐ 2. Nch Open Drain
• R04 1. Complementary	☐ 2. Nch Open Drain
• R05 □ 1. Complementary	☐ 2. Nch Open Drain
• R06 □ 1. Complementary	☐ 2. Nch Open Drain
• R07 ☐ 1. Complementary	☐ 2. Nch Open Drain
• R10 □ 1. Complementary	☐ 2. Nch Open Drain
• R11 ☐ 1. Complementary	☐ 2. Nch Open Drain
• R12 □ 1. Complementary	☐ 2. Nch Open Drain
• R13 🗆 1. Complementary	☐ 2. Nch Open Drain
• R14 □ 1. Complementary	☐ 2. Nch Open Drain
• R15 1. Complementary	☐ 2. Nch Open Drain
• R16 1. Complementary	☐ 2. Nch Open Drain
• R17 □ 1. Complementary	☐ 2. Nch Open Drain
9 LCD POWER SUPPLY	
□ 1. Internal	
□ 1. Internal	
□ ‰. LAttillar	

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## **Outline of mask options**

#### Peripheral circuit board option

These options configure the Peripheral Circuit Board (S5U1C88000P1 with S5U1C88816P2) installed in the ICE (S5U1C88000H5). The selections do not affect the IC's mask option.

### (A) OSC1 system clock

When User Clock is selected, input a clock to the OSC1 terminal. When Internal Clock is selected, the clock frequency is changed according to the oscillation circuit selected by the IC's mask option.

## (B) OSC3 system clock

When User Clock is selected, input a clock to the OSC3 terminal. When Internal Clock is selected, the clock frequency is changed according to the oscillation circuit selected by the IC's mask option.

# S1C88349 mask option

#### (1) OSC1 oscillation circuit

The specification of the OSC1 oscillation circuit can be selected from among four types: "Crystal oscillation", "CR oscillation", "Crystal oscillation (gate capacitor built-in)" and "External clock input". Refer to Section 5.4.3, "OSC1 oscillation circuit", for details.

# (2) OSC3 oscillation circuit

The specification of the OSC3 oscillation circuit can be selected from among four types: "Crystal oscillation", "Ceramic oscillation", "CR oscillation" and "External clock input". Refer to Section 5.4.4, "OSC3 oscillation circuit", for details.

# (3) Multiple key entry reset (K00-K03)

This function resets the IC when several keys are pressed simultaneously. The mask option is used to select whether this function is used or not. Further when the function is used, a combination of the input ports (K00–K03), which are connected to the keys to be pressed simultaneously, can be selected. Refer to Section 4.1.2, "Simultaneous LOW level input at input port terminals K00–K03", for details.

#### (4) Initial bus mode for MPU mode

The bus mode that is initially set in MPU mode can be selected from expanded 512K maximum mode, expanded 512K minimum mode and expanded 64K mode. Refer to Section 5.2, "System controller and Bus Control", for details.

#### (5) Initial reset by SVD circuit

The SVD circuit has a function that generates an initial reset signal when the supply voltage drops to level 0 or less. The mask option is used to select whether this function is used or not. Refer to Section 5.16, "Supply Voltage Detection (SVD) Circuit", for details.

# (6) Input port, MCU/MPU and RESET terminal pull-up resistors

This mask option can select whether the pull-up resistor for the input (K) port terminal is used or not. It is possible to select for each bit of the input ports. Refer to Section 5.5, "Input Ports (K ports)", for details.

Furthermore, a pull-up option is also provided for the MCU/MPU and RESET terminals.

## (7) I/O port pull-up resistors

This mask option can select whether the pull-up resistor for the I/O port terminal (it works during input mode) is used or not. It is possible to select for each bit of the I/O ports. Refer to Section 5.7, "I/O Ports (P ports)", for details.

Since P10 to P13 are shared with the serial interface I/O terminals, the selected P10 and P12 terminal configuration is applied to the serial input (SIN) terminal and serial clock input terminal (SCLK in clock synchronous mode) when the serial interface is used. Refer to Section 5.8. "Serial Interface", for details.

#### (8) Output port specifications

Either complementary output or Nch open drain output can be selected as the output specification for the output ports R00–R07 and R10–R17 in 1 bit units. Refer to Section 5.6.2, "Mask option", for details.

# (9) LCD power supply

Either the internal power supply (internal LCD system voltage regulator and voltage booster that generate VC1 to VC5) or an external power supply can be selected as the LCD system power source. Refer to Section 5.12.2, "Mask option", for details.

# 2 POWER SUPPLY

In this section, we will explain the operating voltage and the configuration of the internal power supply circuit of the S1C88349.

# 2.1 Operating Voltage

The S1C88349 operating power voltage is as follows:

Normal mode: 2.4 V to 5.5 V Low power mode: 1.8 V to 3.5 V High speed mode: 3.5 V to 5.5 V

# 2.2 Internal Power Supply Circuit

The S1C88349 incorporates the power supply circuit shown in Figure 2.2.1. When voltage within the range described above is supplied to VDD (+) and Vss (GND), all the voltages needed for the internal circuit are generated internally in the IC.

Roughly speaking, the power supply circuit is divided into three sections.

The internal logic voltage regulator generates the operating voltage <VD1> for driving the internal logic circuits and the OSC3 oscillation circuit. The VD1 voltage can be selected from the following three types: 1.2 V for low-power mode, 2.2 V for normal mode and 3.1 V for high-speed mode. It should be selected by a program to switch according to the supply voltage and oscillation frequency.

See Section 5.4, "Oscillation Circuits and Operating Mode", for the switching of operating mode.

Note: Under no circumstances should V<sub>D1</sub> terminal output be used to drive external circuit.

The oscillation system voltage regulator generates the operating voltage <Vosc> for the OSC1 oscillation circuit.

The LCD system power supply circuit generates the drive voltage for the LCD. Drive voltage has five potentials VC1-VC5 for 1/5 bias: VC1 and VC2 are generated by the LCD voltage regulator, and are boosted to generate VC3-VC5. These voltages are output from the terminals to drive an external LCD driver (S1D15210 or equivalent).

See Chapter 8, "ELECTRICAL CHARACTERIS-TICS" for the voltage values.

In the S1C88349, the LCD drive voltage is supplied to the built-in LCD driver which drives the LCD panel connected to the SEG and COM terminals.

- Notes: Do not use terminals Vc1–Vc5 except to supply voltage to the expanded LCD driver.
  - Refer to the "S1D15000 Series Technical Manual" for the S1D15210 externaly expanded LCD driver.

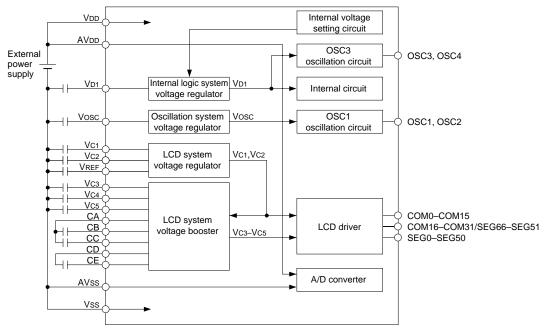


Fig. 2.2.1 Configuration of power supply circuit

# 3 CPU AND BUS CONFIGURATION

In this section, we will explain the CPU, operating mode and bus configuration.

# 3.1 CPU

The S1C88349 utilize the S1C88 8-bit core CPU whose register configuration, command set, etc. are virtually identical to other units in the family of processors incorporating the S1C88.

See the "S1C88 Core CPU Manual" for the S1C88.

Specifically, the S1C88349 employ the Model 3 S1C88 CPU which has a maximum address space of 512K bytes  $\times$  4.

# 3.2 Internal Memory

The S1C88349 is equipped with internal ROM and RAM as shown in Figure 3.2.1. Small scale applications can be handled by one chip. It is also possible to utilize internal memory in combination with external memory.

Furthermore, internal ROM can be disconnected from the bus and the resulting space released for external applications.

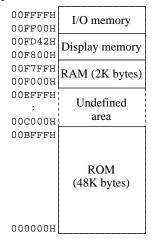


Fig. 3.2.1 Internal memory map

# 3.2.1 ROM

The S1C88349 has a built-in 48K-byte ROM. The ROM is allocated to 000000H–00BFFFH. The ROM areas shown above can be released to external memory depending on the setting of the MCU/MPU terminal. (See "3.5 Chip Mode".)

#### 3.2.2 RAM

The internal RAM capacity is 2K bytes and is allocated to 00F000H-00F7FFH.

Even when external memory which overlaps the internal RAM area is expanded, the RAM area is not released to external memory. Access to this area is via internal RAM.

# 3.2.3 I/O memory

A memory mapped I/O method is employed in the S1C88349 for interfacing with internal peripheral circuit. Peripheral circuit control bits and data register are arranged in data memory space. Control and data exchange are conducted via normal memory access. I/O memory is arranged in page 0: 00FF00H–00FFFFH area.

See Section 5.1, "I/O Memory Map", for details of the I/O memory.

Even when external memory which overlaps the I/O memory area is expanded, the I/O memory area is not released to external memory. Access to this area is via I/O memory.

# 3.2.4 Display memory

The S1C88349 is equipped with an internal display memory which stores a display data for LCD driver

Display memory is arranged in page 0: 00Fx00H–00Fx42H (x = 8-DH) in the data memory area. See Section 5.12, "LCD Controller", for details of the display memory. Like the I/O memory, display memory cannot be released to external memory.

# 3.3 Exception Processing Vectors

000000H–000025H in the program area of the S1C88349 is assigned as exception processing vectors. Furthermore, from 000028H to 0000FFH, software interrupt vectors are assignable to any two bytes which begin with an even address. Table 3.3.1 lists the vector addresses and the exception processing factors to which they correspond.

Table 3.3.1 Exception processing vector table

Vector address	Exception processing factor	Priority
000000H	Reset	High
000002H	Zero division	$\uparrow$
000004H	Watchdog timer (NMI)	
000006Н	Programmable timer 1 interrupt	
000008H	Programmable timer 0 interrupt	
00000AH	K10, K11 input interrupt	
00000CH	K04–K07 input interrupt	
00000EH	K00–K03 input interrupt	
000010H	Serial I/F error interrupt	
000012H	Serial I/F receiving complete interrupt	
000014H	Serial I/F transmitting complete interrupt	
000016H	Stopwatch timer 100 Hz interrupt	
000018H	Stopwatch timer 10 Hz interrupt	
00001AH	Stopwatch timer 1 Hz interrupt	
00001CH	Clock timer 32 Hz interrupt	
00001EH	Clock timer 8 Hz interrupt	
000020H	Clock timer 2 Hz interrupt	
000022H	Clock timer 1 Hz interrupt	$\downarrow$
000024H	A/D converter interrupt	Low
000026H	System reserved (cannot be used)	No
000028H		
:	Software interrupt	priority
0000FEH		rating

For each vector address and the address after it, the start address of the exception processing routine is written into the subordinate and super ordinate sequence. When an exception processing factor is generated, the exception processing routine is executed starting from the recorded address. When multiple exception processing factors are generated at the same time, execution starts with the highest priority item.

The priority sequence shown in Table 3.3.1 assumes that the interrupt priority levels are all the same. The interrupt priority levels can be set by software in each system. (See Section 5.17 "Interrupt and Standby Status".)

Note: For exception processing other than reset, SC (system condition flag) and PC (program counter) are evacuated to the stack and branches to the exception processing routines. Consequently, when returning to the main routine from exception processing routines, please use the RETE instruction.

See the "S1C88 Core CPU Manual" for information on CPU operations when an exception processing factor is generated.

# 3.4 CC (Customized Condition Flag)

The S1C88349 does not use the customized condition flag (CC) in the core CPU. Accordingly, it cannot be used as a branching condition for the conditional branching instruction (JRS, CARS).

# 3.5 Chip Mode

## 3.5.1 MCU mode and MPU mode

The chip operating mode can be set to one of two settings using the  $MCU/\overline{MPU}$  terminal.

■ MCU mode...Set the MCU/MPU terminal to HIGH Switch to this setting when using internal ROM. With respect to areas other than internal memory, external memory can even be expanded. See Section 3.5.2, "Bus mode", for the memory map.

In the MCU mode, during initial reset, only systems in internal memory are activated. Internal ROM is normally fixed as the top portion of the program memory from the common area (logical space 0000H-7FFFH). Exception processing vectors are assigned in internal ROM. Furthermore, the application initialization routines that start with reset exception processing must likewise be written to internal ROM. Since bus and other settings which correlate with external expanded memory can be executed in software, this processing is executed in the initialization routine written to internal ROM. Once these bus mode settings are made, external memory can be accessed.

When accessing internal memory in this mode, the chip enable  $(\overline{CE})$  and read  $(\overline{RD})$ /write  $(\overline{WR})$  signals are not output to external memory, and the data bus (D0-D7) goes into high impedance status (pull-up status when the "pull-up resistors for P00-P07 enabled" have been selected by the mask option). Consequently, in cases where addresses overlap in external and internal memory, the areas in

■ MPU mode...Set the MCUMPU terminal to LOW Internal ROM area is released to an external device source. Internal ROM then becomes unusable and when this area is accessed, chip enable (CE) and read (RD)/write (WR) signals are output to external memory and the data bus (D0–D7) become active. These signals are not output to an external source when other areas of internal memory are accessed.

external memory will be unavailable.

In the MPU mode, the system is activated by external memory.

For this reason, in order to adjust bus settings to conform to the configuration of external memory during initial reset, the user can select the applicable system configuration using the mask option. (See "3.5.2 Bus mode".)

When employing this mode, the exception processing vectors and initialization routine must be assigned within the common area (000000H–007FFFH).

You can select whether to use the built-in pull-up resistor of the  $MCU/\overline{MPU}$  terminal by the mask option.

Note: Setting of MCU/MPU terminal is latched at the rising edge of a reset signal input from the RESET terminal. Therefore, if the setting is to be changed, the RESET terminal must be set to LOW level once again.

# 3.5.2 Bus mode

In order to set bus specifications to match the configuration of external expanded memory, four different bus modes described below are selectable in software.

#### ■ Single chip mode

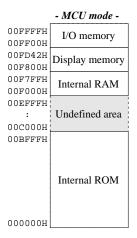


Fig. 3.5.2.1 Memory map for the single chip mode

The single chip mode setting applies when the S1C88349 is used as a single chip microcomputer without external expanded memory. Since this mode employs internal ROM, the system can only be operated in the MCU mode discussed in Section 3.5.1. In the MPU mode, the system cannot be set to the single chip mode. Since there is no need for an external bus line in this mode, terminals normally set for bus use can be used as general purpose output ports or I/O ports.

Accordingly, the output ports are in a 34-bit configuration in the S1C88349 and the I/O ports are in a 16-bit configuration.

CPU operation in this mode is equivalent to the S1C88 core CPU Model 3 minimum mode. Addresses assigned to internal memory within physical space 000000H to 00FFFFH are only effective as a target for accessing.

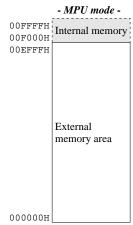
#### **■** Expanded 64K mode (MPU mode)

The expanded 64K mode setting applies when the S1C88349 is used with 64K bytes or less of external expanded memory. This mode is only usable on the MPU mode setting.

Since the internal ROM area is released in the MPU mode, external memory in this model can be assigned to the area from 000000H to 00EFFFH. The area from 00F000H to 00FFFFH is assigned to internal memory (RAM, etc.) and cannot be used to access an external device.

This mode setting is suitable for small- to midscale systems. The address range of the chip enable  $(\overline{CE})$  signal, adapted to memory chips with a capacity of from 8 to 64K bytes, can be selected in software to any one of four settings. See Section 3.6.4, "Chip enable  $(\overline{CE})$  signal", for the  $\overline{CE}$  signal.

CPU operation in this mode is equivalent to the S1C88 core CPU Model 3 minimum mode. The area within physical space 000000H to 00FFFFH is only effective as a target for accessing.



See Figure 3.2.1 for the internal memory

Fig. 3.5.2.2 Memory map for the expanded 64K mode (MPU mode)

#### **■** Expanded 512K minimum mode

The expanded 512K minimum mode setting applies when the S1C88349 is used with over 64K bytes and less than 512K bytes  $\times$  4 of external expanded memory. This mode is usable regardless of the MCU/MPU mode setting.

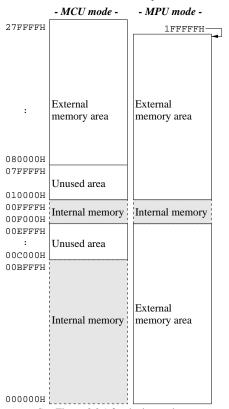
Because internal ROM is being used in the MCU mode, external memory in this model can be assigned to the area from 080000H to 27FFFFH. Since the internal ROM area is released in the MPU mode, external memory in this model can be assigned to the area from 000000H to 1FFFFFH.

However, the area from 00F000H to 00FFFFH is assigned to internal memory and cannot be used to access an external device.

CPU operation in this mode is equivalent to the S1C88 core CPU Model3 minimum mode. The area within physical space 000000H to 1FFFFFH in the MPU mode or physical space 080000H to 27FFFFH + internal memory in the MCU mode is effective as a target for accessing.

Furthermore, since program memory expansion is limited to less than 64K bytes configured with the common area (000000H to 007FFFH) and one optional bank area (internal ROM  $\pm$  32K in the MCU mode), this mode is suitable for small-to mid-scale program memory and large-scale data memory systems.

The address range of chip enable (CE) signals in this mode is fixed at 512K bytes.



See Figure 3.2.1 for the internal memory

Fig. 3.5.2.3 Memory map for the expanded 512K minimum mode

#### **■** Expanded 512K maximum mode

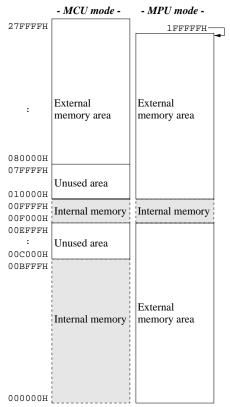
The expanded 512K maximum mode setting applies when the S1C88349 is used with over 64K bytes and less than 512K bytes  $\times$  4 of external expanded memory. This mode is usable regardless of the MCU/MPU mode setting.

Because internal ROM is being used in the MCU mode, external memory in this model can be assigned to the area from 080000H to 27FFFFH. Since the internal ROM area is released in the MPU mode, external memory in this model can be assigned to the area from 000000H to 1FFFFFH.

The area from 00F000H to 00FFFFH is assigned to internal memory and cannot be used to access an external device.

CPU operation in this mode is equivalent to the S1C88 core CPU Model 3 maximum mode, the area within physical space 000000H to 1FFFFFH in the MPU mode or physical space 080000H to 27FFFFH + internal memory in the MCU mode is effective as a target for accessing. In the above mentioned physical space, since program memory and data memory can be secured with an optional (maximum 512K bytes × 4 program + data) size, this mode is suitable for systems with large-scale program and data capacity.

The address range of chip enable  $(\overline{CE})$  signals in this mode is fixed at 512K bytes.



See Figure 3.2.1 for the internal memory

Fig. 3.5.2.4 Memory map for the expanded 512K maximum mode

There is an explanation on how all these settings are actually made in "5.2 System Controller and Bus Control" of this Manual.

# 3.6 External Bus

The S1C88349 has bus terminals that can address a maximum of  $512K \times 4$  bytes and memory (and other) devices can be externally expanded according to the range of each bus mode described in the previous section.

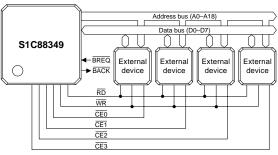


Fig. 3.6.1 External bus lines

Below is an explanation of external bus terminals. For information on control methods, see Section 5.2, "System Controller and Bus Control".

#### 3.6.1 Data bus

The S1C88349 possesses an 8-bit external data bus (D0–D7). The terminals and I/O circuits of data bus D0–D7 are shared with I/O ports P00–P07, switching between these functions being determined by the bus mode setting.

In the single chip mode, the 8-bit terminals are all set as I/O ports P00–P07 and in the other expanded modes, they are set as data bus (D0–D7).

When set as data bus, the data register and I/O control register of each I/O port are detached from the I/O circuits and usable as a general purpose data register with read/write capabilities.

With regard to the pull-up resistors that go ON only in input mode, the mask option can be used to select whether or not to use the pull-up resistor for each data bus line. (The same holds true when the terminals are used as I/O ports.)

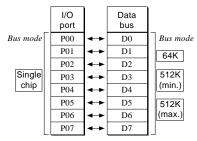


Fig. 3.6.1.1 Correspondence between data bus and I/O ports

#### 3.6.2 Address bus

The S1C88349 possesses a 19-bit external address bus A0–A18. The terminals and output circuits of address bus A0–A18 are shared with output ports R00–R07 (=A0–A7), R10–R17 (=A8–A15) and R20–R22 (=A16–A18), switching between these functions being determined by the bus mode setting. In the single chip mode, the 19-bit terminals are all set as output ports R00–R07, R10–R17 and R20–R22. In the expanded 64K mode, 16 of the 19-bit terminals, A0–A15, are set as the address bus, while the remaining 3 bits, A16–A18, are set as output ports R20–R22.

In the expanded 512K minimum and maximum modes, all of the 19-bit terminals are set as the address bus (A0–A18).

When set as an address bus, the data register and high impedance control register of each output port are detached from the output circuit and used as a general purpose data register with read/write capabilities.

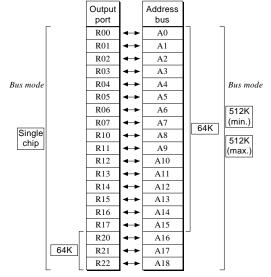


Fig. 3.6.2.1 Correspondence between address bus and output ports

# 3.6.3 Read ( $\overline{RD}$ )/write ( $\overline{WR}$ ) signals

The output terminals and output circuits for the read  $(\overline{RD})/\text{write}$   $(\overline{WR})$  signals directed to external devices are shared respectively with output ports R23 and R24, switching between these functions being determined by the bus mode setting. In the single chip mode, both of these terminals are set as output port terminals and in the other expanded modes, they are set as read  $(\overline{RD})/\text{write}$   $(\overline{WR})$  signal output terminals. When set as read  $(\overline{RD})/\text{write}$   $(\overline{WR})$  signal output terminal, the data register and high impedance control register for each output port (R23, R24) are detached from the output circuit and is usable as a general purpose data register with read/write capabilities.

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These two signals are only output when the memory area of the external device is being accessed. They are not output when internal memory is accessed.

See Section 3.6.5, "WAIT control", for the output timing of the signal.

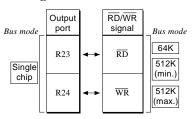


Fig. 3.6.3.1 Correspondence between read  $(\overline{RD})$ /write  $(\overline{WR})$  signal and output ports

# 3.6.4 Chip enable $(\overline{CE})$ signal

The S1C88349 is equipped with address decoders which can output four different chip enable  $(\overline{\text{CE}})$  signals.

Consequently, four devices equipped with a chip enable ( $\overline{\text{CE}}$ ) or chip select ( $\overline{\text{CS}}$ ) terminal can be directly connected without setting the address decoder to an external device.

The four chip enable  $(\overline{CE0}-\overline{CE3})$  signal output terminals and output circuits are shared with output ports R30–R33 and in modes other than the single chip mode, the selection of chip enable  $(\overline{CE})$  or output port can be set in software for each of the four bits. When set for chip enable  $(\overline{CE})$  output, the data register and high impedance control register for each output port are detached from the output circuit and is usable as general purpose data register with read/write capabilities.

In the single chip mode, these terminals are set as output ports R30–R33.

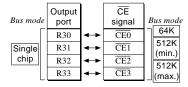


Fig. 3.6.4.1 Correspondence between  $\overline{CE}$  signals and output ports

The address range assigned to the four chip enable  $(\overline{CE})$  signals is determined by the bus mode setting. In the expanded 64K mode, the four different address ranges which match the amount of memory in use can be selected in software. Table 3.6.4.1 shows the address ranges which are assigned to the chip enable  $(\overline{CE})$  signal in each mode. When accessing the internal memory area, the  $\overline{CE}$  signal is not output. Care should be taken here because the address range for these portions of memory involves irregular settings.

The arrangement of memory space for external devices does not necessarily have to be continuous from a subordinate address and any of the chip enable signals can be used to assign areas in memory.

Each of these signals is only output when the memory area of the external device is being accessed. They are not output when internal memory is accessed.

Note: The  $\overline{CE}$  signals will be inactive status when the chip enters the standby mode (HALT mode or SLEEP mode).

See Section 3.6.5, "WAIT control", for the output timing of signal.

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Table 3.6.4.1 \(\overline{CE0}\)\(-\overline{CE3}\) address settings

## (1) Expanded 64K mode (MPU mode only)

CE signal	Address range (selected in software)									
CE signal	8K bytes	16K bytes	32K bytes	64K bytes						
CE0	000000H-001FFFH	000000H-003FFFH	000000H-007FFFH	000000H-00EFFFH						
CE1	002000H-003FFFH	004000H-007FFFH	008000H-00EFFFH	_						
CE2	004000H-005FFFH	008000H-00BFFFH	_	_						
CE3	006000H-007FFFH	00C000H-00EFFFH	_	_						

#### (2) Expanded 512K minimum/maximum modes

CF signal	Addres	s range
CE signal	MCU mode	MPU mode
CE0	200000H-27FFFFH	000000H-00EFFFH, 010000H-07FFFFH
CE1	080000H-0FFFFH	080000H-0FFFFH
CE2	100000H-17FFFFH	100000H-17FFFFH
CE3	180000H-1FFFFFH	180000H-1FFFFH

#### 3.6.5 WAIT control

In order to insure accessing of external low speed devices during high speed operations, the S1C88349 is equipped with a WAIT function which prolongs access time. (See the "S1C88 Core CPU Manual" for details of the WAIT function.)

The WAIT state numbers to be inserted can be selected in software from a series of 8 as shown in Table 3.6.5.1.

Table 3.6.5.1 Selectable WAIT state numbers

Selection No.	1	2	3	4	5	6	7	8
Insert states	0	2	4	6	8	10	12	14

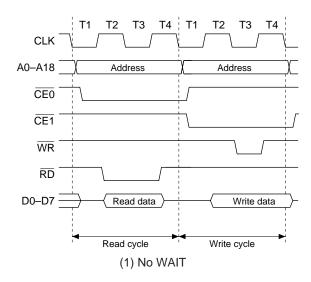
<sup>\*</sup> One state is a 1/2 cycle of the clock in length.

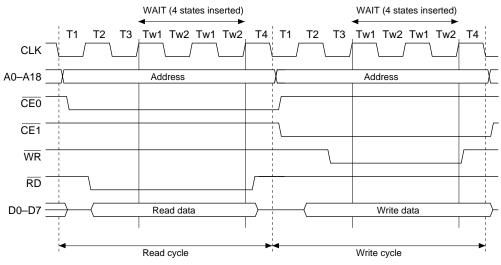
The WAIT states set in software are inserted between bus cycle states T3–T4.

Note, however, that WAIT states cannot be inserted when an internal register and internal memory are being accessed and when operating with the OSC1 oscillation circuit (see "5.4 Oscillation Circuits and Operating Mode").

Consequently, WAIT state settings are meaningless in the single chip mode.

Figure 3.6.5.1 shows the memory read/write timing charts.





(2) WAIT state insertion

Fig. 3.6.5.1 Memory read/write cycle

# 3.6.6 Bus authority release state

The S1C88349 is equipped with a bus authority release function on request from an external device so that DMA (Direct Memory Access) transfer can be conducted between external devices. The internal memory cannot be accessed by this function.

There are two terminals used for this function: the bus authority release request signal ( $\overline{BREQ}$ ) input terminal and the bus authority release acknowledge signal ( $\overline{BACK}$ ) output terminal.

The  $\overline{BREQ}$  input terminal is shared with input port terminal K11 and the  $\overline{BACK}$  output terminal with output port terminal R51, use with setting to  $\overline{BREQ}/\overline{BACK}$  terminals done in software. In the single chip mode, or when using a system which does not require bus authority release, set respective terminals as input and output ports.

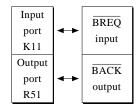


Fig. 3.6.6.1 BREQ/BACK terminals

When the bus authority release request ( $\overline{BREQ}$  = LOW) is received from an external device, the S1C88349 switches the address bus, data bus,  $\overline{RD}/\overline{WR}$  signal, and  $\overline{CE}$  signal lines to a high impedance state, outputs a LOW level from the  $\overline{BACK}$  terminal and releases bus authority.

As soon as a LOW level is output from the  $\overline{BACK}$  terminal, the external device can use the external bus. When  $\overline{DMA}$  is completed, the external device returns the  $\overline{BREQ}$  terminal to HIGH and releases bus authority.

Figure 3.6.6.2 shows the bus authority release sequence.

During bus authority release state, internal memory cannot be accessed from the external device. In cases where external memory has areas which overlap areas in internal memory, the external memory areas can be accessed accordance with the  $\overline{\text{CE}}$  signal output by the external device.

Note: Be careful with the system, such that an external device does not become the bus master, other than during the bus release status.

After setting the BREQ terminal to LOW level, hold the BREQ terminal at LOW level until the BACK terminal becomes LOW level. If the BREQ terminal is returned to HIGH level, before the BACK terminal becomes LOW level, the shift to the bus authorization release status will become indefinite.

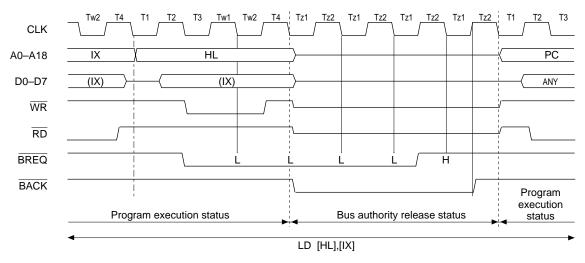


Fig. 3.6.6.2 Bus authority release sequence

# 4 INITIAL RESET

Initial reset in the S1C88349 is required in order to initialize circuits. This section of the Manual contains a description of initial reset factors and the initial settings for internal registers, etc.

# 4.1 Initial Reset Factors

There are three initial reset factors for the S1C88349 as shown below.

- (1) RESET terminal
- (2) Simultaneous LOW level input at input port terminals K00–K03
- (3) Supply voltage detection (SVD) circuit

Figure 4.1.1 shows the configuration of the initial reset circuit.

The CPU and peripheral circuits are initialized by means of initial reset factors. When the factor is canceled, the CPU commences reset exception processing. (See "S1C88 Core CPU Manual".)

When this occurs, reset exception processing vectors, Bank 0, 000000H–000001H from program memory are read out and the program (initialization routine) which begins at the readout address is executed.

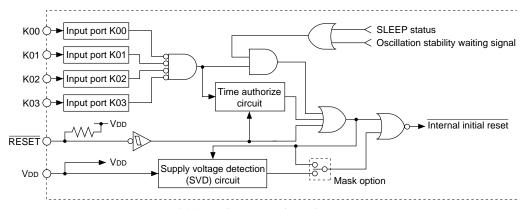


Fig. 4.1.1 Configuration of initial reset circuit

# 4.1.1 RESET terminal

Initial reset can be done by executed externally inputting a LOW level to the  $\overline{RESET}$  terminal. Be sure to maintain the  $\overline{RESET}$  terminal at LOW level for the regulation time after the power on to assure the initial reset.

In addition, be sure to use the  $\overline{RESET}$  terminal for the first initial reset after the power is turned on. The  $\overline{RESET}$  terminal is equipped with a pull-up resistor. You can select whether or not to use by mask option.

# 4.1.2 Simultaneous LOW level input at input port terminals K00–K03

Another way of executing initial reset externally is to input a LOW level simultaneously to the input ports (K00–K03) selected by mask option.

Since there is a built-in time authorize circuit, be sure to maintain the designated input port terminal at LOW level for two seconds (when the oscillation frequency is fOSC1 = 32.768 kHz) or more to perform the initial reset by means of this function.

However, the time authorize circuit is bypassed during the SLEEP (standby) status and oscillation stabilization waiting period, and initial reset is executed immediately after the simultaneous LOW level input to the designated input ports. The combination of input ports (K00–K03) that can be selected by mask option are as follows:

Multiple key entry reset
□ Not use
□ K00 & K01
□ K00 & K01 & K02
☐ K00 & K01 & K02 & K03

For instance, let's say that mask option "K00 & K01 & K02 & K03" is selected.

When the input level at input ports K00–K03 is simultaneously LOW, initial reset will take place.

When using this function, make sure that the designated input ports do not simultaneously switch to LOW level while the system is in normal operation.

# 4.1.3 Supply voltage detection (SVD) circuit

When the SVD circuit detects that supply voltage has dropped below level 0 four successive times (see Chapter 8, "ELECTRICAL CHARACTERISTICS"), it outputs an initial reset signal until the supply voltage has been restored to level 2. You can select whether or not to use this SVD initial reset function by mask option. For more information, see "5.16 Supply Voltage Detection (SVD) Circuit" in this Manual.

Note: The supply voltage must be at least level 2 for the first sampling of the SVD circuit after an initial reset regardless of the mask option selected. At this time, if the power voltage level is less than level 2, the SVD circuit will continue sampling until the supply voltage reaches level 2 or more.

# 4.1.4 Initial reset sequence

After cancellation of the LOW level input to the RESET terminal, when the power is turned on, the start-up of the CPU is held back until the oscillation stabilization waiting time (8,192/fosc1 sec.) and the SVD initial sampling time (248/fosc1 sec.) have elapsed.

Figure 4.1.4.1 shows the operating sequence following initial reset release.

Also, when using the initial reset by simultaneous LOW level input into the input port, you should be careful of the following points.

- (1) During SLEEP status, since the time authorization circuit is bypassed, an initial reset is triggered immediately after a LOW level simultaneous input value. In this case, the CPU starts after waiting the oscillation stabilization time and the SVD circuit initial sampling time, following cancellation of the LOW level simultaneous input.
- (2) Other than during SLEEP status, an initial reset will be triggered 1–2 seconds after a LOW level simultaneous input. In this case, since a reset differential pulse (64/fosc1 sec.) is generated within the S1C88349, the CPU will start even if the LOW level simultaneous input status is not canceled.

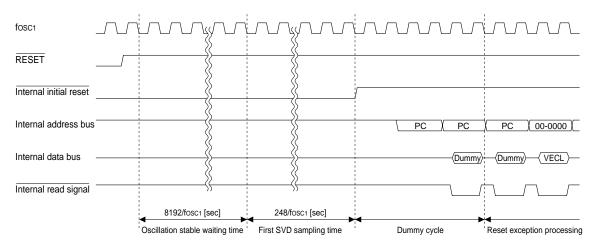


Fig. 4.1.4.1 Initial reset sequence

# 4.2 Initial Settings After Initial Reset

The CPU internal registers are initialized as follows during initial reset.

Table 4.2.1 Initial settings

Register name	Code	Bit length	Setting value
Data register A	A	8	Undefined
Data register B	В	8	Undefined
Index (data) register L	L	8	Undefined
Index (data) register H	Н	8	Undefined
Index register IX	IX	16	Undefined
Index register IY	IY	16	Undefined
Program counter	PC	16	Undefined*
Stack pointer	SP	16	Undefined
Base register	BR	8	Undefined
Zero flag	Z	1	0
Carry flag	C	1	0
Overflow flag	V	1	0
Negative flag	N	1	0
Decimal flag	D	1	0
Unpack flag	U	1	0
Interrupt flag 0	10	1	1
Interrupt flag 1	I1	1	1
New code bank register	NB	8	01H
Code bank register	СВ	8	Undefined*
Expand page register	EP	8	00H
Expand page register for IX	XP	8	00H
Expand page register for IY	YP	8	00H

<sup>\*</sup> Reset exception processing loads the preset values stored in 0 bank, 0000H–0001H into the PC. At the same time, 01H of the NB initial value is loaded into CB.

Initialize the registers which are not initialized at initial reset using software.

Since the internal RAM and display memory are not initialized at initial reset, be sure to initialize using software.

The respectively stipulated initializations are done for internal peripheral circuits. If necessary, the initialization should be done using software. For initial value at initial reset, see the sections on the I/O memory map and peripheral circuit descriptions in the following chapter of this Manual.

# 5 PERIPHERAL CIRCUITS AND THEIR OPERATION

The peripheral circuits of the S1C88349 is interfaced with the CPU by means of the memory mapped I/O method. For this reason, just as with other memory access operations, peripheral circuits can be controlled by manipulating I/O memory. Below is a description of the operation and control method for each individual peripheral circuit.

# 5.1 I/O Memory Map

Table 5.1.1(a) I/O Memory map (00FF00H-00FF02H, MCU mode)

Address	Bit	Name		F	unction		1	0	SR	R/W	Comment
00FF00		BSMD1	Bus mode (						0	R/W	
(MCU)			BSMD1		*	ode					
, ,			1	1	_	Maximum)					Do not set
	D6	BSMD0	1	0	•	Ainimum)			0	R/W	BSMD1-0 to 01B.
			0	1	×						
			0	0	Single c	hip					
	D5	CEMD1	R/W registe	er	~ · · · · · · · ·				1	R/W	
	D4	CEMD0	R/W registe	er					1	R/W	
	D3	CE3	CE3 (R33)	CE3 (R33) CE signal output Enable/Disable				CE3 disable	0	R/W	In the Single chip
	D2	CE2	CE2 (R32)	~			CE2 enable	CE2 disable	0	R/W	mode, these setting
	D1	CE1	CE1 (R31)		e: CE sign	•	CE1 enable	CE1 disable	0	R/W	are fixed at DC
	D0	CE0	CE0 (R30)_	Disabi	e: DC (R3:	x) output	CE0 enable	CE0 disable	0	R/W	output.
00FF01	D7	SPP7	Stack point	er page	address	(MSB)	1	0	0	R/W	
	D6	SPP6					1	0	0	R/W	
	D5	SPP5	< SP page a	llocatab	ole address >	1	0	0	R/W		
	D4	SPP4	Single chi	p mode	only 0 pag	1	0	0	R/W		
	D3	SPP3	• 64K mode	e:	only 0 pag	1	0	0	R/W		
	D2	SPP2	• 512K (mi	n) mode	: 0–27H pag	1	0	0	R/W		
	D1	SPP1	• 512K (ma	x) mode	e:0–27H pag	1	0	0	R/W		
	D0	SPP0				(LSB)	1	0	0	R/W	
00FF02	DZ	EBR	Bus release	enable	register	K11	BREQ	Input port	0	R/W	
	<i>D1</i>		(K11 and R	51 term	inal specific	cation) R51	BACK	Output port		IX/ VV	
			Wait contro	l registe	er	Number			0	R/W	
	D6	WT2	WT2	WT1	WT0	of state					
	L		1	1	1	14			L		
			1 1	1 0	0 1	12			0	R/W	
	D5	WT1	1	0	0	10 8					
	L		0	1	1	6					
			0	1	0	4			0	R/W	
	D4	WT0	0	0	1	2					
			0	0	0	No wait					
	D3	CLKCHG	CPU operat	ing cloc	ck switch		OSC3	OSC1	0	R/W	
	D2	oscc	OSC3 oscil	lation C	n/Off contro	ol	On	Off	0	R/W	
			Operating n	node se	lection				0	R/W	
	D1	VDC1	VDC1	VDC0		ing mode					
			1	×	High speed	d (VD1=3.3V)				R/W	
	DO	VDC0	0	1	Low powe	er (VD1=1.2V)			"	IN/ W	
	00	V DC0	0	0	Normal	(Vd1=2.2V)					
		1	1				I.			1	I.

Note: All the interrupts including \( \overline{NMI} \) are disabled, until you write the optional value into both the "00FF00H" and "00FF01H" addresses.

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Table 5.1.1(b) I/O Memory map (00FF00H-00FF02H, MPU mode)

Address	Bit	Name			unction	T ( *		1	0	SR	R/W	Comment
00FF00		BSMD1	Bus mode (							*	R/W	* Initial setting can
(MPU)			BSMD1			ode						be selected among 3
()			1	1	512K (M		_ n)					types (64K, 512K
	D6	BSMD0	1	0	512K (N					*	R/W	min and 512K max)
			0	1	64K						10	by mask option
			0	0	* Option	selectio	n 📥					setting.
	D5	CEMD1	Chip enable							1	R/W	Only for 64K
			CEMD1		Me	ode						bus mode
			1	1	64K (CE		_					
	D4	CEMD0	1	0	32K (CE					1	R/W	
			0	1	16K (CE							
			0	0	8K (CE							
	D3	CE3	CE3 (R33)	1				CE3 enable	CE3 disable	0	R/W	
		CE2	<del>CE2</del> (R32)		al output E			CE2 enable	CE2 disable	0	R/W	
		CE1	CE1 (R31)		CE signa	_		CE1 enable	CE1 disable	0	R/W	
		CE0	<del>CE0</del> (R30)	Disable	: DC (R3)	x) output		CE0 enable	CE0 disable	1	R/W	
00FF01		SPP7	Stack point	er page a	ddress	(MS	SB)	1	0	0	R/W	
		SPP6		1			,	1	0	0	R/W	
		SPP5	< SP page a	llocatabl	e address >	>		1	0	0	R/W	
		SPP4	• Single chi					1	0	0	R/W	
		SPP3	• 64K mode		only 0 pag			1	0	0	R/W	
	D2	SPP2	• 512K (min	n) mode:	0–27H pag	ge		1	0	0	R/W	
		SPP1	• 512K (ma	x) mode:	0–27H pag	ge		1	0	0	R/W	
		SPP0				(LS	B)	1	0	0	R/W	
00FF02	57	<b>EDD</b>	Bus release	enable re	egister		K11	BREQ	Input port			
	וטו	EBR	(K11 and R	51 termi	nal specific	cation)	R51	BACK	Output port	0	R/W	
			Wait contro	l register	•	Num	ber			0	R/W	
	D6	WT2	WT2	WT1	WT0	of st						
			1	1	1	14	1					
			1	1	0	12				0	R/W	
	D5	WT1	1 1	0	1 0	10						
			0	1	1	8						
			0	1	0	4				0	R/W	
	D4	WT0	0	0	1	2						
			0	0	0	No v	vait					
	D3 CLKCHG CPU operating clock switch						OSC3	OSC1	0	R/W		
	D2	oscc	OSC3 oscillation On/Off control					On	Off	0	R/W	
			Operating n	node sele	ection					0	R/W	
	D1	VDC1	VDC1	VDC0	Operat	ing mod	e					
			$\frac{\text{VDCI}}{1}$		High speed				<u> </u>			
			0		Low powe					0	R/W	
	D0	00 VDC0	0		Normal	(VD1=2						
			Ŭ	•		(121=2	,					

Note: All the interrupts including \( \overline{NMI} \) are disabled, until you write the optional value into both the "00FF00H" and "00FF01H" addresses.

Table 5.1.1(c) I/O Memory map (00FF09H-00FF13H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF09	D7	_	_	_	-	-		
	D6	_	_	_	-	_		Constantly "0" when
	D5	_	_	_	-	_		being read
	D4	_	_	-	-	_		
	D3	VCCHG	Reserved	1	0	0	R/W	
	D2	LCDB	Reserved	1	0	0	R/W	
	D1	LCDAJ	Power TYPE A (4.5V)/B (5.5V) switch	TYPE A	TYPE B	0	R/W	
	D0	DUTY8	LCD drive duty switch	1/8 duty	1/16, 1/32	0	R/W	*1
00FF10	D7	_	_	_	-	_		
	D6	_	_	_	-	_		Constantly "0" when
	D5	_	_	_	-	_		being read
	D4	LCCLK	CL output control for expanded LCD driver	On	Off	0	R/W	
	D3	LCFRM	FR output control for expanded LCD driver	On	Off	0	R/W	
	D2	DTFNT	LCD dot font selection	5 x 5 dots	5 x 8 dots	0	R/W	
	D1	LDUTY	LCD drive duty selection	1/16 duty	1/32 duty	0	R/W	*1
	D0	SGOUT	R/W register	1	0	0	R/W	Reserved register
00FF11	D7	_	_	-	-	_		"0" when being read
	D6	DSPAR	LCD display memory area selection	Display area 1	Display area 0	0	R/W	
	D5	LCDC1	LCD display control			0	R/W	
			LCDC1 LCDC0 LCD display					These bits are reset
			1 1 All LCDs lit					to (0, 0) when
	D4	LCDC0	1 0 All LCDs out			0	R/W	SLP instruction
			0 1 Normal display					is executed.
			0 0 Drive off					
	D3	LC3	LCD contrast adjustment			0	R/W	
	D2	LC2	LC3 LC2 LC1 LC0 Contrast 1 1 1 Dark			0	R/W	
	D1	LC1	1 1 1 1 Daix			0	R/W	
		LC0	: : : : : 0 0 0 0 Light			0	R/W	
00FF12	D7	LCU	0 0 0 C Light				K/W	C
001112	D6		_	_	_	_		Constantly "0" when
		SVDSP	SVD auto-sampling control			0	R/W	being read These registers are
	DS	34036	SVD auto-sampling control	On	Off	0	IN/ W	reset to "0" when
	DΛ	SVDON	SVD continuous sampling control/status R	Ducy	Doody	1 ,0*2	R/W	SLP instruction is
	D4	SVDOIN	W Continuous sampning control/status	Busy On	Ready Off	0	IX/ W	executed.
	DЗ	SVD3	SVD detection level	Oil	OII	X	R	*3
		SVD3	SVD3 SVD2 SVD1 SVD0 Detection level			X	R	3
		SVD2	1 1 1 1 Level 15 1 1 1 0 Level 14			X	R	
		SVD1				X	R	
00FF13	D7	_	0 0 0 0 Level 0	1	_	-		
001113	D6	_	_	_	_	_		Constantly "0" when
	D5		_	_		_		being read
	D3		_	_	_	_		oonig icad
			Comparator 1 On/Off control	On	Off	0	R/W	
			Comparator 1 On/Off control	On	Off	0	R/W	
		CMP1DT				0	R/W	
		CMP1D1	Comparator 1 data  Comparator 0 data	+>-	+<-	0	R	
	טט	CIVILADI	Comparator o data	+>-	+<-	U	I,	

<sup>\*1</sup> Writing "1" to DUTY8 (FF09•D0) disables 1/16 and 1/32 duty selection using LDUTY (FF10•D1).

<sup>\*2</sup> After initial reset, this status is set "1" until conclusion of hardware first sampling.

<sup>\*3</sup> Initial value is set according to the supply voltage detected at first hardware sampling after an initial reset. Until conclusion of first sampling, SVD0–SVD3 data are undefined.

Table 5.1.1(d) I/O Memory map (00FF20H-00FF25H)

Address	Bit	Name	Function	1		0	SR	R/W	Comment
00FF20	D7	PK01	V00 V07 :				0	D/W	
	D6	PK00	K00–K07 interrupt priority register	PK01	PK0	0	0	R/W	
	D5	PSIF1	Social intenfers interment missity resistor	PSIF1 PSW1			0	D/W	
	D4	PSIF0	Serial interface interrupt priority register	PTM1		•	0	R/W	
	D3	PSW1	Stomyyotoh timon intomynt majonity, no siston	1	1	Level 3 Level 2	0	R/W	
	D2	PSW0	Stopwatch timer interrupt priority register	0	1	Level 1	U	K/W	
	D1	PTM1	Clock timer interrupt priority register	0	0	Level 0	0	R/W	
	D0	PTM0	Clock times interrupt priority register				U	IX/ VV	
00FF21	D7	_	_	-		-	-		
	D6	_	_	-		_	_		Constantly "0" when
	D5	_	_	-		_	_		being read
	D4	_	_	_		-	-		
	D3	PPT1	Programmable timer interrupt priority register	PPT1 PK11			0	R/W	
	D2	PPT0	Trogrammable timer interrupt priority register	1	1	Level 3	U	IX/ VV	
	D1	PK11	K10 and K11 interrupt priority register	1 0	0	Level 2 Level 1	0	R/W	
	D0	PK10	KTO and KTT interrupt priority register	0	0	Level 0	U	IX/ VV	
00FF22	D7	_	_	-		-	-		"0" when being read
	D6	ESW100	Stopwatch timer 100 Hz interrupt enable register						
	D5	ESW10	Stopwatch timer 10 Hz interrupt enable register						
	D4	ESW1	Stopwatch timer 1 Hz interrupt enable register	Intom		Interrupt disable	0	R/W	
	D3	ETM32	Clock timer 32 Hz interrupt enable register	Inter	-				
	D2	ETM8	Clock timer 8 Hz interrupt enable register	enal	oie	disable			
	D1	ETM2	Clock timer 2 Hz interrupt enable register						
	D0	ETM1	Clock timer 1 Hz interrupt enable register						
00FF23	D7	EPT1	Programmable timer 1 interrupt enable register						
	D6	EPT0	Programmable timer 0 interrupt enable register	Interrupt					
	D5	EK1	K10 and K11 interrupt enable register			Interrupt	0	R/W	
	D4	EK0H	K04-K07 interrupt enable register						
	D3	EK0L	K00-K03 interrupt enable register	enal	ole	disable	U	IN/ W	
	D2	ESERR	Serial I/F (error) interrupt enable register						
	D1	ESREC	Serial I/F (receiving) interrupt enable register						
	D0	ESTRA	Serial I/F (transmitting) interrupt enable register						
00FF24	D7	-	_	-		-			"0" when being read
			Stopwatch timer 100 Hz interrupt factor flag	(R	.)	(R)			
		FSW10	Stopwatch timer 10 Hz interrupt factor flag	Intern	rupt	No interrupt			
	D4	FSW1	Stopwatch timer 1 Hz interrupt factor flag	facto	r is	factor is			
		FTM32	Clock timer 32 Hz interrupt factor flag	gener	ated	generated	0	R/W	
		FTM8	Clock timer 8 Hz interrupt factor flag	(W	7)	(W)			
		FTM2	Clock timer 2 Hz interrupt factor flag	Res		No operation			
		FTM1	Clock timer 1 Hz interrupt factor flag	Res		1.0 operation			
00FF25		FPT1	Programmable timer 1 interrupt factor flag	(R	.)	(R)			
		FPT0	Programmable timer 0 interrupt factor flag	Intern	rupt	No interrupt			
		FK1	K10 and K11 interrupt factor flag	facto	r is	factor is			
		FK0H	K04-K07 interrupt factor flag	gener	ated	generated	0	R/W	
		FK0L	K00-K03 interrupt factor flag				J	10, 11	
		FSERR	Serial I/F (error) interrupt factor flag	(W	7)	(W)			
		FSREC	Serial I/F (receiving) interrupt factor flag	Res	set	No operation			
	D0	FSTRA	Serial I/F (transmitting) interrupt factor flag						

Table 5.1.1(e) I/O Memory map (00FF28H-00FF31H)

Address	Bit	Name	Fui	nction	1	0	SR	R/W	Comment
00FF28	D7	PADC1	A/D converter interru	pt priority register	PADC1 PAD		0	R/W	
	D6	PADC0			$\begin{bmatrix} 1 & 1 \\ 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}$	Level 3 Level 2 Level 1 Level 0	0	R/W	
	D5	_	Reserved		Prohibited	-	0		Do not write "1".
	D4	_	Reserved		Prohibited	-	0		
	D3	_	_		-	-	_		
	D2	_	_		-	-	_		Constantly "0" when
	D1	_	_		-	-	_		being read
	D0	_	_		-	_	_		
00FF2A	D7	EAD	A/D converter interrup	t enable register	Enable	Disable	0	R/W	
	D6	_	Reserved		-	-	0	R/W	
	D5	_	_		-	-	_		
	D4	_	_		-	-	_		
	D3	_	_		-	-	_		Constantly "0" when
	D2	_	_		-	-	_		being read
	D1	_	_		-	_	_		
	D0	-	_		-	-	_		
00FF2C	D7	FAD	A/D converter interrup	ot factor flag R)	Generated	Not generated	0	R/W	
				W	Reset	No operation			
	D6	_	Reserved		-	-	0	R/W	
	D5	_	_		_	_	_		
	D4	_	_		_	_	_		
	D3	_	_		_	_	_		Constantly "0" when
	D2	_	_		_	_	_		being read
	D1	_	_		_	_	_		
	D0	_	_		_	_	-		
00FF30	D7	_	_		-	_	_		Constantly "0" when
	D6	_	_	-	-	-		_	
	D5	_	_		-	_	_		being read
	D4	MODE16	8/16-bit mode selection	on	16-bit x 1	8-bit x 2	0	R/W	
	D3	CHSEL	TOUT output channe	l selection	Timer 1	Timer 0	0	R/W	
	D2	PTOUT	TOUT output control		On	Off	0	R/W	
	D1	CKSEL1	Prescaler 1 source clo	ock selection	fosc3	foscı	0	R/W	
	D0	CKSEL0	Prescaler 0 source clo	ock selection	fosc3	foscı	0	R/W	
00FF31		EVCNT	Timer 0 counter mode	e selection	Event counter	Timer	0	R/W	
	D6	FCSEL	Timer 0	In timer mode	Pulse width	Normal	0	R/W	
			function selection		measurement	mode			
				In event counter mode	With	Without			
					noise rejector	noise rejector			
	D5	PLPOL	Timer 0	Down count timing	Rising edge	Falling edge	0	R/W	
			pulse polarity	in event counter mode	of K10 input	of K10 input			
			selection	In pulse width	High level measurement	Low level measurement			
				measurement mode	1	for K10 input			
	D4	PSC01	Timer 0 prescaler div	•			0	R/W	
			PSC01 PSC00	Prescaler dividing ratio					
			1 1	Source clock / 64					
	D3	PSC00	1 0	Source clock / 16			0	R/W	
			0 1	Source clock / 4					
			0 0	Source clock / 1					
		CONT0	Timer 0 continuous/o	ne-shot mode selection	Continuous	One-shot	0	R/W	
	D1 PSET0 Timer 0 preset			Preset	No operation	-	W	"0" when being read	
	D0	PRUN0	Timer 0 Run/Stop con	ntrol	Run	Stop	0	R/W	

Table 5.1.1(f) I/O Memory map (00FF32H-00FF36H)

Address	Bit	Name	Table 5.1.1(f) 1/O Memory map (00FF32H=00FF36H)  Function 1 0	SR	R/W	Comment
00FF32		_ Name		- -	12/77	Comment
001152	D6	_		_		Constantly "0" when
	D5	_		_		being read
		PSC11	Timer 1 prescaler dividing ratio selection	0	R/W	
		PSC10	PSC11 PSC10 Prescaler dividing ratio	0	R/W	
			1 1 Source clock / 64			
			1 0 Source clock / 16			
			0 1 Source clock / 4			
			0 0 Source clock / 1			
	D2	CONT1	Timer 1 continuous/one-shot mode selection Continuous One-shot	0	R/W	
	D1	PSET1	Fimer 1 preset Preset No operation	_	W	"0" when being read
	D0	PRUN1	Timer 1 Run/Stop control Run Stop	0	R/W	
00FF33	D7	RLD07	Timer 0 reload data D7 (MSB)		R/W	
	D6	RLD06	Fimer 0 reload data D6			
	D5	RLD05	Fimer 0 reload data D5			
	D4	RLD04	Fimer 0 reload data D4			
	D3	RLD03	Fimer 0 reload data D3	1		
	D2	RLD02	Fimer 0 reload data D2			
	D1	RLD01	Timer 0 reload data D1			
	D0	RLD00	Timer 0 reload data D0 (LSB)			
00FF34	D7	RLD17	Timer 1 reload data D7 (MSB)	1	R/W	
	D6	RLD16	Гimer 1 reload data D6			
	D5	RLD15	Гimer 1 reload data D5			
	D4	RLD14	Timer 1 reload data D4 High Low			
	D3	RLD13	Fimer 1 reload data D3			
		RLD12	Timer 1 reload data D2			
		RLD11	Timer 1 reload data D1			
		RLD10	Timer 1 reload data D0 (LSB)			
00FF35		PTD07	Γimer 0 counter data D7 (MSB)			
		PTD06	Timer 0 counter data D6			
		PTD05	Fimer 0 counter data D5	1	R	
		PTD04	Timer 0 counter data D4 Low			
		PTD03	Timer 0 counter data D3			
		PTD02	Timer 0 counter data D2			
		PTD01	Timer 0 counter data D1			
005500		PTD00	Timer 0 counter data D0 (LSB)			
00FF36		PTD17	Finer 1 counter data D7 (MSB)			
		PTD16	Timer 1 counter data D6			
		PTD15	Finer 1 counter data D5			
		PTD14	Finer 1 counter data D4  High Low	1	R	
		PTD13	Finant Logarton data D2			
		PTD12	Finant 1 counter data D1			
		PTD11	Finer 1 counter data D1			
	טט	PTD10	Timer 1 counter data D0 (LSB)			

Table 5.1.1(g) I/O Memory map (00FF40H–00FF44H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF40	D7	_	_	-	-	_		"0" when being read
	D6	FOUT2	FOUT frequency selection			0	R/W	
			FOUT2 FOUT1 FOUT0 Frequency					
			0 0 0 fosc1 / 1					
	D5	FOUT1	0 0 1 fosc1/2			0	R/W	
			0 1 0 fosc1 / 4 0 1 1 fosc1 / 8					
			1 0 0 fosc3 / 1					
	D4	FOUT0	1 0 1 fosc3/2			0	R/W	
			1 1 0 fosc3 / 4					
			1 1 1 fosc3 / 8					
	D3	FOUTON	FOUT output control	On	Off	0	R/W	
	D2	WDRST	Watchdog timer reset	Reset	No operation	_	W	Constantly "0" when
	D1	TMRST	Clock timer reset	Reset	No operation	_	W	being read
	D0	TMRUN	Clock timer Run/Stop control	Run	Stop	0	R/W	
00FF41		TMD7	Clock timer data 1 Hz					
	D6	TMD6	Clock timer data 2 Hz					
		TMD5	Clock timer data 4 Hz					
		TMD4	Clock timer data 8 Hz	High	Low	0	R	
		TMD3	Clock timer data 16 Hz	mgn	Low	Ü	``	
		TMD2	Clock timer data 32 Hz					
		TMD1	Clock timer data 64 Hz					
	D0	TMD0	Clock timer data 128 Hz					
00FF42	D7	-	_	-	-	_		
	D6	_	_	-	-	_		
	D5	-	_	-	-	_		Constantly "0" when
	D4	_	_	-	-			being read
	D3	_	_	-	-			
	D2	_	_	-	-			
		SWRST	Stopwatch timer reset	Reset	No operation	_	W	
			Stopwatch timer Run/Stop control	Run	Stop	0	R/W	
00FF43		SWD7	Stopwatch timer data					
		SWD6						
		SWD5	BCD (1/10 sec)					
		SWD4				0	R	
		SWD3	Stopwatch timer data					
		SWD2						
		SWD1	BCD (1/100 sec)					
005511		SWD0						
00FF44	D7	- D70TD	-	-		_		Constantly "0" when
		BZSTP	One-shot buzzer forcibly stop	Forcibly stop	-	_	W	being read
	ט5	BZSHT	One-shot buzzer trigger/status R	Busy Trigger	Ready No operation	0	R/W	
	D4	SHTPW	One-shot buzzer duration width selection	125 msec	31.25 msec	0	R/W	
		ENRTM	Envelope attenuation time	1 sec	0.5 sec	0	R/W	
		ENRST	Envelope reset	Reset	No operation	-	W	"0" when being read
		ENON	Envelope On/Off control	On	Off	0	R/W	*1
	D0	BZON	Buzzer output control	On	Off	0	R/W	

<sup>\*1</sup> Reset to "0" during one-shot output.

Table 5.1.1(h) I/O Memory map (00FF45H-00FF49H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF45	D7	_	_	-	-	_		"0" when being read
	D6	DUTY2	Buzzer signal duty ratio selection			0	R/W	
			DUTY2-0 Buzzer frequency (Hz)					
			2 1 0 4096.0 3276.8 2730.7 2340.6 2048.0 1638.4 1365.3 1170.3					
	D5	DUTY1	0 0 0 8/16 8/20 12/24 12/28			0	R/W	
			0 0 1 7/16 7/20 11/24 11/28 0 1 0 6/16 6/20 10/24 10/28					
			0 1 1 5/16 5/20 9/24 9/28					
	D4	DUTY0	1 0 0 4/16 4/20 8/24 8/28			0	R/W	
			1 0 1 3/16 3/20 7/24 7/28 1 1 0 2/16 2/20 6/24 6/28					
			1 1 1 1/16 1/20 5/24 5/28					
	D3	_	_	_	-	-		"0" when being read
	D2	BZFQ2	Buzzer frequency selection			0	R/W	
			BZFQ2 BZFQ1 BZFQ0 Frequency (Hz)					
			0 0 0 4096.0					
	D1	BZFQ1	$\begin{bmatrix} 0 & 0 & 1 & 3276.8 \\ 0 & 1 & 0 & 2730.7 \end{bmatrix}$			0	R/W	
			0 1 0 2730.7 0 1 1 2340.6					
			1 0 0 2048.0					
	D0	BZFQ0	1 0 1 1638.4			0	R/W	
			1 1 0 1365.3					
			1 1 1 1170.3					
00FF48	D7	_	_	-	-	_		"0" when being read
		EPR	Parity enable register	With parity	Non parity	0	R/W	Only for
		PMD	Parity mode selection	Odd	Even	0	R/W	asynchronous mode
	D4	SCS1	Clock source selection			0	R/W	
			SCS1 SCS0 Clock source					In the clock synchro-
			1 1 Programmable timer					nous slave mode,
	D3	SCS0	1 0 fosc3 / 4			0	R/W	external clock is
			0 1 fosc3 / 8					selected.
			0 0 fosc3 / 16					
	D2	SMD1	Serial I/F mode selection			0	R/W	
			SMD1 SMD0 Mode					
		0.450	1 1 Asynchronous 8-bit					
	D1	SMD0	1 0 Asynchronous 7-bit			0	R/W	
			0 1 Clock synchronous slave					
	Do	ESIF	0 0 Clock synchronous master	G : 177	1/0	0	D /337	
00FF49	_		Serial I/F enable register	Serial I/F	I/O port	0	R/W	"O"1 h-i 1
007749	D7	– FER	Framing arror flag	- E	No omen	-	D /XX7	"0" when being read
	סט	i-EK	Framing error flag R	Error	No error	0	R/W	
	DE	PER	Parity error flag R	Reset (0)	No operation	0	R/W	Only for
	DS	FLK	W W	Error	No error	U	IX/ VV	asynchronous mode
	D4	OER	i i	Reset (0)	No operation	0	R/W	asynchronous mode
	<b>υ</b> 4	OLK	Overrun error flag R	Error	No error	U	IX/ W	
	DЗ	RXTRG	Receive trigger/status R	Reset (0)	No operation Stop	0	R/W	
	DS	IXXIIXO	W W		No operation	U	IX/ VV	
	D2	RXEN	Receive enable	Trigger Enable	Disable	0	R/W	
	D1	TXTRG	Transmit trigger/status R	Run	Stop	0	R/W	
	٠,		W W	Trigger	No operation	J	10 11	
	DΩ	TXEN	Transmit enable	Enable	Disable	0	R/W	
	20	. / 1 4	Transmit endoic	Lindoic	Disable	J	10 11	

Table 5.1.1(i) I/O Memory map (00FF4AH-00FF54H)

Address	Bit	Name	Function Function	1	0	SR	R/W	Comment
00FF4A	D7	TRXD7	Transmit/Receive data D7 (MSB)					
	D6	TRXD6	Transmit/Receive data D6					
	D5	TRXD5	Transmit/Receive data D5					
	D4	TRXD4	Transmit/Receive data D4		_			
	D3	TRXD3	Transmit/Receive data D3	High	Low	X	R/W	
	D2	TRXD2	Transmit/Receive data D2					
	D1	TRXD1	Transmit/Receive data D1					
	D0	TRXD0	Transmit/Receive data D0 (LSB)					
00FF50	D7	SIK07	K07 interrupt selection register					
	D6	SIK06	K06 interrupt selection register					
	D5	SIK05	K05 interrupt selection register					
	D4	SIK04	K04 interrupt selection register	Interrupt	Interrupt			
		SIK03	K03 interrupt selection register	enable	disable	0	R/W	
		SIK02	K02 interrupt selection register					
		SIK01	K01 interrupt selection register					
		SIK00	K00 interrupt selection register					
00FF51	D7	_	_	_	_	_		
001101	D6	_	_	_	_	_		
	D5	_	_	_	_	_		Constantly "0" when
	D4	_			_	_		being read
	D3		_		_	_		being read
	D2		_			_		
		SIK11	V11 interpret calcution register	- Intomorat	Interrupt	_		
		SIK11	K11 interrupt selection register	Interrupt		0	R/W	
00FF52		KCP07	K10 interrupt selection register K07 interrupt comparison register	enable	disable			
00FF32		KCP07						
		KCP05	K06 interrupt comparison register	T	T			
		KCP03 KCP04	K05 interrupt comparison register	Interrupt	Interrupt			
			K04 interrupt comparison register	generated	generated	1	R/W	
		KCP03	K03 interrupt comparison register	at falling	at rising			
		KCP02	K02 interrupt comparison register	edge	edge			
		KCP01	K01 interrupt comparison register					
005550		KCP00	K00 interrupt comparison register					
00FF53	D7	_	_	_	-	-		
	D6	_	_	_	-	-		
	D5	_	_	_	_	_		Constantly "0" when
	D4		_	_	-	_		being read
	D3	_	_	_	-	_	-	
	D2	- KOD44	<del>-</del>	-	-	_		
		KCP11	K11 interrupt comparison register	Falling	Rising	1	R/W	
005557		KCP10	K10 interrupt comparison register	edge	edge			
00FF54		K07D	K07 input port data					
		K06D	K06 input port data					
	D5		K05 input port data					
		K04D	K04 input port data	High level	Low level	_	R	
		K03D	K03 input port data	input	input			
F -		K02D	K02 input port data					
		K01D	K01 input port data	.				
	D0	K00D	K00 input port data					

Table 5.1.1(j) I/O Memory map (00FF55H-00FF70H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF55	D7	-	_	-	-	-		
	D6	_	_	-	-	-		
	D5	_	_	-	-	-		Constantly "0" when
	D4	_	_	-	-	_		being read
	D3	_	_	-	-	-		
	D2	_	_	-	_	-		
	D1	K11D	K11 input port data	High level	Low level		R	
	D0	K10D	K10 input port data	input	input	_	K	
00FF60	D7	IOC07	P07 I/O control register					
	D6	IOC06	P06 I/O control register					
	D5	IOC05	P05 I/O control register					
	D4	IOC04	P04 I/O control register	0-44	T		R/W	
	D3	IOC03	P03 I/O control register	Output	Input	0		
	D2	IOC02	P02 I/O control register					
	D1	IOC01	P01 I/O control register					
	D0	IOC00	P00 I/O control register					
00FF61	D7	IOC17	P17 I/O control register					
	D6	IOC16	P16 I/O control register					
	D5	IOC15	P15 I/O control register					
<del>-</del>		IOC14	P14 I/O control register		T		D /33/	
- F		IOC13	P13 I/O control register	Output	Input	0	R/W	
H		IOC12	P12 I/O control register					
H		IOC11	P11 I/O control register					
	D0	IOC10	P10 I/O control register					
	_	P07D	P07 I/O port data					
	D6	P06D	P06 I/O port data					
	D5	P05D	P05 I/O port data				R/W	
	D4	P04D	P04 I/O port data	High	Low			
	D3	P03D	P03 I/O port data			1		
	D2	P02D	P02 I/O port data					
	D1	P01D	P01 I/O port data					
	D0	P00D	P00 I/O port data					
00FF63	D7	P17D	P17 I/O port data					
	D6	P16D	P16 I/O port data					
	D5	P15D	P15 I/O port data					
	D4	P14D	P14 I/O port data					
	D3	P13D	P13 I/O port data	High	Low	1	R/W	
H		P12D	P12 I/O port data					
H		P11D	P11 I/O port data					
H		P10D	P10 I/O port data					
		HZR51	R51 high impedance control	High	Comple-			
H		HZR50	R50 high impedance control	impedance	mentary	0	R/W	
H		HZR4H	R/W register	1		_		
F		HZR4L	R/W register	1	0	0	R/W	Reserved register
H		HZR1H	R14–R17 high impedance control					
H		HZR1L	R10–R13 high impedance control	High	Comple-	_		
D1		HZR0H	R04–R07 high impedance control	impedance	mentary	0	R/W	
	HZR0L	R00–R03 high impedance control	Impedance	mentary				

Table 5.1.1(k) I/O Memory map (00FF71H-00FF76H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF71	D7	HZR27	R27 high impedance control					
	D6	HZR26	R26 high impedance control					
	D5	HZR25	R25 high impedance control					
	D4	HZR24	R24 high impedance control	High	Comple-		D/XX	
	D3	HZR23	R23 high impedance control	impedance	mentary	0	R/W	
	D2	HZR22	R22 high impedance control					
	D1	HZR21	R21 high impedance control					
	D0	HZR20	R20 high impedance control					
00FF72	D7	HZR37	R37 high impedance control					
	D6	HZR36	R36 high impedance control					
	D5	HZR35	R35 high impedance control					
	D4	HZR34	R34 high impedance control	High	Comple-		D/XX	
	D3	HZR33	R33 high impedance control	impedance	mentary	0	R/W	
	D2	HZR32	R32 high impedance control					
	D1	HZR31	R31 high impedance control					
	D0	HZR30	R30 high impedance control					
00FF73	D7	R07D	R07 output port data					
	D6	R06D	R06 output port data					
	D5	R05D	R05 output port data					
	D4	R04D	R04 output port data	-		١.		
	D3	R03D	R03 output port data	High	Low	1	R/W	
	D2	R02D	R02 output port data					
	D1	R01D	R01 output port data	•				
	D0	R00D	R00 output port data					
00FF74	D7	R17D	R17 output port data					
	D6	R16D	R16 output port data					
	D5	R15D	R15 output port data					
	D4	R14D	R14 output port data		_			
	D3	R13D	R13 output port data	High	Low	1	R/W	
	D2	R12D	R12 output port data					
	D1	R11D	R11 output port data					
	D0	R10D	R10 output port data					
00FF75	D7	R27D	R27 output port data					
	D6	R26D	R26 output port data					
	D5	R25D	R25 output port data					
	D4	R24D	R24 output port data		Į,	1	D/XX	
	D3	R23D	R23 output port data	High	Low	1	R/W	
		R22D	R22 output port data					
	D1	R21D	R21 output port data					
	D0	R20D	R20 output port data					
00FF76		R37D	R37 output port data					
	D6	R36D	R36 output port data					
		R35D	R35 output port data	1				
		R34D	R34 output port data		Į į		D 411	
		R33D	R33 output port data	High	Low	1	R/W	
D2 D1		R32D	R32 output port data					
		R31D	R31 output port data					
	R30D	R30 output port data	1					

Table 5.1.1(1) I/O Memory map (00FF77H-00FF82H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF77	D7	R47D	R/W register					
	D6	R46D	R/W register					
	D5	R45D	R/W register					
		R44D	R/W register					
	D3	R43D	R/W register	1	0	1	R/W	Reserved register
		R42D	R/W register					
	D1	R41D	R/W register					
		R40D	R/W register					
00FF78	D7	_	_	_	_	_		
	D6	_	_	_	_	_		
	D5		_	_	_	_		Constantly "0" when
	D4		_	_	_	_		being read
	D3		_	_	_	_		Joenng rout
	D2	_	_	_	_	_		
		R51D	R51 output port data	High	Low	1	R/W	
		R50D	R50 output port data	High	Low	0	R/W	
00FF80	D7	_		-	_	_	10 11	
001100	D6	_		_				Constantly "0" when
	D5	_		_		_		being read
	D3		_	_	_	_		being read
		PRAD	A/D converter clock control		- Off	0	R/W	
		PSAD2	A/D converter division ratio	On	Off	0	R/W	
	DZ	PSAD2				0	K/W	
			$\frac{PSAD2}{1} \frac{PSAD1}{1} \frac{PSAD0}{1} \frac{Division ratio}{fosc1 / 1}$					
	D4	PSAD1	1 1 0 fosc3 / 64				D/W	
	וט	PSADI	1 0 1 fosc3 / 32			0	R/W	
			1 0 0 fosc3 / 16					
		PSAD0	0 1 1 fosc3 / 8				D /XX	
	טט	PSADU	0 1 0 fosc3 / 4 0 0 1 fosc3 / 2			0	R/W	
			0 0 1 103C3 / 2 0 0 fosc3 / 1					
00FF81	D7	PAD7	P17 A/D converter input control			0	R/W	
001101	_	PAD6	P16 A/D converter input control	A/D		0	R/W	
		PAD5	P15 A/D converter input control	converter	I/O port	0	R/W	
		PAD4	P14 A/D converter input control	input		0	R/W	
	D3	_	-	_	_	0	10/11	
	D2	_	_					Constantly "0" when
	D1	_	_	_	_	_		being read
	D0			_				being read
00FF82		ADRUN	A/D conversion start control register	Start	Invalid	0	W	
001102	D6			– Start	-	-	<b>**</b>	
	D5		_	_	_			Constantly "0" when
	D3		_					being read
	D3		_		_	_		ocing read
	D3 D2			_	_	_		
		CHS1	Analog input channel selection	_	_	_	D /337	
	וט	СПОТ				0	R/W	
			CHS1 CHS0 Input channel					
		CUCO	1 1 AD7				D /337	
	טט	CHS0	1 0 AD6			0	R/W	
			0 1 AD5					
			0 0 AD4					

#### 5 PERIPHERAL CIRCUITS AND THEIR OPERATION (I/O Memory Map)

Table 5.1.1(m) I/O Memory map (00FF83H-00FF84H)

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF83	D7	ADDR9	A/D conversion result D9 (MSB)					_
	D6	ADDR8	A/D conversion result D8					
	D5	ADDR7	A/D conversion result D7					
	D4	ADDR6	A/D conversion result D6				D.	
	D3	ADDR5	A/D conversion result D5			_	R	
	D2	ADDR4	A/D conversion result D4					
	D1	ADDR3	A/D conversion result D3					
	D0	ADDR2	A/D conversion result D2					
00FF84	D7	_	_	-	-	_		
	D6	_	_	-	-	_		
	D5	_	_	-	-	_		Constantly "0" when
	D4	_	_	-	-	_		being read
	D3	_	_	-	-	_		
	D2	_	_	-	-	_		
	D1 ADDR1 A/D convers		A/D conversion result D1				В	
	D0	ADDR0	A/D conversion result D0 (LSB)			-	R	

#### 5.2 System Controller and Bus Control

The system controller is a management unit which sets such items as the bus mode in accordance with memory system configuration factors. For the purposes of controlling the system, the following settings can be performed in software:

- (1) Bus mode (CPU mode) settings
- (2) Chip enable (CE) signal output settings
- (3) WAIT state settings for external memory
- (4) Bus authority release request / acknowledge signal (BREQ/BACK) settings
- (5) Page address setting of the stack pointer

Below is a description of the how these settings are to be made.

#### 5.2.1 Bus mode settings

As explained in "3.5.2 Bus mode", the S1C88349 has four bus modes. Settings for bus modes must be made in software and must match the capacity of the external memory.

As shown in Table 5.2.1.1, bus mode settings are performed on the basis of the preset values for each mode written to the registers BSMD0 and BSMD1.

Table 5.2.1.1 Bus mode settings

Setting	g value	Bus mode	Configuration of outernal mamory					
BSMD1	BSMD0	Bus mode	Configuration of external memory					
1	1	Expanded 512K maximum mode	ROM+RAM>64K bytes (Program>64K bytes)					
1	0	Expanded 512K minimum mode	ROM+RAM>64K bytes (Program≤64K bytes)					
0	1	Expanded 64K mode	ROM+RAM≤64K bytes					
0	0	Single chip mode (MCU)	None					
		Optional setting of one of the expanded	See above					
		modes (MPU)						

\* The single chip mode setting is only possible when this IC is used in the MCU mode. The single chip mode setting is incompatible with the MPU mode, since this mode does not utilize internal ROM.

When using in the MPU mode, it is necessary to select the bus mode at the time of the initial resetting and at the time of the <BSMD1 = "0" and BSMD0 = "0"> setting from among the three types of expanded modes (expanded 64K mode, expanded 512K minimum mode and expanded 512K maximum mode) by mask option. Select the expanded 512K maximum mode for this option, when the MPU mode is not used at all.

The function of I/O terminals is set as shown in Table 5.2.1.2 in accordance with mode selection.

Table 5.2.1.2 I/O terminal settings

	1 0000 0.211	Bus mode				
Terminal	Single chip	Expanded 64K mode	Expanded 512K mode			
R00	Output port R00	Address	s bus A0			
R01	Output port R01	Address	s bus A1			
R02	Output port R02	Address	s bus A2			
R03	Output port R03	Address	s bus A3			
R04	Output port R04	Address	s bus A4			
R05	Output port R05	Address bus A5				
R06	Output port R06	Address bus A6				
R07	Output port R07	Address bus A7				
R10	Output port R10	Address bus A8				
R11	Output port R11	Address bus A9				
R12	Output port R12	Address bus A10				
R13	Output port R13	Address bus A11				
R14	Output port R14	Address bus A12				
R15	Output port R15	Address	bus A13			
R16	Output port R16	Address	bus A14			
R17	Output port R17	Address	bus A15			
R20	Output 1	port R20	Address bus A16			
R21	Output 1	port R21	Address bus A17			
R22	Output 1	port R22	Address bus A18			
R23	Output port R23	$\overline{ ext{RD}}$ s	ignal			
R24	Output port R24	WR s	signal			
P00	I/O port P00	Data b	ous D0			
P01	I/O port P01	Data b	ous D1			
P02	I/O port P02	Data b	ous D2			
P03	I/O port P03	Data b	ous D3			
P04	I/O port P04	Data b	ous D4			
P05	I/O port P05	Data b	ous D5			
P06	I/O port P06	Data b	ous D6			
P07	I/O port P07	Data b	ous D7			

At initial reset, the bus mode is set as explained below.

#### In MCU mode:

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At initial reset, the S1C88349 is set in single chip mode.

Accordingly, in MCU mode, even if a memory has been externally expanded, the system is activated by the program written to internal ROM.

In the system with externally expanded memory, perform the applicable bus mode settings during the initialization routine originating in internal ROM.

#### In MPU mode:

When the MPU mode is used, the expanded mode (expanded 64K mode, expanded 512K minimum mode or expanded 512K maximum mode) set during initial reset must be preselected by mask option.

You should set it to conform properly to system configuration.

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#### 5.2.2 Address decoder ( $\overline{CE}$ output) settings

As explained in Section 3.6.4, the S1C88349 is equipped with address decoders that can output a maximum of four chip enable signals  $(\overline{\text{CE0}}-\overline{\text{CE3}})$  to external devices.

The output terminals and output circuits for  $\overline{\text{CE0}}$ – $\overline{\text{CE3}}$  are shared with output ports R30–R33. At initial reset, they are set as output port terminals. For this reason, when operating in a mode other than single chip mode, the ports to be used as  $\overline{\text{CE}}$  signal output terminals must be set as such.

This setting is performed through software which writes "1" to registers CE0–CE3 corresponding the  $\overline{\text{CE}}$  signals to be used.

Table 5.2.2.1 shows the address range assigned to the four chip enable  $(\overline{\text{CE}})$  signals.

The arrangement of memory space for external devices does not necessarily have to be continuous from a subordinate address and any of the chip enable signals can be used to assign areas in memory. However, in the MPU mode, program memory must be assigned to  $\overline{\text{CE0}}$ .

In the expanded 512K mode, the address range of each of the  $\overline{CE}$  signals is fixed. In the expanded 64K mode, the four address ranges, which match the amount of memory in use, are selected with registers CEMD0 and CEMD1.

These signals are only output when the appointed external memory area is accessed and are not output when internal memory is accessed.

#### Table 5.2.2.1 Address settings of $\overline{CE0}$ – $\overline{CE3}$

#### (1) Expanded 64K mode (MPU mode only)

CEMD1	CEMD0	Chip size	CE0	CE1	CE2	CE3
1	1	64K bytes	000000H-00EFFFH	-	-	-
1	0	32K bytes	000000H-007FFFH	008000H-00EFFFH	-	-
0	1	16K bytes	000000H-003FFFH	004000H-007FFFH	008000H-00BFFFH	00C000H-00EFFFH
0	0	8K bytes	000000H-001FFFH	002000H-003FFFH	004000H-005FFFH	006000H-007FFFH

#### (2) Expanded 512K minimum/maximum modes

CE signal	Addres	s range
CE signal	MCU mode	MPU mode
CE0	200000H-27FFFH	000000H-00EFFFH, 010000H-07FFFFH
CE1	080000H-0FFFFH	080000H-0FFFFH
CE2	100000H-17FFFFH	100000H-17FFFFH
CE3	180000H-1FFFFH	180000H-1FFFFH

#### 5.2.3 WAIT state settings

In order to insure accessing of external low speed devices during high speed operations, the S1C88349 is equipped with a WAIT function which prolongs access time.

The number of wait states inserted can be selected from a choice of eight as shown in Table 5.2.3.1 by means of registers WT0–WT2.

Table 5.2.3.1 Setting the number of WAIT states

WT2	WT1	WT0	Number of inserted states
1	1	1	14
1	1	0	12
1	0	1	10
1	0	0	8
0	1	1	6
0	1	0	4
0	0	1	2
0	0	0	No wait

<sup>\*</sup> A state is 1/2 cycles of the clock in length.

WAIT states set in software are inserted between bus cycle states T3–T4.

Note, however, that WAIT states cannot be inserted when an internal register and internal memory are being accessed and when operating with the OSC1 oscillation circuit (see "5.4 Oscillation Circuits and Operating Mode").

Consequently, WAIT state settings in single chip mode are meaningless.

With regard to WAIT insertion timing, see Section 3.6.5, "WAIT control".

# 5.2.4 Setting the bus authority release request signal

With systems performing DMA transfer, the bus authority release request signal ( $\overline{BREQ}$ ) input terminal and acknowledge signal ( $\overline{BACK}$ ) output terminal have to be set.

The  $\overline{BREQ}$  input terminal is shared with input port terminal K11 and the  $\overline{BACK}$  output terminal with output port terminal R51. At initial reset, these terminal facilities are set as input port terminal and output port terminal, respectively. The terminals can be altered to function as  $\overline{BREQ}/\overline{BACK}$  terminals by writing a "1" to register EBR.

For details on bus authority release, see "3.6.6 Bus authority release state" and "S1C88 Core CPU Manual".

#### 5.2.5 Stack page setting

Although the stack area used to evacuate registers during subroutine calls can be arbitrarily moved to any area in data RAM using the stack pointer SP, its page address is set in registers SPP0–SPP7 in I/O memory.

At initial reset, SPP0-SPP7 are set to "00H" (page 0).

Since the internal RAM is arranged on page 0 (00F000H–00F7FFH), the stack area in single chip mode is inevitably located in page 0.

In expanded 64K mode where RAM is externally expanded, stack page is likewise limited to page 0. In order to place the stack area at the final address in internal RAM, the stack pointer SP is placed at an initial setting of "F800H". (SP is pre-decremented.)

In the expanded 512K mode, to place the stack in external expanded RAM, set a corresponding page to SPP0–SPP7. The page addresses to which SPP0–SPP7 can be set are 00H–27H and must be within a RAM area.

A page is each recurrent 64K division of data memory beginning at address zero.

#### 5.2.6 Control of system controller

Table 5.2.6.1 shows the control bits for the system controller.

Table 5.2.6.1(a) System controller control bits (MCU mode)

Address	Bit	Name		F	unction		1	0	SR	R/W	Comment
00FF00	D7	BSMD1	Bus mode (						0	R/W	00
(MCU)			BSMD1			ode					
` ,			1	1		Maximum)					Do not set
	D6	BSMD0	1	0	512K (N	Minimum)			0	R/W	BSMD1-0 to 01B.
			0	1	×	ŕ					
			0	0	Single c	hip					
	D5	CEMD1	R/W registe	r		<u> </u>			1	R/W	
	D4	CEMD0	R/W registe	r					1	R/W	
	D3	CE3	<del>CE3</del> (R33)	1			CE3 enable	CE3 disable	0	R/W	In the Single chip
	D2	CE2	<del>CE2</del> (R32)	CE sig	_	Enable/Disable	CE2 enable	CE2 disable	0	R/W	mode, these setting
	D1	CE1	CE1 (R31)		: CE sign	_	CE1 enable	CE1 disable	0	R/W	are fixed at DC
	D0	CE0	<del>CE0</del> (R30)	Disable	e: DC (R3	x) output	CE0 enable	CE0 disable	0	R/W	output.
00FF01	D7	SPP7	Stack pointe	er page a	address	(MSB)	1	0	0	R/W	-
	D6	SPP6	1				1	0	0	R/W	
	D5	SPP5	< SP page a	llocatab	le address :	>	1	0	0	R/W	
	D4	SPP4	Single chip	p mode:	only 0 pag	1	0	0	R/W		
	D3	SPP3	• 64K mode	:	only 0 pag	ge	1	0	0	R/W	
	D2	SPP2	• 512K (mir	n) mode	: 0–27H pa	ge	1	0	0	R/W	
	D1	SPP1	• 512K (max	x) mode	:0–27H pa	ge	1	0	0	R/W	
	D0	SPP0				(LSB)	1	0	0	R/W	
00FF02	D7 EBR		Bus release	enable 1	register	K11	BREQ	Input port	_	D/XI	
	וט	EBR	(K11 and R	51 term	inal specific	cation) R51	BACK	Output port	0	R/W	
			Wait contro	l registe	r	Number			0	R/W	
	D6	WT2	WT2	WT1	WT0	of state					
			1	1	1	14					
			1 1	1 0	0 1	12			0	R/W	
	D5	WT1	1 1	0	0	10 8					
			0	1	1	6					
			0	1	0	4			0	R/W	
	D4	WT0	0	0	1	2					
			0	0	0	No wait					
	D3	CLKCHG	CPU operat	ing cloc	k switch		OSC3	OSC1	0	R/W	
	D2	oscc	OSC3 oscill	lation O	n/Off contr	ol	On	Off	0	R/W	
			Operating n	node sel	ection				0	R/W	
	D1	VDC1	VDC1	VDC0	Operat	ting mode					
			1	×		d (VD1=3.3V)				<u></u>	
			0	1		er (VD1=1.2V)			0	R/W	
	D0	VDC0	0	0	Normal	(VD1=2.2V)					

Note: All the interrupts including \( \overline{NMI} \) are disabled, until you write the optional value into both the "00FF00H" and "00FF01H" addresses.

Table 5.2.6.1(b) System controller control bits (MPU mode)

Address	Bit	Name		Fu	nction		1	0	SR	R/W	Comment
00FF00	D7	BSMD1	Bus mode (	CPU mod	le)				*	R/W	* Initial setting can
(MPU)			BSMD1	BSMD0	Mo	de					be selected among 3
			1	1	512K (M	aximum)					types (64K, 512K
	D6	BSMD0	1	0	512K (M				*	R/W	min and 512K max)
			0	1	64K						by mask option
			0	0	* Option	selection 🗸					setting.
	D5	CEMD1	Chip enable	mode					1	R/W	Only for 64K
			CEMD1		Mo	de					bus mode
			1	1	64K (CE	<u>(0)</u>					
	D4	CEMD0	1	0	32K (CE				1	R/W	
			0	1	16K (CE						
			0	0	8K (CE						
	D3	CE3	CE3 (R33)	1	<u> </u>	·	CE3 enable	CE3 disable	0	R/W	
	D2	CE2	<del>CE2</del> (R32)	_	_	nable/Disable	CE2 enable	CE2 disable	0	R/W	
		CE1	CE1 (R31)		CE signa	_	CE1 enable	CE1 disable	0	R/W	
		CE0	CE0 (R30)_	Disable:	DC (R3x	) output	CE0 enable	CE0 disable	1	R/W	
00FF01	_	SPP7	Stack points	er page ac	ldress	(MSB)	1	0	0	R/W	
	D6	SPP6	•	1 0		, ,	1	0	0	R/W	
		SPP5	< SP page a	llocatable	e address >	1	0	0	R/W		
		SPP4	• Single chi				1	0	0	R/W	
		SPP3	• 64K mode		only 0 page		1	0	0	R/W	
		SPP2	• 512K (mir	n) mode:	0–27H pag	e	1	0	0	R/W	
	D1	SPP1	• 512K (ma				1	0	0	R/W	
		SPP0	,		1 0	(LSB)	1	0	0	R/W	
00FF02			Bus release	enable re	gister	K11	BREQ	Input port			
	טו	EBR	(K11 and R	51 termin	al specifica	ation) R51	BACK	Output port	0   R/W		
			Wait contro	l register		Number			0	R/W	
	D6	WT2	WT2	WT1	WT0	of state					
			1	1	1	14					
			1	1	0	12			0	R/W	
	D5	WT1	1 1	0	1 0	10					
			0	1	1	8 6					
			0	1	0	4			0	R/W	
	D4	WT0	0	0	1	2					
			0	0	0	No wait					
'	D3	CLKCHG	CPU operat	ing clock	switch		OSC3	OSC1	0	R/W	
	D2	oscc	OSC3 oscil	lation On	Off contro	1	On	Off	0	R/W	
			Operating n	node sele	ction				0	R/W	
	D1	VDC1	VDC1	VDC0	Operati	ng mode					
			1			(VD1=3.3V)				L	
			0			(VD1=3.3V) (VD1=1.2V)			0	R/W	
	D0	VDC0	0		Low power Normal	(VD1=1.2V) (VD1=2.2V)					
			U	0	Tormar	(*DI-2.2*)					

Note: All the interrupts including \( \overline{NMI} \) are disabled, until you write the optional value into both the "00FF00H" and "00FF01H" addresses.

#### BSMD0, BSMD1: 00FF00H•D6, D7

Bus modes are set as shown in Table 5.2.6.2.

Table 5.2.6.2 Bus mode settings

Setting	values	- Bus mode			
BSMD1	BSMD0				
1	1	Expanded 512K maximum mode			
1	0	Expanded 512K minimum mode			
0	1	Expanded 64K mode			
0	0	Single chip mode (MCU)			
		Optional setting of one of the			
		expanded modes (MPU)			

The single chip mode setting is only possible when this IC is used in the MCU mode. The single chip mode setting is incompatible with the MPU mode, since this mode does not utilize internal ROM. When using in the MPU mode, it is necessary to select the bus mode at the time of the initial resetting and at the time of the <BSMD1 = "0" and BSMD0 = "0"> setting from among the three types of expanded modes (expanded 64K mode, expanded 512K minimum mode and expanded 512K maximum mode) by mask option.

Select the expanded 512K maximum mode for this option, when the MPU mode is not used at all. At initial reset, in the MCU mode the unit is set to single chip mode and in the MPU mode the mask option is used to select the applicable mode.

#### CEMD0, CEMD1: 00FF00H•D4, D5

Sets the  $\overline{\text{CE}}$  signal address range (valid only in the expanded 64K mode\*).

Settings are made according to external memory chip size as shown in Table 5.2.6.3.

Table 5.2.6.3  $\overline{CE}$  signal settings

CEMD1	CEMD0	Address range	Usable terminals
1	1	64K bytes	CE0
1	0	32K bytes	CEO, CE1
0	1	16K bytes	CEO-CE3
0	0	8K bytes	CEO-CE3

These settings are invalid for any mode other than expanded  $64 \mathrm{K}$  mode.

At initial reset, each register is set to "1" (64K bytes).

\* Settings of these registers are valid only in the MPU mode. CEMD0 and CEMD1 can be used as general purpose registers with read/write capabilities in the MCU mode.

#### *CE0-CE3: 00FF00H•D0-D3*

Sets the  $\overline{\text{CE}}$  output terminals being used.

When "1" is written:  $\overline{CE}$  output enable When "0" is written:  $\overline{CE}$  output disable

Reading: Valid

 $\overline{\text{CE}}$  output is enabled when a "1" is written to registers CE0–CE3 which correspond to the  $\overline{\text{CE}}$  output being used. A "0" written to any of the registers disables  $\overline{\text{CE}}$  signal output from that terminal and it reverts to its alternate function as an output port terminal (R30–R33).

At initial reset, register CE0 is set to "0" in the MCU mode and in the MPU mode, "1" is set in the register. Registers CE1–CE3 are always set to "0" regardless of the MCU/MPU mode setting.

Note: To avoid a malfunction from an interrupt generated before the bus configuration is initialized, all interrupts including NMI are masked until you write an optional value into address "00FF00H".

#### SPP0-SPP7: 00FF01H

Sets the page address of stack area. In single chip mode and expanded 64K mode, set page address to "00H".

In expanded 512K mode, it can be set to any value within the range "00H"-"27H".

Since a carry and borrow from/to the stack pointer SP is not reflected in register SPP, the upper limit on continuous use of the stack area is 64K bytes. At initial reset, this register is set to "00H" (page 0).

Note: To avoid a malfunction from an interrupt generated before the bus configuration is initialized, all interrupts including \overline{NMI} are disabled, until you write an optional value into "00FF01H" address. Furthermore, to avoid generating an interrupt while the stack area is being set, all interrupts including \overline{NMI} are disabled in one instruction execution period after writing to address "00FF01H".

#### WT0-WT2: 00FF02H•D4-D6

How WAIT state settings are performed. The number of WAIT states to be inserted based on register settings is as shown in Table 5.2.6.4.

Table 5.2.6.4 Setting WAIT states

WT2	WT1	WT0	Number of inserted states
1	1	1	14
1	1	0	12
1	0	1	10
1	0	0	8
0	1	1	6
0	1	0	4
0	0	1	2
0	0	0	No wait

<sup>\*</sup> A state is 1/2 cycles of the clock in length.

At initial reset, this register is set to "0" (no wait).

#### EBR: 00FF02H•D7

Sets the  $\overline{BREQ}/\overline{BACK}$  terminals function.

When "1" is written:  $\overline{BREQ}/\overline{BACK}$  enabled When "0" is written:  $\overline{BREQ}/\overline{BACK}$  disabled

Reading: Valid

How  $\overline{BREQ}$  and  $\overline{BACK}$  terminal functions are set. Writing "1" to EBR enables  $\overline{BREQ}/\overline{BACK}$  input/output. Writing "0" sets the  $\overline{BREQ}$  terminal as input port terminal K11 and the  $\overline{BACK}$  terminal as output port terminal R51.

At initial reset, EBR is set to "0"  $(\overline{BREQ}/\overline{BACK})$  disabled).

#### 5.2.7 Programming notes

- (1) All the interrupts including  $\overline{NMI}$  are masked, until you write the optional value into both the "00FF00H" and "00FF01H" addresses. Consequently, even if you do not change the content of this address (You use the initial value, as is.), you should still be sure to perform the writing operation using the initialization routine.
- (2) When setting stack fields, including page addresses as well, you should write them in the order of the register SPP ("00FF01H") and the stack pointer SP.

Example: When setting the "178000H" address

LD EP, #00H

LD HL, #0FF01H

LD [HL], #17H During this period the interrupts (including NMI) are masked.

#### 5.3 Watchdog Timer

#### 5.3.1 Configuration of watchdog timer

The S1C88349 is equipped with a watchdog timer driven by OSC1 as source oscillation. The watchdog timer must be reset periodically in software, and if reset of more than 3–4 seconds (when fosc1 = 32.768 kHz) does not take place, a non-maskable interrupt signal is generated and output to the CPU.

Figure 5.3.1.1 is a block diagram of the watchdog timer.

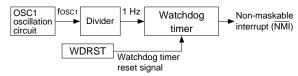


Fig. 5.3.1.1 Block diagram of watchdog timer

By running watchdog timer reset during the main routine of the program, it is possible to detect program runaway as if watchdog timer processing had not been applied. Normally, this routine is integrated at points that are regularly being processed.

The watchdog timer continues to operate during HALT and when a HALT state is continuous for longer than 3–4 seconds, the CPU shifts to exception processing.

During SLEEP, the watchdog timer is stopped.

D2 WDRST Watchdog timer reset

Clock timer reset

D0 TMRUN Clock timer Run/Stop control

D1 TMRST

#### 5.3.2 Interrupt function

In cases where the watchdog timer is not periodically reset in software, the watchdog timer outputs an interrupt signal to the CPU's  $\overline{\text{NMI}}$  (level 4) input. Unmaskable and taking priority over other interrupts, this interrupt triggers the generation of exception processing. See the "S1C88 Core CPU Manual" for more details on  $\overline{\text{NMI}}$  exception processing.

This exception processing vector is set at 000004H.

#### 5.3.3 Control of watchdog timer

Table 5.3.3.1 shows the control bits for the watchdog timer.

#### WDRST: 00FF40H•D2

Resets the watchdog timer.

When "1" is written: Watchdog timer is reset When "0" is written: No operation

Reading: Constantly "0"

By writing "1" to WDRST, the watchdog timer is reset, after which it is immediately restarted. Writing "0" will mean no operation. Since WDRST is for writing only, it is constantly set to "0" during readout.

#### 5.3.4 Programming notes

- (1) The watchdog timer must reset within 3-second cycles by software.
- (2) Do not execute the SLP instruction for 2 msec after a  $\overline{NMI}$  interrupt has occurred (when fosci is 32.768 kHz).

W

W

R/W

Constantly "0" when

being read

Address	Bit	Name			Function	1	1	0	SR	R/W	Comment
00FF40	D7	_	-				-	-	_		"0" when being read
	D6	FOUT2	FOUT fr	equency	selection				0	R/W	
	D5	FOUT1		FOUT1 0 0 1 1 0 0 0 1 1 1 1 1 1 1 1 1 1 1 1		Frequency  fosc1 / 1  fosc1 / 2  fosc1 / 4  fosc1 / 8  fosc3 / 1  fosc3 / 2  fosc3 / 4			0	R/W	
			1	1	1	fosc3 / 8					
	D3	FOUTON	FOUT or	utput con	trol		On	Off	0	R/W	

Reset

Reset

Run

No operation

No operation

Stop

Table 5.3.3.1 Watchdog timer control bits

# 5.4 Oscillation Circuits and Operating Mode

#### 5.4.1 Configuration of oscillation circuits

The S1C88349 is twin clock system with two internal oscillation circuits (OSC1 and OSC3). OSC1 oscillation circuit generates the 32.768 kHz (Typ.) main clock and OSC3 oscillation circuit the sub-clock when the CPU and some peripheral circuits (output port, serial interface and programmable timer) are in high speed operation. Figure 5.4.1.1 shows the configuration of the oscillation circuit.

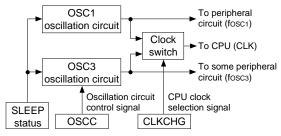
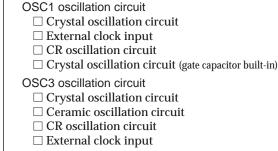


Fig. 5.4.1.1 Configuration of oscillation circuits

At initial reset, OSC1 oscillation circuit is selected for the CPU operating clock and OSC3 oscillation circuit is in a stopped state. ON/OFF switching of the OSC3 oscillation circuit and switching of the system clock between OSC1 and OSC3 are controlled in software. OSC3 circuit is utilized when high speed operation of the CPU and some peripheral circuits become necessary. Otherwise, OSC1 should be used to generate the operating clock and OSC3 circuit placed in a stopped state in order to reduce current consumption.

#### 5.4.2 Mask option



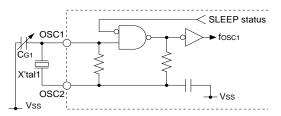
In terms of the oscillation circuit types for OSC1, either crystal oscillation, CR oscillation, crystal oscillation (gate capacitor built-in) or external clock input can be selected with the mask option. In terms of oscillation circuit types for OSC3, either crystal oscillation, ceramic oscillation, CR oscillation or external clock input can be selected with the mask option, in the same way as OSC1.

#### 5.4.3 OSC1 oscillation circuit

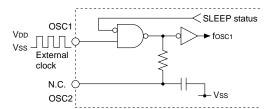
The OSC1 oscillation circuit generates the 32.768 kHz (Typ.) system clock which is utilized during low speed operation (low power mode) of the CPU and peripheral circuits. Furthermore, even when OSC3 is utilized as the system clock, OSC1 continues to generate the source clock for the clock timer and stopwatch timer.

This oscillation circuit stops when the SLP instruction is executed. However, in case the SVD circuit is executing an SLP instruction, oscillation is stopped in synchronization with the completion of sampling. In terms of the oscillation circuit types, either crystal oscillation, CR oscillation, crystal oscillation (gate capacitor built-in) or external clock input can be selected with the mask option.

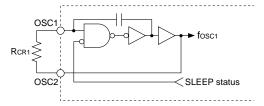
Figure 5.4.3.1 shows the configuration of the OSC1 oscillation circuit.



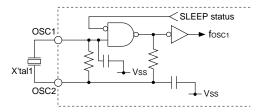
#### (1) Crystal oscillation circuit



#### (2) External clock input



#### (3) CR oscillation circuit



(4) Crystal oscillation circuit (gate capacitor built-in)

Fig. 5.4.3.1 OSC1 oscillation circuit

When crystal oscillation is selected, a crystal oscillation circuit can be easily formed by connecting a crystal oscillator X'tal1 (Typ. 32.768 kHz) between the OSC1 and OSC2 terminals along with a trimmer capacitor Cg1 (5–25 pF) between the OSC1 terminal and Vss.

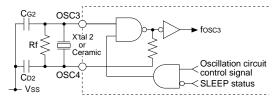
In addition, the gate capacitor CG1 (5 pF) can be built into the circuit by the mask option. When CR oscillation is selected, connect a resistor (RCR1) between the OSC1 and OSC2 terminals. When external input is selected, release the OSC2 terminal and input the rectangular wave clock into the OSC1 terminal.

#### 5.4.4 OSC3 oscillation circuit

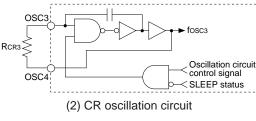
The OSC3 oscillation circuit generates the system clock when the CPU and some peripheral circuits (output port, serial interface and programmable timer) are in high speed operation.

This oscillation circuit stops when the SLP instruction is executed, or the OSCC register is set to "0". In terms of oscillation circuit types, any one of crystal oscillation, ceramic oscillation, CR oscillation or external clock input can be selected with the mask option.

Figure 5.4.4.1 shows the configuration of the OSC3 oscillation circuit.



(1) Crystal/Ceramic oscillation circuit



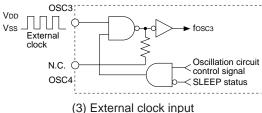


Fig. 5.4.4.1 OSC3 oscillation circuit

When crystal or ceramic oscillation circuit is selected, the crystal or ceramic oscillation circuit are formed by connecting either a crystal oscillator (X'tal2) or a combination of ceramic oscillator (Ceramic) and feedback resistor (Rf) between OSC3 and OSC4 terminals and connecting two capacitors (CG2, CD2) between the OSC3 terminal and Vss, and between the OSC4 terminal and Vss, respectively. When CR oscillation is selected, the CR oscillation circuit is formed merely by connecting a resistor (RCR3) between OSC3 and OSC4 terminals. When external input is selected, release the OSC4 terminal and input the rectangular wave clock into the OSC3 terminal.

#### 5.4.5 Operating mode

power operation.

You can select three types of operating modes using software, to obtain a stable operation and good characteristics (operating frequency and current consumption) over a broad operation voltage. Here below are indicated the features of the respective modes.

# Normal mode (VDD = 2.4 V-5.5 V) This mode is set following the initial reset. It permits the OSC3 oscillation circuit (Max. 4.2 MHz) to be used and also permits relative low

Low power mode (VDD = 1.8 V-3.5 V)
 This is a lower power mode than the normal mode. It makes ultra-low power consumption possible by operation on the OSC1 oscillation circuit, although the OSC3 circuit cannot be used.

# High speed mode (VDD = 3.5 V-5.5 V) This mode permits higher speed operation than the normal mode. Since the OSC3 oscillation circuit (Max. 8.2 MHz) can be used, you should use this mode, when you require operation at 4.2 MHz or more. However, the current consumption will increase relative to the normal mode.

Using software to switch over among the above three modes to meet your actual usage circumstances will make possible a low power system. For example, you will be able to reduce current consumption by switching over to the normal mode when using the OSC3 as the CPU clock and, conversely, changing over to the low power mode when using the OSC1 as the CPU clock (OSC3 oscillation circuit is OFF).

Note: Do not turn the OSC3 oscillation circuit ON in the low power mode.

Do not switch over the operating mode (normal mode  $\leftrightarrow$  high speed mode) in the OSC3 oscillation circuit ON status, as this will cause faulty operation.

You can not use two modes, the low power mode and the high speed mode on one application, with respect to the operating voltages.

#### 5.4.6 Switching the CPU clocks

You can use either OSC1 or OSC3 as the system clock for the CPU and you can switch over by means of software.

You can save power by turning the OSC3 oscillation circuit off while the CPU is operating in OSC1. When you must operate on OSC3, you can change to high speed operation by turning the OSC3 oscillation circuit ON and switching over the system clock.

In this case, since several msec to several tens of msec are necessary for the oscillation to stabilize after turning the OSC3 oscillation circuit ON, you should switch over the clock after stabilization time has elapsed. (The oscillation start time will vary somewhat depending on the oscillator and on the externally attached parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)

When switching over from the OSC3 to the OSC1, turn the OSC3 oscillation circuit OFF immediately following the clock changeover. The basic clock switching procedure is as described above, however, you must also combine it with the changeover of the operating mode to permit low current consumption and high speed operation. Figure 5.4.6.1 indicates the status transition diagram for the operation mode and clock changeover.

Note: When turning ON the OSC3 oscillation circuit after switching the operating mode, you should allow a minimum waiting time of 5 msec.

#### **Program Execution Status**

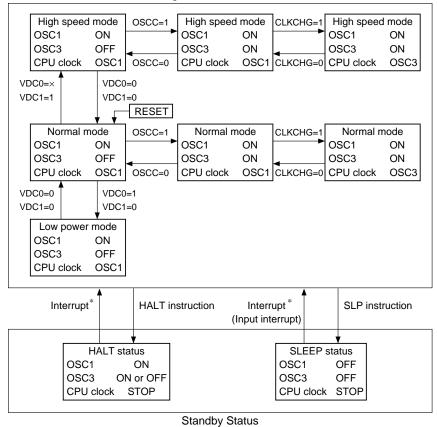


Fig. 5.4.6.1 Status transition diagram for the operation mode and clock changeover

<sup>\*</sup> The return destination from the standby status becomes the program execution status prior to shifting to the standby status

#### 5.4.7 Control of oscillation circuit and operating mode

Table 5.4.7.1 shows the control bits for the oscillation circuits and operating modes.

Table 5.4.7.1 Oscillation circuit and operating mode control bits

Address	Bit	Name		Function				1	0	SR	R/W	Comment
00FF02	D7	EBR	Bus releas	se enable r	egister	ļ	K11	BREQ	Input port	0	R/W	
	<i>D1</i>	LDK	(K11 and	R51 termi	nal specific	cation)	R51	BACK	Output port	U	IX/ VV	
			Wait cont	rol registe	r	Nun	ıber			0	R/W	
	D6	WT2	WT2	WT1	WT0	of s	tate					
			1	1	1	1	4					
			1	1	0	13				0	R/W	
	D5	WT1	I 1	0	0	10						
			0	1	1	8						
		D4 WT0	0	1	0	6 4	) L			0	R/W	
	D4		0	0	1	2	2					
			0	0	0	No v	wait					
	D3	CLKCHG	CPU oper	ating cloc	k switch			OSC3	OSC1	0	R/W	
	D2	oscc	OSC3 osc	illation O	n/Off contro	ol		On	Off	0	R/W	
			Operating	mode sel	ection					0	R/W	
	D1	VDC1	VDC1	VDC0	Operat	ing mod	le					
			1		High speed	d (VD1=	3.3V)					
			0	1	Low powe					0	R/W	
	D0	VDC0	0	0	Normal	(VD1=	2.2V)					

#### VDC1, VDC0: 00FF02H•D1, D0

Selects the operating mode according to supply voltage and operating frequency.

Table 5.4.7.2 shows the correspondence between register preset values and operating modes.

Table 5.4.7.2 Correspondence between register preset values and operating modes

Operating mode	VDC1	VDC0	V <sub>D</sub> 1	Power voltage	Operating frequency
Normal mode	0	0	2.2 V	2.4-5.5 V	4.2 MHz (Max.)
Low power mode	0	1	1.2 V	1.8-3.5 V	50 kHz (Max.)
High speed mode	1	×	3.3 V	3.5-5.5 V	8.2 MHz (Max.)

<sup>\*</sup> The VDI voltage is the value where VSS has been made the standard (GND).

At initial reset, this register is set to "0" (normal mode).

#### OSCC: 00FF02H•D2

Controls the ON and OFF settings of the OSC3 oscillation circuit.

When "1" is written: OSC3 oscillation ON When "0" is written: OSC3 oscillation OFF

Reading: Valid

When the CPU and some peripheral circuits (output port, serial interface and programmable timer) are to be operated at high speed, OSCC is to be set to "1". At all other times, it should be set to "0" in order to reduce current consumption.

At initial reset, OSCC is set to "0" (OSC3 oscillation OFF).

#### CLKCHG: 00FF02H•D3

Selects the operating clock for the CPU.

When "1" is written: OSC3 clock When "0" is written: OSC1 clock Reading: Valid

When the operating clock for the CPU is switched to OSC3, CLKCHG should be set to "1" and when the clock is switched to OSC1, CLKCHG should be set to "0".

At initial reset, CLKCHG is set to "0" (OSC1 clock).

#### 5.4.8 Programming notes

- When the high speed CPU operation is not necessary, you should operate the peripheral circuits according to the setting outline indicate below.
  - CPU operating clock OSC1
  - OSC3 oscillation circuit
     OFF (When the OSC3 clock is not necessary for some peripheral circuits.)
  - Operating mode
     Low power mode (When VDD-VSS is 3.5 V or less)
     or Normal mode (When VDD-VSS is 3.5 V or more)
- (2) Do not turn the OSC3 oscillation circuit ON in the low power mode. Do not switch over the operating mode (normal mode ↔ high speed mode) in the OSC3 oscillation circuit ON status, as this will cause faulty operation.
- (3) When turning ON the OSC3 oscillation circuit after switching the operating mode, you should allow a minimum waiting time of 5 msec.
- (4) Since several msec to several tens of msec are necessary for the oscillation to stabilize after turning the OSC3 oscillation circuit ON.

  Consequently, you should switch the CPU operating clock (OSC1 → OSC3) after allowing for a sufficient waiting time once the OSC3 oscillation goes ON. (The oscillation start time will vary somewhat depending on the oscillator and on the externally attached parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)
- (5) When switching the clock from OSC3 to OSC1, be sure to switch OSC3 oscillation OFF with separate instructions. Using a single instruction to process simultaneously can cause a malfunction of the CPU.

#### 5.5 Input Ports (K ports)

#### 5.5.1 Configuration of input ports

The S1C88349 is equipped with 10 input port bits (K00–K07, K10 and K11) all of which are usable as general purpose input port terminals with interrupt function.

K10 terminal doubles as the external clock (EVIN) input terminal of the programmable timer (event counter) with input port functions sharing the input signal as is. (See "5.11 Programmable Timer")

Furthermore, it should be noted, however, that K11 terminal is shared with the bus authority release request signal ( $\overline{BREQ}$ ) input terminal. Function assignment of this terminal can be selected in software. When this terminal is selected for  $\overline{BREQ}$  signal, K11 cannot be used as an input port. (See "5.2 System Controller and Bus Control") In the explanation below, it is assumed that K11 is set as an input port.

Each input port is equipped with a pull-up resistor. The mask option can be used to select either "With resistor" or "Gate direct" for each input port. Figure 5.5.1.1 shows the structure of the input port.

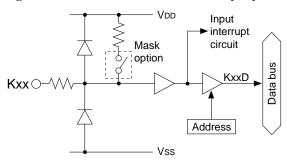


Fig. 5.5.1.1 Structure of input port

Each input port terminal is directly connected via a three-state buffer to the data bus. Furthermore, the input signal state at the instant of input port readout is read in that form as data.

#### 5.5.2 Mask option

Input port pull-up resistors						
K00   With resistor	$\square$ Gate direct					
K01 □ With resistor	$\square$ Gate direct					
K02 □ With resistor	$\square$ Gate direct					
K03 □ With resistor	☐ Gate direct					
K04 □ With resistor	☐ Gate direct					
K05 □ With resistor	$\square$ Gate direct					
K06 □ With resistor	$\square$ Gate direct					
K07   With resistor	$\square$ Gate direct					
K10 □ With resistor	☐ Gate direct					
K11 $\square$ With resistor	$\square$ Gate direct					

Input ports K00–K07, K10 and K11 are all equipped with pull-up resistors. The mask option can be used to select 'With resistor' or 'Gate direct' for each port (bit).

The 'With resistor' option is rendered suitable for purposes such as push switch or key matrix input. When changing the input terminal from LOW level to HIGH with the built-in pull-up resistor, a delay in the waveform rise time will occur depending on the time constant of the pull-up resistor and the load capacitance of the terminal. It is necessary to set an appropriate wait time for introduction of an input port. In particular, special attention should be paid to key scan for key matrix formation. Make this wait time the amount of time or more calculated by the following expression.

### Wait time = RIN x (CIN + load capacitance on the board) x 1.6 [sec]

RIN: Pull up resistance Max. value CIN: Terminal capacitance Max. value

When 'Gate direct' is selected, the pull-up resistor is detached and the port is rendered suitable for purposes such as slide switch input and interfacing with other LSIs.

In this case, take care that a floating state does not occur in input.

For unused input ports, select the default setting of "With resistor".

## 5.5.3 Interrupt function and input comparison register

Input port K00–K07, K10 and K11 are all equipped with an interrupt function. These input ports are divided into three groupings: K00–K03 (K0L), K04–K07 (K0H) and K10–K11 (K1). Furthermore, the interrupt generation condition for each series of terminals can be set by software.

When the interrupt generation condition set for each series of terminals is met, the interrupt factor flag FK0L, FK0H or FK1 corresponding to the applicable series is set at "1" and an interrupt is generated.

K00 ()

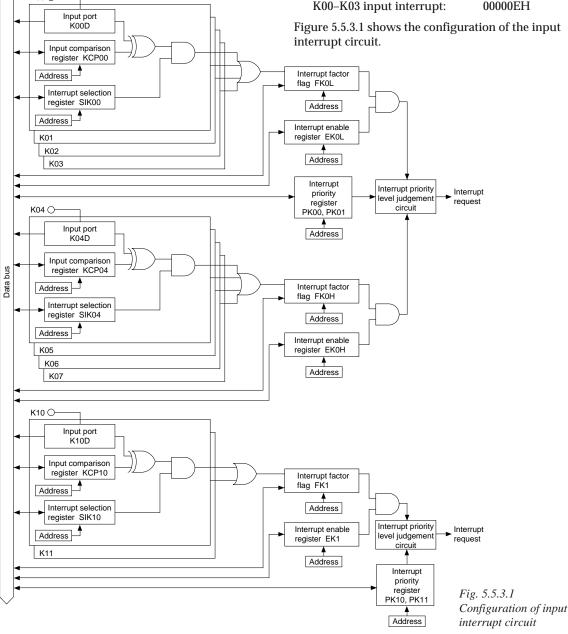
Interrupt can be prohibited by setting the interrupt enable registers EK0L, EK0H and EK1 for the corresponding interrupt factor flags.

Furthermore, the priority level for input interrupt can be set at the desired level (0–3) using the interrupt priority registers PK00–PK01 and PK10–PK11 corresponding to each of two groups K0x (K00–K07) and K1x (K10–K11).

For details on the interrupt control registers for the above and on operations subsequent to interrupt generation, see "5.17 Interrupt and Standby Status".

The exception processing vectors for each interrupt factor are set as follows:

K10 and K11 input interrupt: 00000AH K04-K07 input interrupt: 00000CH K00-K03 input interrupt: 00000EH



The interrupt selection registers SIK00–SIK03, SIK04–SIK07 and SIK10–SIK11 and input comparison registers KCP00–KCP03, KCP04–KCP07 and KCP10–KCP11 for each port are used to set the interrupt generation condition described above.

Input port interrupt can be permitted or prohibited by the setting of the interrupt selection register SIK. In contrast to the interrupt enable register EK which masks the interrupt factor for each series of terminals, the interrupt selection register SIK is masks the bit units.

The input comparison register KCP selects whether the interrupt for each input port will be generated on the rising edge or the falling edge of input.

When the data content of the input terminals in which interrupt has been permitted by the interrupt selection register SIK and the data content of the input comparison register KCP change from a conformity state to a non-conformity state, the interrupt factor flag FK should be set to "1" and an interrupt is generated.

Figure 5.5.3.2 shows an example of interrupt generation in the series of terminals K0L (K00–K03).

Because interrupt has been prohibited for K00 by the interrupt selection register SIK00, with the settings as shown in (2), an interrupt will not be generated.

Since K03 is "0" in the next settings (3) in the figure, the non-conformity between the input terminal data K01–K03 where interrupt is permitted and the data from the input comparison registers KCP01–KCP03 generates an interrupt.

In line with the explanation above, since the change in the contents of input data and input comparison registers KCP from a conformity state to a non-conformity state introduces an interrupt generation condition, switching from one non-conformity state to another, as is the case in (4) in the figure, will not generate an interrupt. Consequently, in order to be able to generate a second interrupt, either the input terminal must be returned to a state where its content is once again in conformity with that of the input comparison register KCP, or the input comparison register KCP must be reset. Input terminals for which interrupt is prohibited will not influence an interrupt generation condition.

Interrupt is generated in exactly the same way in the other two series of terminals K0H (K04-K07) and K1 (K10 and K11).

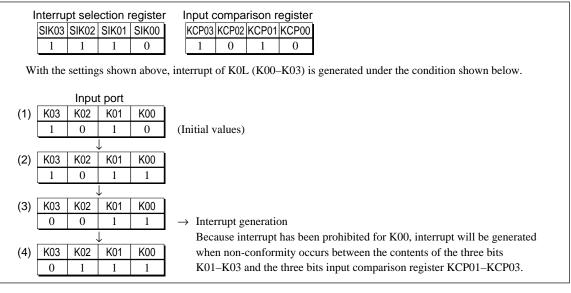


Fig. 5.5.3.2 Interrupt generation example in K0L (K00–K03)

#### 5.5.4 Control of input ports

Table 5.5.4.1 shows the input port control bits.

Table 5.5.4.1(a) Input port control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF50	D7	SIK07	K07 interrupt selection register					
	D6	SIK06	K06 interrupt selection register					
	D5	SIK05	K05 interrupt selection register					
		SIK04	K04 interrupt selection register	Interrupt	Interrupt			
			K03 interrupt selection register	enable	disable	0	R/W	
	D2	SIK02	K02 interrupt selection register					
	D1	SIK01	K01 interrupt selection register					
	D0	SIK00	K00 interrupt selection register					
00FF51	D7	_	-	_	_	_		
	D6	_	_	_	_	_		-
	D5	_	_	_	_	_		Constantly "0" when
	D4	_	_	_	_	_		being read
	D3	_	_	_	_	_		
	D2	_	_	_	_	_		
	D1	SIK11	K11 interrupt selection register	Interrupt	Interrupt			
	D0	SIK10	K10 interrupt selection register	enable	disable	0	R/W	
00FF52		KCP07	K07 interrupt comparison register	chaole	distroit			
001102		KCP06	K06 interrupt comparison register					
		KCP05	K05 interrupt comparison register	Interrupt	Interrupt			
		KCP04	K04 interrupt comparison register	generated	generated			
		KCP03	K03 interrupt comparison register	at falling	at rising	1	R/W	
		KCP02	K02 interrupt comparison register	edge	edge			
		KCP01	K01 interrupt comparison register	cuge	cugc			
	D0	KCP00	K00 interrupt comparison register					
00FF53	D7	_		_	_	_		
001100	D6	_	_		_	_		
	D5	_	_		_	_		Constantly "0" when
	D4	_	_	_	_	<u> </u>		being read
	D3	_	_		_	_		being read
	D2	_	_			_		-
		KCP11	K11 interrupt comparison register	Falling	Rising			
	D0	KCP10	K10 interrupt comparison register	edge	edge	1	R/W	
00FF54		K07D	K07 input port data	cuge	cuge			
001104		K06D	K06 input port data					
		K05D	K05 input port data					
		K04D	K04 input port data	High level	Low level			
		K03D	K03 input port data	input	input	-	R	
		K02D	K02 input port data	mput	mput			
		K01D	K01 input port data					
	D0	K00D	K00 input port data					
00FF55	D7	_		_	_			
501155	D6			_	_	+-		-
	D5	_	_	_	_			Constantly "0" when
	D5 D4	_	_	_		<del>-</del>		being read
	D3	_	_	_	_	<del>  -</del>		ocing read
	D3		_	_	_	<del>  -</del>		-
		– K11D	K11 input port data	Think 11	- I om 11	<del>-</del>		
		K11D K10D		High level	Low level	-	R	
	טט	KIUD	K10 input port data	input	input			

Table 5.5.4.1(b) Input port control bits

Address	Bit	Name	Function	1		0	SR	R/W	Comment
00FF20	D7	PK01	K00–K07 interrupt priority register				0	R/W	
	D6	PK00	K00–K07 interrupt priority register	PK01 PK00		0	U	K/W	
	D5	PSIF1	Serial interface interrupt priority register	al interface interrupt priority register PSW1 PSW0 Priority 0	R/W				
	D4	PSIF0	Serial interface interrupt priority register	PTM1	PTM	0 level		) IX W	
	D3	PSW1	Stopwatch timer interrupt priority register	1 1	1	Level 3 Level 2	0	R/W	
	D2	PSW0	Stopwatch timer interrupt priority register	0	1	Level 1		IX VV	
		PTM1	Clock timer interrupt priority register	0	0	Level 0	0	R/W	
	D0	PTM0	Clock timer interrupt priority register				0	IV VV	
00FF21	D7	_	_	-		-	_		
	D6	_	_	-		-	_		Constantly "0" when
	D5	_	_	-		-	_		being read
	D4	_	_	-		-	_		
		PPT1	Programmable timer interrupt priority register	PPT1 PK11	PPT PK1		0	R/W	
	_	PPT0	Trogrammatic timer metrupt priority register	1					
		PK11			Level 2 Level 1	0	R/W		
	<u> </u>	PK10	11 , 0	ő	0	Level 0	0	10 11	
00FF23		EPT1	Programmable timer 1 interrupt enable register						
	_	EPT0	Programmable timer 0 interrupt enable register						
		EK1	K10 and K11 interrupt enable register						
		EK0H	K04–K07 interrupt enable register	Interr	upt	Interrupt	0	R/W	
		EK0L	K00–K03 interrupt enable register	enab	ole	disable		10 11	
		ESERR	Serial I/F (error) interrupt enable register						
		ESREC	Serial I/F (receiving) interrupt enable register						
	_	ESTRA	Serial I/F (transmitting) interrupt enable register						
00FF25	D7	FPT1	Programmable timer 1 interrupt factor flag	(R)	)	(R)			
	_	FPT0	Programmable timer 0 interrupt factor flag	Interr	upt	No interrupt			
	D5	FK1	K10 and K11 interrupt factor flag	facto	r is	factor is			
		FK0H	K04–K07 interrupt factor flag	genera	ated	generated	0	R/W	
	_	FK0L	K00–K03 interrupt factor flag					" "	
		FSERR	Serial I/F (error) interrupt factor flag	(W	)	(W)			
		FSREC	Serial I/F (receiving) interrupt factor flag	Res	et	No operation			
	D0	FSTRA	Serial I/F (transmitting) interrupt factor flag						

K00D-K07D: 00FF54H K10D, K11D: 00FF55H•D0, D1

Input data of input port terminal Kxx can be read out.

When "1" is read: HIGH level
When "0" is read: LOW level
Writing: Invalid

The terminal voltage of each of the input port K00–K07, K10 and K11 can be directly read out as either a "1" for HIGH (VDD) level or a "0" for LOW (Vss) level.

This bit is exclusively for readout and are not usable for write operations.

SIK00-SIK07: 00FF50H SIK10, SIK11: 00FF51H•D0, D1

Sets the interrupt generation condition (interrupt permission/prohibition) for input port terminals K00–K07, K10 and K11.

When "1" is written: Interrupt permitted When "0" is written: Interrupt prohibited

Reading: Valid

SIKxx is the interrupt selection register which correspond to the input port Kxx. A "1" setting permits interrupt in that input port and a "0" prohibits it. Changes of state in an input terminal in which interrupt is prohibited, will not influence interrupt generation.

At initial reset, this register is set to "0" (interrupt prohibited).  $\,$ 

#### KCP00-KCP07: 00FF52H KCP10, KCP11: 00FF53H•D0, D1

Sets the interrupt generation condition (interrupt generation timing) for input port terminals K00–K07, K10 and K11.

When "1" is written: Falling edge When "0" is written: Rising edge Reading: Valid

KCPxx is the input comparison register which correspond to the input port Kxx. Interrupt in those ports which have been set to "1" is generated on the falling edge of the input and in those set to "0" on the rising edge.

At initial reset, this register is set to "1" (falling edge).

#### PK00, PK01: 00FF20H•D6, D7 PK10, PK11: 00FF21H•D0, D1

Sets the input interrupt priority level. The two bits PK00 and PK01 are the interrupt priority registers corresponding to the interrupts for K00–K07 (K0L and K0H). Corresponding to K10–K11 (K1), the two bits PK10 and PK11 perform the same function. Table 5.5.4.2 shows the interrupt priority level which can be set by this register.

Table 5.5.4.2 Interrupt priority level settings

PK11	PK10	
PK01	PK00	Interrupt priority level
1	1	Level 3 (IRQ3)
1	0	Level 2 (IRQ2)
0	1	Level 1 (IRQ1)
0	0	Level 0 (None)

At initial reset, this register is set to "0" (level 0).

#### EKOL, EKOH, EK1: 00FF23H•D3, D4, D5

How interrupt generation to the CPU is permitted or prohibited.

When "1" is written: Interrupt permitted When "0" is written: Interrupt prohibited

Reading: Valid

The interrupt enable register EK0L corresponds to K00–K03, EK0H to K04–K07, and EK1 to K10–K11. Interrupt is permitted in those series of terminals set to "1" and prohibited in those set to "0". At initial reset, this register is set to "0" (interrupt prohibited).

#### FK0L, FK0H, FK1: 00FF25H•D3, D4, D5

Indicates the generation state for an input interrupt.

When "1" is read: Interrupt factor present When "0" is read: Interrupt factor not present

When "1" is written: Reset factor flag

When "0" is written: Invalid

The interrupt factor flag FK0L corresponds to K00–K03, FK0H to K04–K07, and FK1 to K10–K11 and they are set to "1" by the occurrence of an interrupt generation condition.

When set in this manner, if the corresponding interrupt enable register is set to "1" and the corresponding interrupt priority register is set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. Regardless of the interrupt enable register and interrupt priority register settings, the interrupt factor flag will be set to "1" by the occurrence of an interrupt generation condition.

To accept the subsequent interrupt after interrupt generation, re-setting of the interrupt flags (set interrupt flag to lower level than the level indicated by the interrupt priority registers, or execute the RETE instruction) and interrupt factor flag reset are necessary. The interrupt factor flag is reset to "0" by writing "1".

At initial reset, this flag is all reset to "0".

#### 5.5.5 Programming note

When changing the input terminal from LOW level to HIGH with the built-in pull-up resistor, a delay in the waveform rise time will occur depending on the time constant of the pull-up resistor and the load capacitance of the terminal. It is necessary to set an appropriate wait time for introduction of an input port. In particular, special attention should be paid to key scan for key matrix formation. Make this wait time the amount of time or more calculated by the following expression.

Wait time = RIN x (CIN + load capacitance on the board) x 1.6 [sec]

RIN: Pull up resistance Max. value CIN: Terminal capacitance Max. value

#### 5.6 Output Ports (R ports)

#### 5.6.1 Configuration of output ports

The S1C88349 is equipped with 34 bits of output ports (R00–R07, R10–R17, R20–R27, R30–R37, R50, R51).

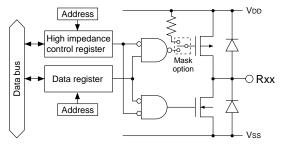
Depending on the bus mode setting, the configuration of the output ports may vary as shown in the table below.

Table 5.6.1.1 Configuration of output ports

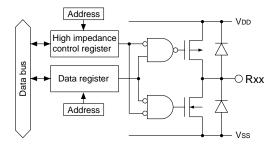
Tamainal	Bus mode					
Terminal	Single chip	Expanded 64K	Expanded 512K			
R00	Output port R00	Addres	s A0			
R01	Output port R01	Addres	s A1			
R02	Output port R02	Address A2				
R03	Output port R03	Addres	s A3			
R04	Output port R04	Addres	s A4			
R05	Output port R05	Addres	s A5			
R06	Output port R06	Addres	s A6			
R07	Output port R07	Addres	s A7			
R10	Output port R10	Addres	s A8			
R11	Output port R11	Addres	s A9			
R12	Output port R12	Address	A10			
R13	Output port R13	Address	A11			
R14	Output port R14	Address A12				
R15	Output port R15	Address A13				
R16	Output port R16	Address A14				
R17	Output port R17	Address A15				
R20	Output p	port R20	Address A16			
R21	Output p	port R21	Address A17			
R22	Output p	oort R22	Address A18			
R23	Output port R23	RD sig	gnal			
R24	Output port R24	WR si	gnal			
R25		Output port R25				
R26		Output port R26				
R27		Output port R27				
R30	Output port R30	Output port R3	0/ <del>CE0</del> signal			
R31	Output port R31	Output port R3	1/CE1 signal			
R32	Output port R32	Output port R3	2/CE2 signal			
R33	Output port R33	Output port R3	3/CE3 signal			
R34		Output port R34				
R35		Output port R35				
R36	Output port R36					
R37		Output port R37				
R50	Output port R50					
R51	Output port R51	Output port R51	BACK signal			

Only the configuration of the output ports in single chip mode will be discussed here. With respect to bus control, see "5.2 System Controller and Bus Control".

Figure 5.6.1.1 shows the basic structure (excluding special output circuits) of the output ports.



Nch open drain can be set for R00–R07 and R10–R17 by the mask option.



R20-R27, R30-R37, R50, R51

Fig. 5.6.1.1 Structure of output ports

In modes other than single chip mode, the data registers and high impedance control registers of the output ports used for bus function can be used as general purpose registers with read/write capabilities. This will not in any way affect bus signal output.

The output specification of each output port is as complementary output with high impedance control in software possible.

Besides normal DC output, output ports R25–R27, R34, and R50 have a special output function, which can be selected by software.

#### 5.6.2 Mask option

Output ports R00-R07 and	R10-R17 output
specifications	
R00 □ Complementary	□ Nch open drain
R01 □ Complementary	☐ Nch open drain
R02 □ Complementary	☐ Nch open drain
R03 □ Complementary	☐ Nch open drain
R04 □ Complementary	☐ Nch open drain
R05 □ Complementary	☐ Nch open drain
R06 □ Complementary	☐ Nch open drain
R07   Complementary	☐ Nch open drain
R10   Complementary	☐ Nch open drain
R11 □ Complementary	☐ Nch open drain
R12   Complementary	☐ Nch open drain
R13 □ Complementary	☐ Nch open drain
R14   Complementary	☐ Nch open drain
R15   Complementary	☐ Nch open drain
R16   Complementary	☐ Nch open drain
R17 □ Complementary	☐ Nch open drain

Output ports R00–R07 and R10–R17 can be used to select output specification for each port (1 bit) by mask option.

The output specification can be selected for either complementary output or Nch open drain output.

Nch open drain output is rendered suitable for purposes as key matrix common output.

For unused input ports, select the default setting of "Complementary".

Note: When Nch open drain has been selected, voltage in excess of the supply voltage range must not applied to the output port terminal.

#### 5.6.3 High impedance control

The output port can be high impedance controlled in software.

This makes it possible to share output signal lines with an other external device.

A high impedance control register is set for each series of output port terminals as shown below. Either complementary output and high impedance state can be selected with this register.

Table 5.6.3.1 Correspondence between output ports and high impedance control registers

Register	Output port terminal
HZR0L	R00-R03
HZR0H	R04-R07
HZR1L	R10-R13
HZR1H	R14–R17
HZR20	R20
HZR21	R21
HZR22	R22
HZR23	R23
HZR24	R24
HZR25	R25
HZR26	R26
HZR27	R27
HZR30	R30
HZR31	R31
HZR32	R32
HZR33	R33
HZR34	R34
HZR35	R35
HZR36	R36
HZR37	R37
HZR4L*	_
HZR4H*	_
HZR50	R50
HZR51	R51

<sup>\*</sup> This is a 2-bit reserved register, it can be used as a general purpose register with read/write capabilities.

When a high impedance control register HZRxx is set to "1", the corresponding output port terminal becomes high impedance state and when set to "0", it becomes complementary output.

#### 5.6.4 DC output

As Figure 5.6.1.1 shows, when "1" is written to the output port data register, the output terminal switches to HIGH (VDD) level and when "0" is written it switches to LOW (Vss) level. When output is in a high impedance state, the data written to the data register is output from the terminal at the instant when output is switched to complementary.

#### 5.6.5 Special output

Besides normal DC output, output ports R25–R27, R34 and R50 can also be assigned special output functions in software as shown in Table 5.6.5.1.

Table 5.6.5.1 Special output ports

Output port	Special output	
R25	CL output	
R26	FR output	
R27	TOUT output	
R34	FOUT output	
R50	BZ output	

#### ■ CL and FR output (R25 and R26)

In order for the S1C88349 to handle connection to an externally expanded LCD driver, output ports R25 and R26 can be used to output a CL signal (LCD synchronous signal) and FR signal (LCD frame signal), respectively.

The configuration of output ports R25 and R26 are shown in Figure 5.6.5.1.

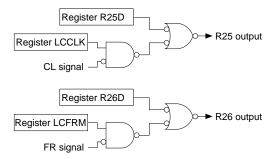


Fig. 5.6.5.1 Configuration of R25 and R26

The output control for the CL signal is done by the register LCCLK. When you set "1" for the LCCLK, the CL signal is output from the output port terminal R25, when "0" is set, the HIGH (VDD) level is output. At this time, "1" must always be set for the data register R25D.

The output control for the FR signal is done by the register LCFRM. When you set "1" for the LCFRM, the FR signal is output from the output port terminal R26, when "0" is set, the HIGH (VDD) level is output. At this time, "1" must always be set for the data register R26D.

The frequencies of each signal are changed as shown in Table 5.6.5.2 according to the drive duty selection.

Table 5.6.5.2 Frequencies of CL and FR signals

Drive duty	CL signal (Hz)	FR signal (Hz)
1/32	2,048	32
1/16	1,024	32
1/8	1,024	64

Since the signals are generated asynchronously from the registers LCCLK and LCFRM, when the signals are turned ON or OFF by the register settings, a hazard of a 1/2 cycle or less is generated. Figure 5.6.5.2 shows the output waveforms of the CL and FR signals.



Fig. 5.6.5.2 Output waveforms of CL and FR signals

#### ■ TOUT output (R27)

In order for the S1C88349 to provide clock signal to an external device, the output port terminal R27 can be used to output a TOUT signal (clock output by the programmable timer). The configuration of output port R27 is shown in Figure 5.6.5.3.

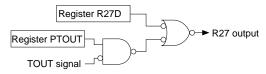


Fig. 5.6.5.3 Configuration of R27

The output control for the TOUT signal is done by the register PTOUT. When you set "1" for the PTOUT, the TOUT signal is output from the output port terminal R27, when "0" is set, the HIGH (VDD) level is output. At this time, "1" must always be set for the data register R27D.

The TOUT signal is the programmable timer underflow divided by 1/2.

With respect to frequency control, see "5.11 Programmable Timer".

Since the TOUT signal is generated asynchronously from the register PTOUT, when the signal is turned ON or OFF by the register settings, a hazard of a 1/2 cycle or less is generated.

Figure 5.6.5.4 shows the output waveform of the TOUT signal.



Fig. 5.6.5.4 Output waveform of TOUT signal

#### **■** FOUT output (R34)

In order for the S1C88349 to provide clock signal to an external device, a FOUT signal (oscillation clock fosc1 or fosc3 dividing clock) can be output from the output port terminal R34.

Figure 5.6.5.5 shows the configuration of output port R34.

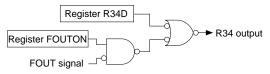


Fig. 5.6.5.5 Configuration of R34

The output control for the FOUT signal is done by the register FOUTON. When you set "1" for the FOUTON, the FOUT signal is output from the output port terminal R34, when "0" is set, the HIGH (VDD) level is output. At this time, "1" must always be set for the data register R34D.

The frequency of the FOUT signal can be selected in software by setting the registers FOUT0–FOUT2. The frequency is selected any one from among eight settings as shown in Table 5.6.5.3.

Table 5.6.5.3 FOUT frequency setting

FOUT2	FOUT1	FOUT0	FOUT frequency
0	0	0	foscı / 1
0	0	1	foscı / 2
0	1	0	fosc1 / 4
0	1	1	fosc1 / 8
1	0	0	fosc3 / 1
1	0	1	fosc3 / 2
1	1	0	fosc3 / 4
1	1	1	fosc3 / 8

fosc1: OSC1 oscillation frequency fosc3: OSC3 oscillation frequency

When the FOUT frequency is made "fosc3/n", you must turn on the OSC3 oscillation circuit before outputting FOUT. A time interval of several msec to several 10 msec, from the turning ON of the OSC3 oscillation circuit to until the oscillation stabilizes, is necessary, due to the oscillation element that is used. Consequently, if an abnormality occurs as the result of an unstable FOUT signal being output externally, you should allow an adequate waiting time after turning ON of the OSC3 oscillation, before turning outputting FOUT. (The oscillation start time will vary somewhat depending on the oscillator and on the externally attached parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)

At initial reset, OSC3 oscillation circuit is set to OFF state.

Since the FOUT signal is generated asynchronously from the register FOUTON, when the signal is turned ON or OFF by the register settings, a hazard of a 1/2 cycle or less is generated.

Figure 5.6.5.6 shows the output waveform of the FOUT signal.



Fig. 5.6.5.6 Output waveform of FOUT signal

#### ■ BZ output (R50)

In order for the S1C88349 to drive an external buzzer, a BZ signal (sound generator output) can be output from the output port terminal R50. The configuration of the output port R50 is shown in Figure 5.6.5.7.

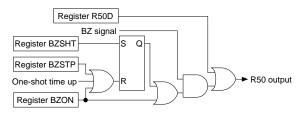


Fig. 5.6.5.7 Configuration of R50

The output control for the BZ signal is done by the registers BZON, BZSHT and BZSTP. When you set "1" for the BZON or BZSHT, the BZ signal is output from the output port terminal R50, when "0" is set for the BZON or "1" is set for the BZSTP, the LOW (Vss) level is output. At this time, "0" must always be set for the data register R50D.

The BZ signal which is output makes use of the output of the sound generator. With respect to control of frequency and envelope, see "5.13 Sound Generator".

Since the BZ signal is generated asynchronously from the registers BZON, BZSHT and BZSTP, when the signal is turned ON or OFF by the register settings, a hazard of a 1/2 cycle or less is generated. Figure 5.6.5.8 shows the output waveform of the BZ signal.



Fig. 5.6.5.8 Output waveform of BZ signal

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#### 5.6.6 Control of output ports

Table 5.6.6.1 shows the output port control bits.

Table 5.6.6.1(a) Output port control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF70	D7	HZR51	R51 high impedance control	High	Comple-	_		
,	D6	HZR50	R50 high impedance control	impedance	mentary	0	R/W	
,	D5	HZR4H	R/W register		-			
,	D4	HZR4L R/W register		1	0	0	R/W	Reserved register
,	D3	HZR1H	R14–R17 high impedance control					
,		HZR1L	R10–R13 high impedance control	High	Comple-			
,	D1	HZR0H	R04–R07 high impedance control	impedance	mentary	0	R/W	
,	D0	HZR0L	R00–R03 high impedance control		-			
00FF71	D7	HZR27	R27 high impedance control					
,	D6	HZR26	R26 high impedance control					
,	D5	HZR25	R25 high impedance control					
,	D4	HZR24	R24 high impedance control	High	Comple-			
,	D3	HZR23	R23 high impedance control	impedance	mentary	0	R/W	
,	D2	HZR22	R22 high impedance control		-			
,	D1	HZR21	R21 high impedance control					
,	D0	HZR20	R20 high impedance control					
00FF72	D7	HZR37	R37 high impedance control					
,	D6	HZR36	R36 high impedance control					
,	D5	HZR35	R35 high impedance control					
,		HZR34	R34 high impedance control	High	Comple-			
,	D3	HZR33	R33 high impedance control	impedance	mentary	0	R/W	
,		HZR32	R32 high impedance control	1				
,		HZR31	R31 high impedance control	•				
,	D0	HZR30	R30 high impedance control					
00FF73	D7	R07D	R07 output port data					
,	D6	R06D	R06 output port data					
,	D5	R05D	R05 output port data					
,	D4	R04D	R04 output port data		_	١.		
,	D3	R03D	R03 output port data	High	Low	1	R/W	
,	D2	R02D	R02 output port data					
,	D1	R01D	R01 output port data					
,	D0	R00D	R00 output port data					
00FF74	D7	R17D	R17 output port data					
,	D6	R16D	R16 output port data					
,	D5	R15D	R15 output port data					
,	D4	R14D	R14 output port data				D ///	
ı	D3	R13D	R13 output port data	High	Low	1	R/W	
,	D2	R12D	R12 output port data					
,	D1	R11D	R11 output port data					
,	D0	R10D	R10 output port data					
00FF75	D7	R27D	R27 output port data					
,	D6	R26D	R26 output port data					
,		R25D	R25 output port data					
,		R24D	R24 output port data	***			D	
,			R23 output port data	High	Low	1	R/W	
,		R22D	R22 output port data					
,		R21D	R21 output port data					
,		R20D	R20 output port data					

Table 5.6.6.1(b) Output port control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF76	D7	R37D	R37 output port data					
	D6	R36D	R36 output port data	1				
		R35D	R35 output port data			ı		
	D4	R34D	R34 output port data					
	D3	R33D	R33 output port data	High	Low	1	R/W	
	D2 R32D R32 output port data							
	D1	R31D	R31 output port data	1				
	D0	R30D	R30 output port data					
00FF77	D7	R47D	R/W register					
	D6	R46D	R/W register					
	D5	R45D	R/W register					
	D4	R44D	R/W register				D /337	
	D3	R43D	R/W register	1	0	1	K/W	Reserved register
	D2	R42D	R/W register					
	D1	R41D	R/W register	•				
	D0	R40D	R/W register					
00FF78	D7	_	_	-	-	_		
	D6	-	-	-	-	_		
	D5	-	_	-	-	_		Constantly "0" when
	D4	-	_	-	-	_		being read
	D3	1	_	_	_	-		
	D2	-	-	_	_	-		
	D1	R51D	R51 output port data	High	Low	1	R/W	
	D0	R50D	R50 output port data	High	Low	0	R/W	
00FF10	D7	-		-	-	_		Constantly "0" when
	D6	_		-	-	_		being read
	D5	_	_	-	-	_		being read
	D4	LCCLK	CL output control for expanded LCD driver	On	Off	0	R/W	
	D3	LCFRM	FR output control for expanded LCD driver	On	Off	0	R/W	
	D2	DTFNT	LCD dot font selection	5 x 5 dots	5 x 8 dots	0	R/W	
	D1	LDUTY	LCD drive duty selection	1/16 duty	1/32 duty	0	R/W	*1
	D0	SGOUT	R/W register	1	0	0	R/W	Reserved register
00FF30	D7	_		-	-	_		Constantly "0" when
	D6	_		-	-	_		being read
	D5	_	_	-	-	_		
			8/16-bit mode selection	16-bit x 1	8-bit x 2	0	R/W	
	-	CHSEL	TOUT output channel selection	Timer 1	Timer 0	0	R/W	
		PTOUT	TOUT output control	On	Off	0	R/W	
			Prescaler 1 source clock selection	fosc3	foscı	0	R/W	
		CKSEL0	Prescaler 0 source clock selection	fosc3	foscı	0	R/W	
00FF44	D7	_	_	-	-	_		Constantly "0" when
	_	BZSTP	One-shot buzzer forcibly stop	Forcibly stop	No operation	-	W	being read
	D5	BZSHT	One-shot buzzer trigger/status R	Busy Trigger	Ready No operation	0	R/W	
	D4	SHTPW	One-shot buzzer duration width selection	125 msec	31.25 msec	0	R/W	
	D3	ENRTM	Envelope attenuation time	1 sec	0.5 sec	0	R/W	_
	D2	ENRST	Envelope reset	Reset	No operation	-	W	"0" when being read
	D1	ENON	Envelope On/Off control	On	Off	0	R/W	*2
	D0	BZON	Buzzer output control	On	Off	0	R/W	

<sup>\*1</sup> Writing "1" to DUTY8 (FF09•D0) disables 1/16 and 1/32 duty selection using LDUTY (FF10•D1).

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<sup>\*2</sup> Reset to "0" during one-shot output.

Table 5.6.6.1(c) Output port control bits

Address	Bit	Name			Function	1	1	0	SR	R/W	Comment
00FF40	D7	-	-				-	-	_		"0" when being read
	D6	FOUT2	FOUT fro	equency	selection				0	R/W	
			FOUT2	$\frac{\text{FOUT1}}{0}$	$\frac{\text{FOUT0}}{0}$	Frequency fosc1 / 1					
		FOUT1	0	0	1	fosci / 2				D /XX	
	DS	FOUTT	0	1	0	fosci / 4			0	R/W	
			0	1	1	fosc1 / 8					
			1	0	0	fosc3 / 1					
	D4	FOUT0	1	0	1	fosc3 / 2			0	R/W	
			1	1	0	fosc3 / 4					
			1	1	1	fosc3 / 8					
	D3	FOUTON	FOUT or	itput con	trol		On	Off	0	R/W	
	D2	WDRST	Watchdo	g timer r	eset		Reset	No operation	_	W	Constantly "0" when
	D1	TMRST	Clock tin	ner reset			Reset	No operation	_	W	being read
	D0	TMRUN	Clock tin	ner Run/	Stop cont	rol	Run	Stop	0	R/W	

#### ■ High impedance control

HZR0L, HZR0H: 00FF70H•D0, D1 HZR1L, HZR1H: 00FF70H•D2, D3

HZR20-HZR27: 00FF71H HZR30-HZR37: 00FF72H

HZR4L, HZR4H: 00FF70H•D4, D5 \*1 HZR50, HZR51: 00FF70H•D6, D7

Sets the output terminals to a high impedance state.

When "1" is written: High impedance When "0" is written: Complementary

Reading: Valid

HZRxx is the high impedance control register which correspond as shown in Table 5.6.3.1 to the various output port terminals.

When "1" is set to the HZRxx register, the corresponding output port terminal becomes high impedance state and when "0" is set, it becomes complementary output.

At initial reset, this register is set to "0" (complementary).

\*1 HZR4L and HZR4H is 2-bit reserved register, it can be used as a general purpose register with read/write capabilities.

#### ■ DC output control

R00D-R07D: 00FF73H R10D-R17D: 00FF74H R20D-R27D: 00FF75H R30D-R37D: 00FF76H R40D-R47D: 00FF77H\*1 R50D, R51D: 00FF78H•D0, D1

Sets the data output from the output port terminal Rxx.

When "1" is written: HIGH level output When "0" is written: LOW level output

Reading: Valid

RxxD is the data register for each output port. When "1" is set, the corresponding output port terminal switches to HIGH (VDD) level, and when "0" is set, it switches to LOW (VSS) level. At initial reset, R50D is set to "0" (LOW level output), all other registers are set to "1" (HIGH level output).

The output data registers set for bus signal output can be used as general purpose registers with read/write capabilities which do not affect the output terminals.

\*1 R40D-R47D is 8-bit reserved register, it can be used as a general purpose register with read/write capabilities.

#### ■ Special output control

#### LCCLK: 00FF10H•D4

Controls the CL (LCD synchronous) signal output.

When "1" is written: CL signal output When "0" is written: HIGH level (DC) output

Reading: Valid

LCCLK is the output control register for CL signal. When "1" is set, the CL signal is output from the output port terminal R25 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R25D.

At initial reset, LCCLK is set to "0" (HIGH level output).

#### LCFRM: 00FF10H•D3

Controls the FR (LCD frame) signal output.

When "1" is written: FR signal output When "0" is written: HIGH level (DC) output

Reading: Valid

LCFRM is the output control register for FR signal. When "1" is set, the FR signal is output from the output port terminal R26 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R26D.

At initial reset, LCFRM is set to "0" (HIGH level output).

#### PTOUT: 00FF30H•D2

Controls the TOUT (programmable timer output clock) signal output.

When "1" is written: TOUT signal output When "0" is written: HIGH level (DC) output

Reading: Valid

PTOUT is the output control register for TOUT signal. When "1" is set, the TOUT signal is output from the output port terminal R27 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R27D. At initial reset, PTOUT is set to "0" (HIGH level output).

#### FOUTON: 00FF40H•D3

Controls the FOUT (fosc1/fosc3 dividing clock) signal output.

When "1" is written: FOUT signal output When "0" is written: HIGH level (DC) output

Reading: Valid

FOUTON is the output control register for FOUT signal. When "1" is set, the FOUT signal is output from the output port terminal R34 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R34D. At initial reset, FOUTON is set to "0" (HIGH level output).

#### FOUT0, FOUT1, FOUT2: 00FF40H•D4, D5, D6

FOUT signal frequency is set as shown in Table 5.6.6.2.

Table 5.6.6.2 FOUT frequency settings

FOUT2	FOUT1	FOUT0	FOUT frequency
0	0	0	foscı / 1
0	0	1	foscı / 2
0	1	0	fosc1 / 4
0	1	1	fosc1 / 8
1	0	0	fosc3 / 1
1	0	1	fosc3 / 2
1	1	0	fosc3 / 4
1	1	1	fosc3 / 8

fosc1: OSC1 oscillation frequency fosc3: OSC3 oscillation frequency

At initial reset, this register is set to "0" (fosc1/1).

#### BZON: 00FF44H•D0

Controls the BZ (buzzer) signal output.

When "1" is written: BZ signal output When "0" is written: LOW level (DC) output

Reading: Valid

BZON is the output control register for BZ signal. When "1" is set, the BZ signal is output from the output port terminal R50 and when "0" is set, LOW (Vss) level is output. At this time, "0" must always be set for the data register R50D.

At initial reset, BZON is set to "0" (LOW level output).

#### BZSHT: 00FF45H•D5

Controls the one-shot buzzer output.

When "1" is written: Trigger When "0" is written: No operation

When "1" is read: **Busy** When "0" is read: Ready

Writing "1" into BZSHT causes the one-shot output circuit to operate and the BZ signal to be output. The buzzer output is automatically turned OFF after the time set by SHTPW has elapsed. At this time, "0" must always be set for the data register

The one-shot output is only valid when the normal buzzer output is OFF (BZON = "0") state. The trigger is invalid during ON (BZON = "1") state. When a re-trigger is assigned during a one-shot output, the one-shot output time set with SHTPW is measured again from that point. (time extension) The operation status of the one-shot output circuit can be confirmed by reading BZSHT, when the oneshot output is ON, BZSHT reads "1" and when the output is OFF, it reads "0".

#### BZSTP: 00FF45H•D6

Forcibly stops the one-shot buzzer output.

At initial reset, BZSHT is set to "0" (ready).

When "1" is written: Forcibly stop When "0" is written: No operation Constantly "0"

By writing "1" into BZSTP, the one-shot buzzer output can be stopped prior to the elapsing of the time set with SHTPW.

Writing "0" is invalid and writing "1" except during one-shot output is also invalid.

When "1" is written to BZSHT and BZSTP simultaneously, BZSTP takes precedence and one-shot output becomes stop status.

Since BZSTP is for writing only, during readout it is constantly set to "0".

#### 5.6.7 Programming notes

- (1) Since the special output signals (CL, FR, TOUT, FOUT and BZ) are generated asynchronously from the output control registers (LCCLK, LCFRM, PTOUT, FOUTON, BZON, BZSHT and BZSTP), when the signals is turned ON or OFF by the output control register settings, a hazard of a 1/2 cycle or less is generated.
- (2) When the FOUT frequency is made "fosc3/n", you must turn on the OSC3 oscillation circuit before outputting FOUT. A time interval of several msec to several 10 msec, from the turning ON of the OSC3 oscillation circuit to until the oscillation stabilizes, is necessary, due to the oscillation element that is used. Consequently, if an abnormality occurs as the result of an unstable FOUT signal being output externally, you should allow an adequate waiting time after turning ON of the OSC3 oscillation, before turning outputting FOUT. (The oscillation start time will vary somewhat depending on the oscillator and on the externally attached parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHAR-ACTERISTICS".) At initial reset, OSC3 oscillation circuit is set to
  - OFF state.
- (3) The SLP instruction has executed when the special output signals (TOUT, FOUT and BZ) are in the enable status, an unstable clock is output for the special output at the time of return from the SLEEP state. Consequently, when shifting to the SLEEP state, you should set the special output signal to the disable status prior to executing the SLP instruction.

#### 5.7 I/O Ports (P ports)

#### 5.7.1 Configuration of I/O ports

The S1C88349 is equipped with 16 bits of I/O ports (P00–P07, P10–P17). The configuration of these I/O ports will vary according to the bus mode as shown below.

Table 5.7.1.1 Configuration of I/O ports

T	Bus mode						
Terminal	Single chip	Expanded 64K	Expanded 512K				
P00	I/O port P00	Data b	us D0				
P01	I/O port P01	Data b	us D1				
P02	I/O port P02	Data b	us D2				
P03	I/O port P03	Data b	us D3				
P04	I/O port P04	Data b	us D4				
P05	I/O port P05	Data b	Data bus D5				
P06	I/O port P06	Data bus D6					
P07	I/O port P07	Data b	Data bus D7				
P10	I/O p	ort P10 (SIN)					
P11	I/O port P11 (SOUT)						
P12	I/O p	ort P12 (SCLK)					
P13	I/O port P13 (SRDY)						
P14	I/O port P14 (CMPP0/AD4)						
P15	I/O port P15 (CMPM0/AD5)						
P16	I/O port P16 (CMPP1/AD6)						
P17	I/O p	ort P17 (CMPM1	/AD7)				

With respect to the data bus, see "5.2 System Controller and Bus Control".

Figure 5.7.1.1 shows the structure of an I/O port.

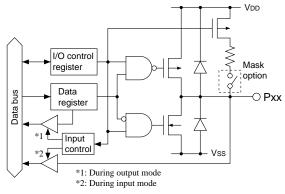


Fig. 5.7.1.1 Structure of I/O port

 $\rm I/O$  port can be set for input or output mode in one bit unit. These settings are performed by writing data to the  $\rm I/O$  control registers.

I/O port terminals P10–P13 are shared with serial interface input/output terminal and P14–P17 are shared with analog comparator and A/D converter input terminals. The function of each terminals is switchable in software.

With respect to serial interface, analog comparator and A/D converter, see "5.8 Serial Interface", "5.14 Analog Comparator" and "5.15 A/D converter", respectively.

The data registers and I/O control registers of I/O ports set for data bus and serial interface output terminals use are usable as general purpose registers with read/write capabilities which do not affect I/O activities of the terminal.

The same as above, the I/O control register of I/O port set for serial interface input terminal use is usable as general purpose register.

#### 5.7.2 Mask option

I/O port pull-up resistors	
P00 With resistor	☐ Gate direct
P01 □ With resistor	☐ Gate direct
P02 □ With resistor	$\square$ Gate direct
P03   With resistor	☐ Gate direct
P04 □ With resistor	☐ Gate direct
P05 □ With resistor	☐ Gate direct
P06 ☐ With resistor	☐ Gate direct
P07 $\square$ With resistor	$\square$ Gate direct
P10   With resistor	☐ Gate direct
P11 □ With resistor	☐ Gate direct
P12   With resistor	☐ Gate direct
P13 □ With resistor	☐ Gate direct
P14 □ With resistor	$\square$ Gate direct
P15   With resistor	☐ Gate direct
P16 □ With resistor	☐ Gate direct
P17 $\square$ With resistor	$\square$ Gate direct

I/O ports P00–P07 and P10–P17 are equipped with a pull-up resistor which goes ON in the input mode. Whether this resistor is used or not can be selected for each port (one bit unit).

In cases where the 'With resistor' option is selected, the pull-up resistor goes ON when the port is in input mode.

When changing the port terminal from LOW level to HIGH with the built-in pull-up resistor, a delay in the waveform rise time will occur depending on the time constant of the pull-up resistor and the load capacitance of the terminal. It is necessary to set an appropriate wait time for introduction of an I/O port. Make this wait time the amount of time or more calculated by the following expression.

Wait time = RIN x (CIN + load capacitance on the board) x 1.6 [sec]

RIN: Pull up resistance Max. value CIN: Terminal capacitance Max. value

When the analog comparator or A/D converter is used, select "Gate direct" for I/O ports (P14–P17) which then become input terminals.

For unused I/O ports, select the default setting of "With resistor".

#### 5.7.3 I/O control registers and I/O mode

I/O ports P00–P07 and P10–P17 are set either to input or output modes by writing data to the I/O control registers IOC00–IOC07 and IOC10–IOC17 which correspond to each bit.

To set an I/O port to input mode, write "0" to the I/O control register.

An I/O port which is set to input mode will shift to a high impedance state and functions as an input port. Readout in input mode consists simply of a direct readout of the input terminal state: the data being "1" when the input terminal is at HIGH (VDD) level and "0" when it is at LOW (VSS) level.

When the "With resistor" option is selected using the mask option, the resistor is pulled up onto the port terminal in input mode. Even in input mode, data can be written to the data registers without affecting the terminal state. To set an I/O port to output mode, write "1" to the I/O control register. An I/O port which is set to output mode functions as an output port. When port output data is "1", a HIGH (VDD) level is output and when it is "0", a LOW (Vss) level is output. Readout in output mode consists of the contents of the data register. At initial reset, I/O control registers are set to "0"

#### 5.7.4 Control of I/O ports

(I/O ports are set to input mode).

Table 5.7.4.1 shows the I/O port control bits.

Table 5.7.4.1 I/O port control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF60	D7	IOC07	P07 I/O control register					
	D6	IOC06	P06 I/O control register					
	D5	IOC05	P05 I/O control register					
	D4	IOC04	P04 I/O control register	0	Y.,4	0	R/W	
	D3	IOC03	P03 I/O control register	Output	Input	0	K/W	
	D2	IOC02	P02 I/O control register					
	D1	IOC01	P01 I/O control register					
	D0	IOC00	P00 I/O control register					
00FF61	D7	IOC17	P17 I/O control register					
	D6	IOC16	P16 I/O control register					
	D5	IOC15	P15 I/O control register					
	D4	IOC14	P14 I/O control register	Output	Tomas	0	R/W	
		IOC13	P13 I/O control register	Output	Input		IX/ VV	
	D2	IOC12	P12 I/O control register					
	D1	IOC11	P11 I/O control register					
	D0	IOC10	P10 I/O control register					
00FF62		P07D	P07 I/O port data					
	D6	P06D	P06 I/O port data					
		P05D	P05 I/O port data					
	D4	P04D	P04 I/O port data	High	Low	1	R/W	
		P03D	P03 I/O port data	Ingii	Low	•	10, 11	
		P02D	P02 I/O port data					
		P01D	P01 I/O port data					
		P00D	P00 I/O port data					
00FF63		P17D	P17 I/O port data					
		P16D	P16 I/O port data					
		P15D	P15 I/O port data					
		P14D	P14 I/O port data	High	Low	1	R/W	
		P13D	P13 I/O port data	511	25,,	•	''	
		P12D	P12 I/O port data					
		P11D	P11 I/O port data					
	D0	P10D	P10 I/O port data					

P00D-P07D: 00FF62H P10D-P17D: 00FF63H

How I/O port terminal Pxx data readout and output data settings are performed.

#### When writing data:

When "1" is written: HIGH level When "0" is written: LOW level

When the I/O port is set to output mode, the data written is output as is to the I/O port terminal. In terms of port data, when "1" is written, the port terminal goes to HIGH (VDD) level and when "0" is written to a LOW (VSS) level.

Even when the port is in input mode, data can still be written in.

#### When reading out data:

When "1" is read: HIGH level ("1")
When "0" is read: LOW level ("0")

When an I/O port is in input mode, the voltage level being input to the port terminal is read out. When terminal voltage is HIGH (VDD), it is read as a "1", and when it is LOW (VSS), it is read as a "0". Furthermore, in output mode, the contents of the data register are read out.

At initial reset, this register is set to "1" (HIGH level).

Note: The data registers of I/O ports set for the data bus and output terminal of serial interface can be used as general purpose registers with read/write capabilities which do not affect I/O activities of the terminals.

IOC00-IOC07: 00FF60H IOC10-IOC17: 00FF61H

Sets the I/O ports to input or output mode.

When "1" is written: Output mode When "0" is written: Input mode Reading: Valid

IOCxx is the I/O control register which correspond to each I/O port in a bit unit.

Writing "1" to the IOCxx register will switch the corresponding I/O port Pxx to output mode, and writing "0" will switch it to input mode.

When the analog comparator or A/D converter is used, "0" must always be set for the I/O control registers (IOC14–IOC17) of I/O ports which will become input terminals.

At initial reset, this register is set to "0" (input mode).

Note: The data registers of I/O ports set for the data bus and input terminal of serial interface can be used as general purpose registers with read/write capabilities which do not affect I/O activities of the terminals.

#### 5.7.5 Programming notes

(1) When changing the port terminal from LOW level to HIGH with the built-in pull-up resistor, a delay in the waveform rise time will occur depending on the time constant of the pull-up resistor and the load capacitance of the terminal. It is necessary to set an appropriate wait time for introduction of an I/O port. Make this wait time the amount of time or more calculated by the following expression.

## Wait time = RIN x (CIN + load capacitance on the board) x 1.6 [sec]

Rin: Pull up resistance Max. value Cin: Terminal capacitance Max. value

(2) When the analog comparator or A/D converter is used, "0" must always be set for the I/O control registers (IOC14–IOC17) of I/O ports which will become input terminals.

## 5.8 Serial Interface

#### 5.8.1 Configuration of serial interface

The S1C88349 incorporates a full duplex serial interface (when asynchronous system is selected) that allows the user to select either clock synchronous system or asynchronous system.

The data transfer method can be selected in soft-

When the clock synchronous system is selected, 8bit data transfer is possible.

When the asynchronous system is selected, either 7bit or 8-bit data transfer is possible, and a parity check of received data and the addition of a parity bit for transmitting data can automatically be done by selecting in software.

Figure 5.8.1.1 shows the configuration of the serial interface.

Serial interface input/output terminals, SIN, SOUT, SCLK and SRDY are shared with I/O ports P10-P13. In order to utilize these terminals for the serial interface input/output terminals, proper settings have to be made with registers ESIF, SMD0 and SMD1. (At initial reset, these terminals are set as I/O port terminals.)

The direction of I/O port terminals set for serial interface input/output terminals are determined by the signal and transfer mode for each terminal. Furthermore, the settings for the corresponding I/O control registers for the I/O ports become invalid.

Table 5.8.1.1 Configuration of input/output terminals

Terminal	When serial interface is selected
P10	SIN
P11	SOUT
P12	SCLK
P13	$\overline{SRDY}$

<sup>\*</sup> The terminals used may vary depending on the transfer mode.

SIN and SOUT are serial data input and output terminals which function identically in clock synchronous system and asynchronous system. SCLK is exclusively for use with clock synchronous system and functions as a synchronous clock input/ output terminal. SRDY is exclusively for use in clock synchronous slave mode and functions as a sendreceive ready signal output terminal. When asynchronous system is selected, since SCLK

and SRDY are superfluous, the I/O port terminals P12 and P13 can be used as I/O ports. In the same way, when clock synchronous master

mode is selected, since SRDY is superfluous, the I/O port terminal P13 can be used as I/O port.

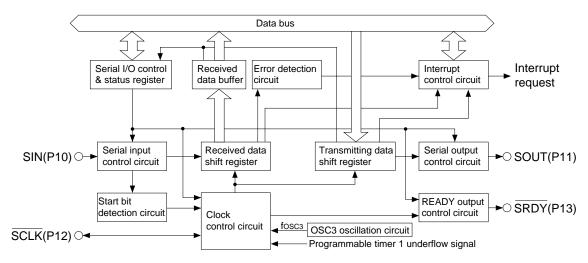
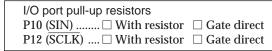


Fig. 5.8.1.1 Configuration of serial interface

#### 5.8.2 Mask option

Since serial interface input/output terminals are shared with the I/O ports, serial interface terminal specifications have necessarily been selected with the mask option for I/O ports.



Each I/O port terminal is equipped with a pull-up resistor which goes ON in input mode. A selection can be made for each port (one bit unit) as to whether or not the resistor will be used. Specifications (whether the pull-up will be used or not) of P10 (SIN) and P12 (SCLK) which will become input terminals when using the serial interface are decided by settings the options for the I/O port.

When "Gate direct" is selected in the serial I/F mode, be sure that the input terminals do not go into a floating state.

#### 5.8.3 Transfer modes

There are four transfer modes for the serial interface and mode selection is made by setting the two bits of the mode selection registers SMD0 and SMD1 as shown in the table below.

Table 5.8.3.1 Transfer modes

SMD1	SMD0	Mode
1	1	Asynchronous 8-bit
1	0	Asynchronous 7-bit
0	1	Clock synchronous slave
0	0	Clock synchronous master

Table 5.8.3.2 Terminal settings corresponding to each transfer mode

		<u>'</u>		
Mode	SIN	SOUT	SCLK	SRDY
Asynchronous 8-bit	Input	Output	P12	P13
Asynchronous 7-bit	Input	Output	P12	P13
Clock synchronous slave	Input	Output	Input	Output
Clock synchronous master	Input	Output	Output	P13

At initial reset, transfer mode is set to clock synchronous master mode.

#### ■ Clock synchronous master mode

In this mode, the internal clock is utilized as a synchronous clock for the built-in shift registers, and clock synchronous 8-bit serial transfers can be performed with this serial interface as the master. The synchronous clock is also output from the SCLK terminal which enables control of the external (slave side) serial I/O device. Since the SRDY terminal is not utilized in this mode, it can be used as an I/O port.

Figure 5.8.3.1(a) shows the connection example of input/output terminals in the clock synchronous master mode.

#### ■ Clock synchronous slave mode

In this mode, a synchronous clock from the external (master side) serial input/output device is utilized and clock synchronous 8-bit serial transfers can be performed with this serial interface as the slave. The synchronous clock is input to the  $\overline{SCLK}$  terminal and is utilized by this interface as the synchronous clock.

Furthermore, the  $\overline{SRDY}$  signal indicating the transmit-receive ready status is output from the  $\overline{SRDY}$  terminal in accordance with the serial interface operating status.

In the slave mode, the settings for registers SCS0 and SCS1 used to select the clock source are invalid. Figure 5.8.3.1(b) shows the connection example of input/output terminals in the clock synchronous slave mode.

#### ■ Asynchronous 7-bit mode

In this mode, asynchronous 7-bit transfer can be performed. Parity check during data reception and addition of parity bit (odd/even/none) during transmitting can be specified and data processed in 7 bits with or without parity. Since this mode employs the internal clock, the SCLK terminal is not used. Furthermore, since the SRDY terminal is not utilized either, both of these terminals can be used as I/O ports.

Figure 5.8.3.1(c) shows the connection example of input/output terminals in the asynchronous mode.

#### Asynchronous 8-bit mode

In this mode, asynchronous 8-bit transfer can be performed. Parity check during data reception and addition of parity bit (odd/even/none) during transmitting can be specified and data processed in 8 bits with or without parity. Since this mode employs the internal clock, the SCLK terminal is not used. Furthermore, since the SRDY terminal is not utilized either, both of these terminals can be used as I/O ports.

Figure 5.8.3.1(c) shows the connection example of input/output terminals in the asynchronous mode.

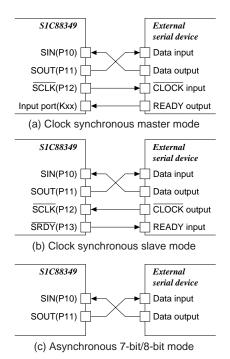


Fig. 5.8.3.1 Connection examples of serial interface I/O terminals

#### 5.8.4 Clock source

There are four clock sources and selection is made by setting the two bits of the clock source selection register SCS0 and SCS1 as shown in table below.

Table 5.8.4.1 Clock source

SCS1	SCS0	Clock source
1	1	Programmable timer
1	0	fosc3 / 4
0	1	fosc3 / 8
0	0	fosc3 / 16

This register setting is invalid in clock synchronous slave mode and the external clock input from the SCLK terminal is used.

When the "programmable timer" is selected, the programmable timer 1 underflow signal is divided by 1/2 and this signal used as the clock source. With respect to the transfer rate setting, see "5.11 Programmable Timer".

At initial reset, the synchronous clock is set to "fosc3/16".

Whichever clock is selected, the signal is further divided by 1/16 and then used as the synchronous clock.

Furthermore, external clock input is used as is for SCLK in clock synchronous slave mode.

Table 5.8.4.2 shows an examples of transfer rates and OSC3 oscillation frequencies when the clock source is set to programmable timer.

When the demultiplied signal of the OSC3 oscillation circuit is made the clock source, it is necessary to turn the OSC3 oscillation ON, prior to using the serial interface.

A time interval of several msec to several 10 msec, from the turning ON of the OSC3 oscillation circuit to until the oscillation stabilizes, is necessary, due to the oscillation element that is used. Consequently, you should allow an adequate waiting time after turning ON of the OSC3 oscillation, before starting transmitting/receiving of serial interface. (The oscillation start time will vary somewhat depending on the oscillator and on the externally attached parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".) At initial reset, the OSC3 oscillation circuit is set to OFF status.

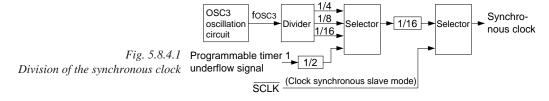


Table 5.8.4.2 OSC3 oscillation frequencies and transfer rates

Transfer rate	OSC3 oscillation frequency / Programmable timer settings						
	fosc3 = 3.072 MHz		fosc3 = 4.608 MHz		fosc3 = 4.9152 MH		
(bps)	PSC1X	RLD1X	PSC1X	RLD1X	PSC1X	RLD1X	
9,600	0 (1/1)	09H	0 (1/1)	0EH	0 (1/1)	0FH	
4,800	0 (1/1)	13H	0 (1/1)	1DH	0 (1/1)	1FH	
2,400	0 (1/1)	27H	0 (1/1)	3BH	0 (1/1)	3FH	
1,200	0 (1/1)	4FH	0 (1/1)	77H	0 (1/1)	7FH	
600	0 (1/1)	9FH	0 (1/1)	EFH	0 (1/1)	FFH	
300	1 (1/4)	4FH	1 (1/4)	77H	1 (1/4)	7FH	
150	1 (1/4)	9FH	1 (1/4)	EFH	1 (1/4)	FFH	

#### 5.8.5 Transmit-receive control

Below is a description of the registers which handle transmit-receive control. With respect to transmitreceive control procedures and operations, please refer to the following sections in which these are discussed on a mode by mode basis.

#### ■ Shift register and received data buffer

Exclusive shift registers for transmitting and receiving are installed in this serial interface. Consequently, duplex communication simultaneous transmit and receive is possible when the asynchronous system is selected.

Data being transmitted are written to TRXD0–TRXD7 and converted to serial through the shift register and is output from the SOUT terminal.

In the reception section, a received data buffer is installed separate from the shift register. Data being received are input to the SIN terminal and is converted to parallel through the shift register and written to the received data buffer. Since the received data buffer can be read even during serial input operation, the continuous data is received efficiently.

However, since buffer functions are not used in clock synchronous mode, be sure to read out data before the next data reception begins.

## Transmit enable register and transmit control bit

For transmitting control, use the transmit enable register TXEN and transmit control bit TXTRG.

The transmit enable register TXEN is used to set the transmitting enable/disable status. When "1" is written to this register to set the transmitting enable status, clock input to the shift register is enabled and the system is ready to transmit data. In the clock synchronous mode, synchronous clock input/output from the  $\overline{SCLK}$  terminal is also enabled.

The transmit control bit TXTRG is used as the trigger to start transmitting data.

Data to be transmitted is written to the transmit data shift register, and when transmitting preparations a recomplete, "1" is written to TXTRG whereupon data transmitting begins.

When interrupt has been enabled, an interrupt is generated when the transmission is completed. If there is subsequent data to be transmitted it can be sent using this interrupt.

In addition, TXTRG can be read as the status. When set to "1", it indicates transmitting operation, and "0" indicates transmitting stop.

For details on timing, see the timing chart which gives the timing for each mode.

When not transmitting, set TXEN to "0" to disable transmitting status.

#### ■ Receive enable register, receive control bit

For receiving control, use the receive enable register RXEN and receive control bit RXTRG.

Receive enable register RXEN is used to set receiving enable/disable status. When "1" is written into this register to set the receiving enable status, clock input to the shift register is enabled and the system is ready to receive data. In the clock synchronous mode, synchronous clock input/output from the  $\overline{SCLK}$  terminal is also enabled.

With the above setting, receiving begins and serial data input from the SIN terminal goes to the shift register.

The operation of the receive control bit RXTRG is slightly different depending on whether a clock synchronous system or an asynchronous system is being used.

In the clock synchronous system, the receive control bit RXTRG is used as the trigger to start receiving data.

When received data has been read and the preparation for next data receiving is completed, write "1" into RXTRG to start receiving. (When "1" is written to RXTRG in slave mode, \$\overline{SRDY}\$ switches to "0".) In an asynchronous system, RXTRG is used to prepare for next data receiving. After reading the received data from the received data buffer, write "1" into RXTRG to signify that the received data buffer is empty. If "1" is not written into RXTRG, the overrun error flag OER will be set to "1" when the next receiving operation is completed. (An overrun error will be generated when receiving is completed between reading the received data and the writing of "1" to RXTRG.)

In addition, RXTRG can be read as the status. In either clock synchronous mode or asynchronous mode, when RXTRG is set to "1", it indicates receiving operation and when set to "0", it indicates that receiving has stopped.

For details on timing, see the timing chart which gives the timing for each mode.

When you do not receive, set RXEN to "0" to disable receiving status.

# 5.8.6 Operation of clock synchronous transfer

Clock synchronous transfer involves the transfer of 8-bit data by synchronizing it to eight clocks. The same synchronous clock is used by both the transmitting and receiving sides.

When the serial interface is used in the master mode, the clock signal selected using SCS0 and SCS1 is further divided by 1/16 and employed as the synchronous clock. This signal is then sent via the SCLK terminal to the slave side (external serial I/O device).

When used in the slave mode, the clock input to the SCLK terminal from the master side (external serial input/output device) is used as the synchronous clock.

In the clock synchronous mode, since one clock line (SCLK) is shared for both transmitting and receiving, transmitting and receiving cannot be performed simultaneously. (Half duplex only is possible in clock synchronous mode.)

Transfer data is fixed at 8 bits and both transmitting and receiving are conducted with the LSB (bit 0) coming first.



Fig. 5.8.6.1 Transfer data configuration using clock synchronous mode

Below is a description of initialization when performing clock synchronous transfer, transmit-receive control procedures and operations. With respect to serial interface interrupt, see "5.8.8 Interrupt function".

#### ■ Initialization of serial interface

When performing clock synchronous transfer, the following initial settings must be made.

(1) Setting of transmitting/receiving disable

To set the serial interface into a status in which
both transmitting and receiving are disabled, "0"
must be written to both the transmit enable
register TXEN and the receive enable register
RXEN. Fix these two registers to a disable status
until data transfer actually begins.

#### (2) Port selection

Because serial interface input/output ports SIN, SOUT, SCLK and SRDY are set as I/O port terminals P10–P13 at initial reset, "1" must be written to the serial interface enable register ESIF in order to set these terminals for serial interface use.

#### (3) Setting of transfer mode

Select the clock synchronous mode by writing the data as indicated below to the two bits of the mode selection registers SMD0 and SMD1.

Master mode: SMD0 = "0", SMD1 = "0" Slave mode: SMD0 = "1", SMD1 = "0"

#### (4) Clock source selection

In the master mode, select the synchronous clock source by writing data to the two bits of the clock source selection registers SCS0 and SCS1. (See Table 5.8.4.1.)

This selection is not necessary in the slave mode.

Since all the registers mentioned in (2)–(4) are assigned to the same address, it's possible to set them all with one instruction. The parity enable register EPR is also assigned to this address, however, since parity is not necessary in the clock synchronous mode, parity check will not take place regardless of how they are set.

#### (5) Clock source control

When the master mode is selected and programmable timer for the clock source is selected, set transfer rate on the programmable timer side. (See "5.11 Programmable Timer".) When the divided signal of OSC3 oscillation circuit is selected for the clock source, be sure that the OSC3 oscillation circuit is turned ON prior to commencing data transfer. (See "5.4 Oscillation Circuits and Operating Mode".)

#### ■ Data transmit procedure

The control procedure and operation during transmitting is as follows.

- Write "0" in the transmit enable register TXEN and the receive enable register RXEN to reset the serial interface.
- (2) Write "1" in the transmit enable register TXEN to set into the transmitting enable status.
- (3) Write the transmitting data into TRXD0-TRXD7.
- (4) In case of the master mode, confirm the receive ready status on the slave side (external serial input/output device), if necessary. Wait until it reaches the receive ready status.
- (5) Write "1" in the transmit control bit TXTRG and start transmitting.

In the master mode, this control causes the synchronous clock to change to enable and to be provided to the shift register for transmitting and output from the  $\overline{SCLK}$  terminal. In the slave mode, it waits for the synchronous clock to be input from the  $\overline{SCLK}$  terminal. The transmitting data of the shift register shifts one bit at a time at each falling edge of the synchronous clock and is output from the SOUT terminal. When the final bit (MSB) is output, the SOUT terminal is maintained at that level, until the next transmitting begins.

The transmitting complete interrupt factor flag FSTRA is set to "1" at the point where the data transmitting of the shift register is completed. When interrupt has been enabled, a transmitting complete interrupt is generated at this point.

Set the following transmitting data using this interrupt.

(6) Repeat steps (3) to (5) for the number of bytes of transmitting data, and then set the transmit disable status by writing "0" to the transmit enable register TXEN, when the transmitting is completed.

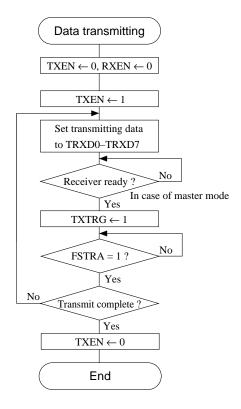


Fig. 5.8.6.2 Transmit procedure in clock synchronous mode

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#### ■ Data receive procedure

The control procedure and operation during receiving is as follows.

- Write "0" in the receive enable register RXEN and transmit enable register TXEN to reset the serial interface.
- (2) Write "1" in the receive enable register RXEN to set into the receiving enable status.
- (3) In case of the master mode, confirm the transmit ready status on the slave side (external serial input/output device), if necessary. Wait until it reaches the transmit ready status.
- (4) Write "1" in the receive control bit RXTRG and start receiving.

In the master mode, this control causes the synchronous clock to change to enable and is provided to the shift register for receiving and output from the  $\overline{SCLK}$  terminal.

In the slave mode, it waits for the synchronous clock to be input from the  $\overline{SCLK}$  terminal. The received data input from the SIN terminal is successively incorporated into the shift register in synchronization with the rising edge of the synchronous clock.

At the point where the data of the 8th bit has been incorporated at the final (8th) rising edge of the synchronous clock, the content of the shift register is sent to the received data buffer and the receiving complete interrupt factor flag FSREC is set to "1". When interrupt has been enabled, a receiving complete interrupt is generated at this point.

- Read the received data from TRXD0-TRXD7 using receiving complete interrupt.
- (6) Repeat steps (3) to (5) for the number of bytes of receiving data, and then set the receive disable status by writing "0" to the receive enable register RXEN, when the receiving is completed.

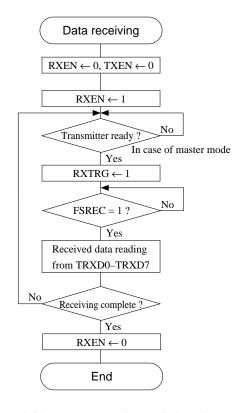
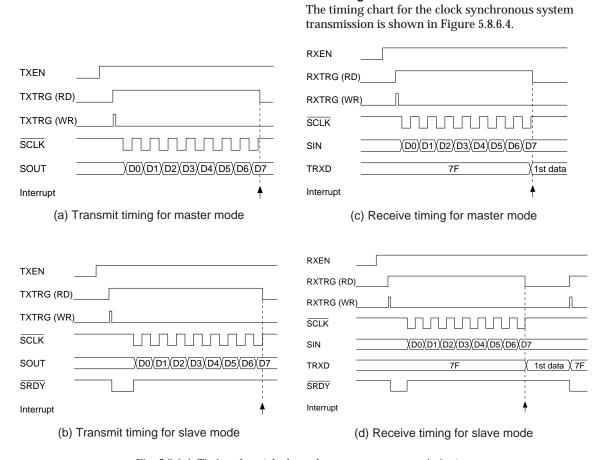


Fig. 5.8.6.3 Receiving procedure in clock synchronous mode

#### ■ Transmit/receive ready (SRDY) signal

When this serial interface is used in the clock synchronous slave mode (external clock input), an  $\overline{SRDY}$  signal is output to indicate whether or not this serial interface can transmit/receive to the master side (external serial input/output device). This signal is output from the  $\overline{SRDY}$  terminal and when this interface enters the transmit or receive enable (READY) status, it becomes "0" (LOW level) and becomes "1" (HIGH level) when there is a BUSY status, such as during transmit/receive operation.



The  $\overline{SRDY}$  signal changes the "1" to "0," immediately after writing "1" into the transmit control bit

returns from "0" to "1", at the point where the first

When you have set in the master mode, control the

transfer by inputting the same signal from the slave

side using the input port or I/O port. At this time,

since the SRDY terminal is not set and instead P13

functions as the I/O port, you can apply this port

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for said control.

■ Timing chart

synchronous clock has been input (falling edge).

TXTRG or the receive control bit RXTRG and

Fig. 5.8.6.4 Timing chart (clock synchronous system transmission)

#### 5.8.7 Operation of asynchronous transfer

Asynchronous transfer is a mode that transfers by adding a start bit and a stop bit to the front and the back of each piece of serial converted data. In this mode, there is no need to use a clock that is fully synchronized clock on the transmit side and the receive side, but rather transmission is done while adopting the synchronization at the start/stop bits that have attached before and after each piece of data. The RS-232C interface functions can be easily realized by selecting this transfer mode.

This interface has separate transmit and receive shift registers and is designed to permit full duplex transmission to be done simultaneously for transmitting and receiving.

For transfer data in the asynchronous 7-bit mode, either 7 bits data (no parity) or 7 bits data + parity bit can be selected. In the asynchronous 8-bit mode, either 8 bits data (no parity) or 8 bits data + parity bit can be selected.

Parity can be even or odd, and parity checking of received data and adding a party bit to transmitting data will be done automatically. Thereafter, it is not necessary to be conscious of parity itself in the program.

The start bit and stop bit are respectively fixed at one bit and data is transmitted and received by placing the LSB (bit 0) at the front.

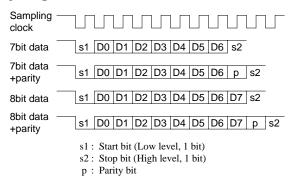


Fig. 5.8.7.1 Transfer data configuration for asynchronous system

Here following, we will explain the control sequence and operation for initialization and transmitting /receiving in case of asynchronous data transfer. See "5.8.8 Interrupt function" for the serial interface interrupts.

#### Initialization of serial interface

The below initialization must be done in cases of asynchronous system transfer.

(1) Setting of transmitting/receiving disable
To set the serial interface into a status in which
both transmitting and receiving are disabled, "0"
must be written to both the transmit enable
register TXEN and the receive enable register
RXEN. Fix these two registers to a disable status
until data transfer actually begins.

#### (2) Port selection

Because serial interface input/output terminals SIN and SOUT are set as I/O port terminals P10 and P11 at initial reset, "1" must be written to the serial interface enable register ESIF in order to set these terminals for serial interface use. SCLK and SRDY terminals set in the clock synchronous mode are not used in the asynchronous mode. These terminals function as I/O port terminals P12 and P13.

# (3) Setting of transfer mode Select the asynchronous mode by writing the data as indicated below to the two bits of the

mode selection registers SMD0 and SMD1.

7-bit mode: SMD0 = "0", SMD1 = "1" 8-bit mode: SMD0 = "1", SMD1 = "1"

#### (4) Parity bit selection

When checking and adding parity bits, write "1" into the parity enable register EPR to set to "with parity check". As a result of this setting, in the asynchronous 7-bit mode, it has a 7 bits data + parity bit configuration and in the asynchronous 8-bit mode it has an 8 bits data + parity bit configuration. In this case, parity checking for receiving and adding a party bit for transmitting is done automatically in hardware. Moreover, when "with parity check" has been selected, "odd" or "even" parity must be further selected in the parity mode selection register PMD. When "0" is written to the EPR register to select "without parity check" in the asynchronous 7-bit mode, data configuration is set to 7 bits data (no parity) and in the asynchronous 8-bit mode (no parity) it is set to 8 bits data (no parity) and parity checking and parity bit adding will not be done.

#### (5) Clock source selection

Select the clock source by writing data to the two bits of the clock source selection registers SCS0 and SCS1. (See Table 5.8.4.1.)

Since all the registers mentioned in (2)–(5) are assigned to the same address, it's possible to set them all with one instruction.

#### (6) Clock source control

When the programmable timer is selected for the clock source, set transfer rate on the programmable timer side. (See "5.11 Programmable Timer".)

When the divided signal of OSC3 oscillation circuit is selected for the clock source, be sure that the OSC3 oscillation circuit is turned ON prior to commencing data transfer. (See "5.4 Oscillation Circuits and Operating Mode".)

#### ■ Data transmit procedure

The control procedure and operation during transmitting is as follows.

- (1) Write "0" in the transmit enable register TXEN to reset the serial interface.
- (2) Write "1" in the transmit enable register TXEN to set into the transmitting enable status.
- (3) Write the transmitting data into TRXD0-TRXD7. Also, when 7-bit data is selected, the TRXD7 data becomes invalid.
- (4) Write "1" in the transmit control bit TXTRG and start transmitting.

This control causes the shift clock to change to enable and a start bit (LOW) is output to the SOUT terminal in synchronize to its rising edge. The transmitting data set to the shift register is shifted one bit at a time at each rising edge of the clock thereafter and is output from the SOUT terminal. After the data output, it outputs a stop bit (HIGH) and HIGH level is maintained until the next start bit is output.

The transmitting complete interrupt factor flag FSTRA is set to "1" at the point where the data transmitting is completed. When interrupt has been enabled, a transmitting complete interrupt is generated at this point.

Set the following transmitting data using this interrupt.

(5) Repeat steps (3) to (4) for the number of bytes of transmitting data, and then set the transmit disable status by writing "0" to the transmit enable register TXEN, when the transmitting is completed.

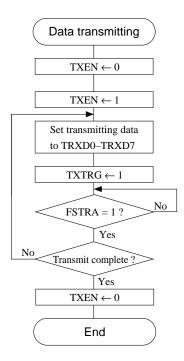


Fig. 5.8.7.2 Transmit procedure in asynchronous mode

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#### ■ Data receive procedure

The control procedure and operation during receiving is as follows.

- Write "0" in the receive enable register RXEN to set the receiving disable status and to reset the respective PER, OER, FER flags that indicate parity, overrun and framing errors.
- (2) Write "1" in the receive enable register RXEN to set into the receiving enable status.
- (3) The shift clock will change to enable from the point where the start bit (LOW) has been input from the SIN terminal and the receive data will be synchronized to the rising edge following the second clock, and will thus be successively incorporated into the shift register. After data bits have been incorporated, the stop bit is checked and, if it is not HIGH, it becomes a framing error and the error interrupt factor flag FSERR is set to "1". When interrupt has been enabled, an error interrupt is generated at this point. When receiving is completed, data in the shift register is transferred to the received data buffer and the receiving complete interrupt flag FSREC is set to "1". When interrupt has been enabled, a receiving complete interrupt is generated at this point. (When an overrun error is generated, the interrupt factor flag FSREC is not set to "1" and a receiving complete interrupt is not generated.) If "with parity check" has been selected, a parity check is executed when data is transferred into the received data buffer from the shift register and if a parity error is detected, the error interrupt factor flag is set to "1". When the interrupt has been enabled, an error interrupt is generated at this point just as in the framing error mentioned above.
- (4) Read the received data from TRXD0-TRXD7 using receiving complete interrupt.
- (5) Write "1" to the receive control bit RXTRG to inform that the receive data has been read out. When the following data is received prior to writing "1" to RXTRG, it is recognized as an overrun error and the error interrupt factor flag is set to "1". When the interrupt has been enabled, an error interrupt is generated at this point just as in the framing error and parity error mentioned above.
- (6) Repeat steps (3) to (5) for the number of bytes of receiving data, and then set the receive disable status by writing "0" to the receive enable register RXEN, when the receiving is completed.

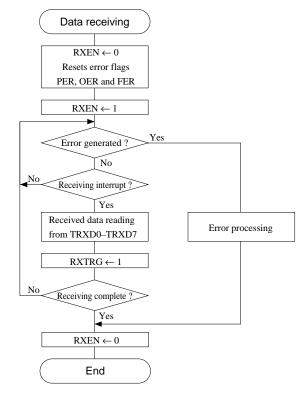


Fig. 5.8.7.3 Receiving procedure in asynchronous mode

#### ■ Receive error

During receiving the following three types of errors can be detected by an interrupt.

#### (1) Parity error

When writing "1" to the EPR register to select "with parity check", a parity check (vertical parity check) is executed during receiving. After each data bit is sent a parity check bit is sent. The parity check bit is a "0" or a "1". Even parity checking will cause the sum of the parity bit and the other bits to be even. Odd parity causes the sum to be odd. This is checked on the receiving side.

The parity check is performed when data received in the shift register is transferred to the received data buffer. It checks whether the parity check bit is a "1" or a "0" (the sum of the bits including the parity bit) and the parity set in the PMD register match. When it does not match, it is recognized as an parity error and the parity error flag PER and the error interrupt factor flag FSERR are set to "1".

When interrupt has been enabled, an error interrupt is generated at this point.

interrupt is generated at this point.
The PER flag is reset to "0" by writing "1".
Even when this error has been generated, the received data corresponding to the error is transferred in the received data buffer and the receive operation also continues.

The received data at this point cannot assured because of the parity error.

#### (2) Framing error

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In asynchronous transfer, synchronization is adopted for each character at the start bit ("0") and the stop bit ("1"). When receiving has been done with the stop bit set at "0", the serial interface judges the synchronization to be off and a framing error is generated. When this error is generated, the framing error flag FER and the error interrupt factor flag FSERR are set to "1". When interrupt has been enabled, an error interrupt is generated at this point. The FER flag is reset to "0" by writing "1". Even when this error has been generated, the received data for it is loaded into the receive data buffer and the receive operation also continues. However, even when it does not become a framing error with the following data receipt, such data cannot be assured. Even when this error has been generated, the received data corresponding to the error is transferred in the received data buffer and the receive operation also continues. However, even when it does not become a framing error with the following data receiving, such data cannot be assured.

#### (3) Overrun error

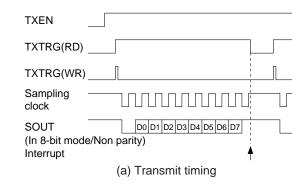
When the next data is received before "1" is written to RXTRG, an overrun error will be generated, because the previous receive data will be overwritten. When this error is generated, the overrun error flag OER and the error interrupt factor flag FSERR are set to "1". When interrupt has been enabled, an error interrupt is generated at this point. The OER flag is reset to "0" by writing "1" into it. Even when this error has been generated, the received data corresponding to the error is transferred in the received data buffer and the receive operation also continues. Furthermore, when the timing for writing "1" to DYTEC and the viscing for the president of the second of the second

Furthermore, when the timing for writing "1" to RXTRG and the timing for the received data transfer to the received data buffer overlap, it will be recognized as an overrun error.

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#### ■ Timing chart

Figure 5.8.7.4 show the asynchronous transfer timing chart.



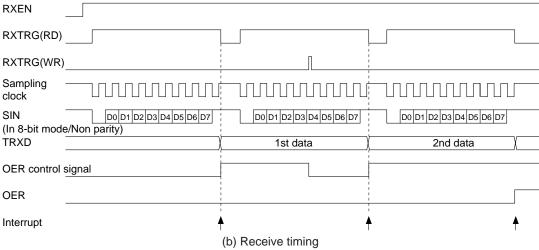


Fig. 5.8.7.4 Timing chart (asynchronous transfer)

#### 5.8.8 Interrupt function

This serial interface includes a function that generates the below indicated three types of interrupts.

- Transmitting complete interrupt
- Receiving complete interrupt
- Error interrupt

The interrupt factor flag FSxxx and the interrupt enable register ESxxx for the respective interrupt factors are provided and then the interrupt enable/disable can be selected by the software. In addition, a priority level of the serial interface interrupt for the CPU can be optionally set at levels 0 to 3 by the interrupt priority registers PSIF0 and PSIF1. For details on the above mentioned interrupt control register and the operation following generation of an interrupt, see "5.17 Interrupt and Standby Status".

Figure 5.8.8.1 shows the configuration of the serial interface interrupt circuit.

#### ■ Transmitting complete interrupt

This interrupt factor is generated at the point where the sending of the data written into the shift register has been completed and sets the interrupt factor flag FSTRA to "1". When set in this manner, if the corresponding interrupt enable register ESTRA is set to "1" and the corresponding interrupt priority registers PSIF0 and PSIF1 are set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. When "0" has been written into the interrupt enable

register ESTRA and interrupt has been disabled, an interrupt is not generated to the CPU. Even in this case, the interrupt factor flag FSTRA is set to "1". The interrupt factor flag FSTRA is reset to "0" by writing "1".

The following transmitting data can be set and the transmitting start (writing "1" to TXTRG) can be controlled by generation of this interrupt factor. The exception processing vector address for this interrupt factor is set at 000014H.

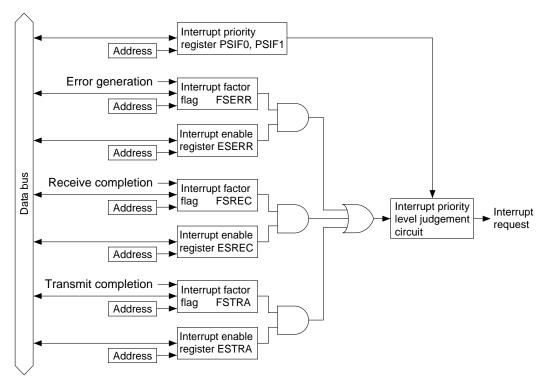


Fig. 5.8.8.1 Configuration of serial interface interrupt circuit

#### ■ Receiving complete interrupt

This interrupt factor is generated at the point where receiving has been completed and the receive data incorporated into the shift register has been transferred into the received data buffer and it sets the interrupt factor flag FSREC to "1". When set in this manner, if the corresponding interrupt enable register ESREC is set to "1" and the corresponding interrupt priority registers PSIF0 and PSIF1 are set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. When "0" has been written into the interrupt enable register ESREC and interrupt has been disabled, an interrupt is not generated to the CPU. Even in this case, the interrupt factor flag FSREC is set to "1". The interrupt factor flag FSREC is reset to "0" by writing "1".

The generation of this interrupt factor permits the received data to be read.

Also, the interrupt factor flag is set to "1" when a parity error or framing error is generated.

The exception processing vector address for this interrupt factor is set at 000012H.

#### ■ Error interrupt

This interrupt factor is generated at the point where a parity error, framing error or overrun error is detected during receiving and it sets the interrupt factor flag FSERR to "1". When set in this manner, if the corresponding interrupt enable register ESERR is set to "1" and the corresponding interrupt priority registers PSIF0 and PSIF1 are set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. When "0" has been written in the interrupt enable register ESERR and interrupt has been disabled, an interrupt is not generated to the CPU. Even in this case, the interrupt factor flag FSERR is set to "1". The interrupt factor flag FSERR is reset to "0" by writing "1".

Since all three types of errors result in the same interrupt factor, you should identify the error that has been generated by the error flags PER (parity error), OER (overrun error) and FER (framing error).

The exception processing vector address for this interrupt factor is set at 000010H.

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## 5.8.9 Control of serial interface

Table 5.8.9.1 show the serial interface control bits.

Table 5.8.9.1(a) Serial interface control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF48	D7	_	_	_	-	_		"0" when being read
	D6	EPR	Parity enable register	With parity	Non parity	0	R/W	Only for
	D5	PMD	Parity mode selection	Odd	Even	0	R/W	asynchronous mode
	D4	SCS1	Clock source selection			0	R/W	
			SCS1 SCS0 Clock source					In the clock synchro-
			1 1 Programmable timer					nous slave mode,
	D3	SCS0	1 0 fosc3 / 4			0	R/W	external clock is
			0 1 fosc3 / 8					selected.
			0 0 fosc3 / 16					
	D2	SMD1	Serial I/F mode selection			0	R/W	
			SMD1 SMD0 Mode					
			1 1 Asynchronous 8-bit					
	D1	SMD0	1 0 Asynchronous 7-bit			0	R/W	
			0 1 Clock synchronous slave					
			0 0 Clock synchronous master					
		ESIF	Serial I/F enable register	Serial I/F	I/O port	0	R/W	
00FF49	D7	_		-	-	_		"0" when being read
	D6	FER	Framing error flag R	Error	No error	0	R/W	
			W	Reset (0)	No operation			
	D5	PER	Parity error flag R	Error	No error	0	R/W	Only for
			W	Reset (0)	No operation			asynchronous mode
	D4	OER	Overrun error flag	Error	No error	0	R/W	
			W	Reset (0)	No operation			
	D3	RXTRG	Receive trigger/status R	Run	Stop	0	R/W	
			¦ W	Trigger	No operation			
		RXEN	Receive enable	Enable	Disable	0	R/W	
	D1	TXTRG	Transmit trigger/status R	Run	Stop	0	R/W	
			¦ W	Trigger	No operation			
		TXEN	Transmit enable	Enable	Disable	0	R/W	
00FF4A		TRXD7	Transmit/Receive data D7 (MSB)					
		TRXD6	Transmit/Receive data D6					
		TRXD5	Transmit/Receive data D5					
		TRXD4	Transmit/Receive data D4	High	Low	X	R/W	
		TRXD3	Transmit/Receive data D3					
		TRXD2	Transmit/Receive data D2					
	D1 TRXD1 Transmit/Receive data D1							
		TRXD0	Transmit/Receive data D0 (LSB)					
00FF20		PK01	K00–K07 interrupt priority register			0	R/W	
		PK00		PK01 PK00 PSIF1 PSIF0 PSW1 PSW0 Priority				
		PSIF1	Serial interface interrupt priority register			0	R/W	
	_	PSIF0		$\frac{\text{PTM1}}{1} \frac{\text{PTM}}{1}$	louel Level 3			
		PSW1	Stopwatch timer interrupt priority register	1 0	Level 2	0	R/W	
		PSW0		$\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}$	Level 1 Level 0		-	
		PTM1	Clock timer interrupt priority register		Level 0	0	R/W	
	טט	PTM0						

Table 5.8.9.1(b) Serial interface control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF23	D7	EPT1	Programmable timer 1 interrupt enable register					
	D6	EPT0	Programmable timer 0 interrupt enable register					
	D5	EK1	K10 and K11 interrupt enable register					
	D4	EK0H	K04–K07 interrupt enable register	Interrupt	Interrupt	0	R/W	
	D3	EK0L	K00-K03 interrupt enable register	enable	disable	U	IN/ W	
	D2	ESERR	Serial I/F (error) interrupt enable register					
	D1	ESREC	Serial I/F (receiving) interrupt enable register					
	D0	ESTRA	Serial I/F (transmitting) interrupt enable register					
00FF25	D7	FPT1	Programmable timer 1 interrupt factor flag	(R)	(R)			
	D6	FPT0	Programmable timer 0 interrupt factor flag	Interrupt	No interrupt			
	D5	FK1	K10 and K11 interrupt factor flag	factor is	factor is			
	D4	FK0H	K04–K07 interrupt factor flag	generated	generated	0	R/W	
	D3	FK0L	K00-K03 interrupt factor flag			U	IN/ W	
	D2	FSERR	Serial I/F (error) interrupt factor flag	(W)	(W)			
	D1	FSREC	Serial I/F (receiving) interrupt factor flag	Reset	No operation			
	D0	FSTRA	Serial I/F (transmitting) interrupt factor flag					

#### ESIF: 00FF48H•D0

Sets the serial interface terminals (P10-P13).

When "1" is written: Serial input/output terminal When "0" is written: I/O port terminal

Reading: Valid

The ESIF is the serial interface enable register and P10–P13 terminals become serial input/output terminals (SIN, SOUT,  $\overline{SCLK}$ ,  $\overline{SRDY}$ ) when "1" is written, and they become I/O port terminals when "0" is written.

Also, see Table 5.8.3.2 for the terminal settings according to the transfer modes.

At initial reset, ESIF is set to "0" (I/O port).

#### SMD0, SMD1: 00FF48H•D1, D2

Set the transfer modes according to Table 5.8.9.2.

Table 5.8.9.2 Transfer mode settings

SMD1	SMD0	Mode
1	1	Asynchronous 8-bit
1	0	Asynchronous 7-bit
0	1	Clock synchronous slave
0	0	Clock synchronous master

SMD0 and SMD1 can also read out. At initial reset, this register is set to "0" (clock synchronous master mode).

#### SCS0, SCS1: 00FF48H•D3, D4

Select the clock source according to Table 5.8.9.3.

Table 5.8.9.3 Clock source selection

SCS1	SCS0	Clock source
1	1	Programmable timer
1	0	fosc3 / 4
0	1	fosc3 / 8
0	0	fosc3 / 16

SCS0 and SCS1 can also be read out.

In the clock synchronous slave mode, setting of this register is invalid.

At initial reset, this register is set to "0" (fosc3/16).

#### EPR: 00FF48H•D6

Selects the parity function.

When "1" is written: With parity When "0" is written: Non parity Reading: Valid

Selects whether or not to check parity of the received data and to add a parity bit to the transmitting data. When "1" is written to EPR, the most significant bit of the received data is considered to be the parity bit and a parity check is executed. A parity bit is added to the transmitting data. When "0" is written, neither checking is done nor is a parity bit added.

Parity is valid only in asynchronous mode and the EPR setting becomes invalid in the clock synchronous mode.

At initial reset, EPR is set to "0" (non parity).

#### PMD: 00FF48H•D5

Selects odd parity/even parity.

When "1" is written: Odd parity When "0" is written: Even parity Reading: Valid

When "1" is written to PMD, odd parity is selected and even parity is selected when "0" is written. The parity check and addition of a parity bit is only valid when "1" has been written to EPR. When "0" has been written to EPR, the parity setting by PMD becomes invalid.

At initial reset, PMD is set to "0" (even parity).

#### TXEN: 00FF49H•D0

Sets the serial interface to the transmitting enable status.

When "1" is written: Transmitting enable When "0" is written: Transmitting disable

Reading: Valid

When "1" is written to TXEN, the serial interface shifts to the transmitting enable status and shifts to the transmitting disable status when "0" is written. Set TXEN to "0" when making the initial settings of the serial interface and similar operations. At initial reset, TXEN is set to "0" (transmitting disable).

#### TXTRG: 00FF49H•D1

Functions as the transmitting start trigger and the operation status indicator (transmitting/stop status).

When "1" is read: During transmitting

When "0" is read: During stop

When "1" is written: Transmitting start

When "0" is written: Invalid

Starts the transmitting when "1" is written to TXTRG after writing the transmitting data. TXTRG can be read as the status. When set to "1", it indicates transmitting operation, and "0" indicates transmitting stop.

At initial reset, TXTRG is set to "0" (during stop).

#### RXEN: 00FF49H•D2

Sets the serial interface to the receiving enable status.

When "1" is written: Receiving enable When "0" is written: Receiving disable

Reading: Valid

When "1" is written to RXEN, the serial interface shifts to the receiving enable status and shifts to the receiving disable status when "0" is written.

Set RXEN to "0" when making the initial settings of the serial interface and similar operations.

At initial reset, RXEN is set to "0" (receiving disable).

#### *RXTRG: 00FF49H•D3*

Functions as the receiving start trigger or preparation for the following data receiving and the operation status indicator (during receiving/during stop).

When "1" is read: During receiving When "0" is read: During stop

When "1" is written: Receiving start/following

data receiving preparation

When "0" is written: Invalid

RXTRG has a slightly different operation in the clock synchronous system and the asynchronous system.

The RXTRG in the clock synchronous system, is used as the trigger for the receiving start. Writes "1" into RXTRG to start receiving at the point where the receive data has been read and the following receive preparation has been done. (In the slave mode,  $\overline{SRDY}$  becomes "0" at the point where "1" has been written into into the RXTRG.)

RXTRG is used in the asynchronous system for preparation of the following data receiving. Reads the received data located in the received data buffer and writes "1" into RXTRG to inform that the received data buffer has shifted to empty. When "1" has not been written to RXTRG, the overrun error flag OER is set to "1" at the point where the following receiving has been completed. (When the receiving has been completed between the operation to read the received data and the operation to write "1" into RXTRG, an overrun error occurs.)

In addition, RXTRG can be read as the status. In either clock synchronous mode or asynchronous mode, when RXTRG is set to "1", it indicates receiving operation and when set to "0", it indicates that receiving has stopped.

At initial reset, RXTRG is set to "0" (during stop).

#### TRXD0-TRXD7: 00FF4AH

#### **During transmitting**

Write the transmitting data into the transmit shift register.

When "1" is written: HIGH level When "0" is written: LOW level

Write the transmitting data prior to starting transmitting.

In the case of continuous transmitting, wait for the transmitting complete interrupt, then write the data. The TRXD7 becomes invalid for the asynchronous 7-bit mode.

Converted serial data for which the bits set at "1" as HIGH (VDD) level and for which the bits set at "0" as LOW (VSS) level are output from the SOUT terminal.

#### **During receiving**

Read the received data.

When "1" is read: HIGH level When "0" is read: LOW level

The data from the received data buffer can be read out. Since the sift register is provided separately from this buffer, reading can be done during the receive operation in the asynchronous mode. (The buffer function is not used in the clock synchronous mode.) Read the data after waiting for the receiving complete interrupt.

When performing parity check in the asynchronous 7-bit mode, "0" is loaded into the 8th bit (TRXD7) that corresponds to the parity bit.

The serial data input from the SIN terminal is level converted, making the HIGH (VDD) level bit "1" and the LOW (Vss) level bit "0" and is then loaded into this buffer.

At initial reset, the buffer content is undefined.

#### OER: 00FF49H•D4

Indicates the generation of an overrun error.

When "1" is read: Error
When "0" is read: No error
When "1" is written: Reset to "0"
When "0" is written: Invalid

OER is an error flag that indicates the generation of an overrun error and becomes "1" when an error has been generated.

An overrun error is generated when the receiving of data has been completed prior to the writing of "1" to RXTRG in the asynchronous mode.

OER is reset to "0" by writing "1".

At initial reset and when RXEN is "0", OER is set to "0" (no error).

#### PER: 00FF49H•D5

Indicates the generation of a parity error.

When "1" is read: Error
When "0" is read: No error
When "1" is written: Reset to "0"
When "0" is written: Invalid

PER is an error flag that indicates the generation of a parity error and becomes "1" when an error has been generated.

When a parity check is performed in the asynchronous mode, if data that does not match the parity is received, a parity error is generated.

PER is reset to "0" by writing "1".

At initial reset and when RXEN is "0", PER is set to "0" (no error).

#### FER: 00FF49H•D6

Indicates the generation of a framing error.

When "1" is read: Error
When "0" is read: No error
When "1" is written: Reset to "0"
When "0" is written: Invalid

FER is an error flag that indicates the generation of a framing error and becomes "1" when an error has been generated.

When the stop bit for the receiving of the asynchronous mode has become "0", a framing error is generated.

FER is reset to "0" by writing "1".

At initial reset and when RXEN is "0", FER is set to "0" (no error).

#### PSIF0, PSIF1: 00FF20H•D4, D5

Sets the priority level of the serial interface interrupt. The two bits PSIF0 and PSIF1 are the interrupt priority register corresponding to the serial interface interrupt. Table 5.8.9.4 shows the interrupt priority level which can be set by this register.

Table 5.8.9.4 Interrupt priority level settings

PSIF1	PSIF0	Interrupt priority level
1	1	Level 3 (IRQ3)
1	0	Level 2 (TRQ2)
0	1	Level 1 (IRQ1)
0	0	Level 0 (None)

At initial reset, this register is set to "0" (level 0).

#### ESTRA, ESREC, ESERR: 00FF23H•D0, D1, D2

Enables or disables the generation of an interrupt for the CPU.

When "1" is written: Interrupt enabled When "0" is written: Interrupt disabled

Reading: Valid

ESTRA, ESREC and ESERR are interrupt enable registers that respectively correspond to the interrupt factors for transmitting complete, receiving complete and receiving error. Interrupts set to "1" are enabled and interrupts set to "0" are disabled. At initial reset, this register is set to "0" (interrupt disabled).

#### FSTRA, FSREC, FSERR: 00FF25H•D0, D1, D2

Indicates the serial interface interrupt generation status.

When "1" is read: Interrupt factor present
When "0" is read: Interrupt factor not present

When "1" is written: Resets factor flag

When "0" is written: Invalid

FSTRA, FSREC and FSERR are interrupt factor flags that respectively correspond to the interrupts for transmitting complete, receiving complete and receiving error and are set to "1" by generation of each factor.

Transmitting complete interrupt factor is generated at the point where the data transmitting of the shift register has been completed.

Receiving complete interrupt factor is generated at the point where the received data has been transferred into the received data buffer.

Receive error interrupt factor is generated when a parity error, framing error or overrun error has been detected during data receiving.

When set in this manner, if the corresponding interrupt enable register is set to "1" and the corresponding interrupt priority register is set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU.

Regardless of the interrupt enable register and interrupt priority register settings, the interrupt factor flag will be set to "1" by the occurrence of an interrupt generation condition.

To accept the subsequent interrupt after interrupt generation, re-setting of the interrupt flags (set interrupt flag to lower level than the level indicated by the interrupt priority registers, or execute the RETE instruction) and interrupt factor flag reset are necessary. The interrupt factor flag is reset to "0" by writing "1".

At initial reset, this flag is reset to "0".

#### 5.8.10 Programming notes

- Be sure to initialize the serial interface mode in the transmitting/receiving disable status (TXEN = RXEN = "0").
- (2) Do not perform double trigger (writing "1") to TXTRG (RXTRG) when the serial interface is in the transmitting (receiving) operation. Furthermore, do not execute the SLP instruction. (When executing the SLP instruction, set TXEN = RXEN = "0".)
- (3) In the clock synchronous mode, since one clock line (SCLK) is shared for both transmitting and receiving, transmitting and receiving cannot be performed simultaneously. (Half duplex only is possible in clock synchronous mode.)

  Consequently, be sure not to write "1" to RXTRG (TXTRG) when TXTRG (RXTRG) is "1".
- (4) When a parity error or flaming error is generated during receiving in the asynchronous mode, the receiving error interrupt factor flag FSERR is set to "1" prior to the receiving complete interrupt factor flag FSREC for the time indicated in Table 5.8.10.1. Consequently, when an error is generated, you should reset the receiving complete interrupt factor flag FSREC to "0" by providing a wait time in error processing routines and similar routines. When an overrun error is generated, the receiving complete interrupt factor flag FSREC is not set to "1" and a receiving complete interrupt is not generated.

Table 5.8.10.1 Time difference between FSERR and FSREC on error generation

Clock source	Time difference		
fosc3 / n	1/2 cycles of fosc3 / n		
Programmable timer	1 cycle of timer 1 underflow		

(5) When the demultiplied signal of the OSC3 oscillation circuit is made the clock source, it is necessary to turn the OSC3 oscillation ON, prior to using the serial interface.

A time interval of several msec to several 10 msec, from the turning ON of the OSC3 oscillation circuit to until the oscillation stabilizes, is necessary, due to the oscillation element that is used. Consequently, you should allow an adequate waiting time after turning ON of the OSC3 oscillation, before starting transmitting/receiving of serial interface. (The oscillation start time will vary somewhat depending on the oscillator and on the externally attached parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".) At initial reset, the OSC3 oscillation circuit is set to

OFF status.

#### 5.9 Clock Timer

#### 5.9.1 Configuration of clock timer

The S1C88349 has built in a clock timer that uses the OSC1 oscillation circuit as clock source. The clock timer is composed of an 8-bit binary counter that uses the 256 Hz signal dividing fosc1 as its input clock and can read the data of each bit (128–1 Hz) by software.

Normally, this clock timer is used for various timing functions such as clocks.

The configuration of the clock timer is shown in Figure 5.9.1.1.

#### 5.9.2 Interrupt function

The clock timer can generate an interrupt by each of the 32 Hz, 8 Hz, 2 Hz and 1 Hz signals. The configuration of the clock timer interrupt circuit is shown in Figure 5.9.2.1.

Interrupts are generated by respectively setting the corresponding interrupt factor flags FTM32, FTM8, FTM2 and FTM1 at the falling edge of the 32 Hz, 8 Hz, 2 Hz and 1 Hz signals to "1". Interrupt can be prohibited by the setting the interrupt enable registers ETM32, ETM8, ETM2 and ETM1 corresponding to each interrupt factor flag. In addition, a priority level of the clock timer interrupt for the CPU can be optionally set at levels 0 to 3 by the interrupt priority registers PTM0 and PTM1

For details on the above mentioned interrupt control register and the operation following generation of an interrupt, see "5.17 Interrupt and Standby Status".

The exception processing vector addresses for each interrupt factor are respectively set as shown below.

32 Hz interrupt: 00001CH 8 Hz interrupt: 00001EH 2 Hz interrupt: 000020H 1 Hz interrupt: 000022H

Figure 5.9.2.2 shows the timing chart for the clock timer.

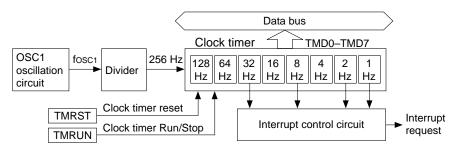


Fig. 5.9.1.1 Configuration of clock timer

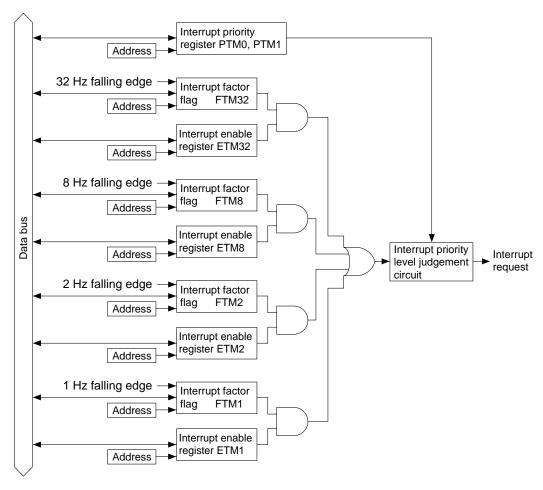


Fig. 5.9.2.1 Configuration of clock timer interrupt circuit

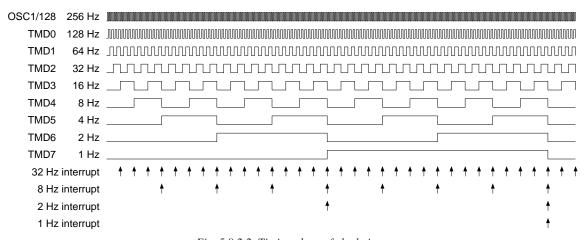


Fig. 5.9.2.2 Timing chart of clock timer

## 5.9.3 Control of clock timer

Table 5.9.3.1 shows the clock timer control bits.

Table 5.9.3.1 Clock timer control bits

Address	Bit	Name	Function			1	0	SR	R/W	Comment
00FF40	D7	_	-		-	-	_		"0" when being read	
	D6	FOUT2	FOUT frequency selection				0	R/W		
	D5	FOUT1	FOUT2 FOUT1 0 0 0 0 0 1	FOUT0 0 1 0	Frequency fosc1 / 1 fosc1 / 2 fosc1 / 4			0	R/W	
	D4	FOUT0	0 1 1 0 1 0 1 1 1 1	1 0 1 0 1	fosc1 / 8 fosc3 / 1 fosc3 / 2 fosc3 / 4 fosc3 / 8			0	R/W	
	D3	FOLITON	FOUT output con	trol		On	Off	0	R/W	
		WDRST	Watchdog timer r			Reset	No operation	_	W	Constantly "0" when
		TMRST	Clock timer reset	CSCI		Reset	No operation		W	being read
		TMRUN	Clock timer Run/	Stop contro	n1	Run	Stop	0	R/W	being read
00FF41		TMD7	Clock timer data	1 Hz	51	Kun	зюр	0	IC W	
001141		TMD6	Clock timer data	2 Hz						
		TMD5	Clock timer data	4 Hz						
		TMD3		8 Hz						
			Clock timer data			High	Low	0	R	
		TMD3	Clock timer data							
		TMD2	Clock timer data							
		TMD1	Clock timer data							
		TMD0	Clock timer data	128 Hz						
00FF20		PK01	K00–K07 interruj	ot priority i	register				R/W	
		PK00	•			PK01 PK0 PSIF1 PSIF				_
		PSIF1 PSIF0	Serial interface in	terrupt prio	ority register	PSW1 PSW PTM1 PTM	V1 PSW0 Priority		R/W	
	D2	PSW1 PSW0	Stopwatch timer i	nterrupt pr	riority register	1 1 Level 3 1 0 Level 2 0 1 Level 1 0 0 Level 0		0	R/W	
		PTM1 PTM0	Clock timer interi	upt priorit	y register			0	R/W	
00FF22	D7	_	_			-	-	_		"0" when being read
	D6	ESW100	Stopwatch timer 10	00 Hz interi	rupt enable register					
	D5	ESW10	Stopwatch timer 10	Hz interru	ıpt enable register					
	D4	ESW1	Stopwatch timer 1	Hz interrup	ot enable register	Intomunt	Intorment			
	D3	ETM32	Clock timer 32 Hz	interrupt er	nable register	Interrupt	Interrupt	0	R/W	
	D2	ETM8	Clock timer 8 Hz i	nterrupt ena	able register	enable	disable			
	D1	ETM2	Clock timer 2 Hz i	nterrupt ena	able register					
	D0	ETM1	Clock timer 1 Hz i	nterrupt ena	able register					
00FF24	D7	_	_			-	-	_		"0" when being read
	D6	FSW100	Stopwatch timer	100 Hz inte	errupt factor flag	(R)	(R)			
		FSW10	Stopwatch timer			Interrupt	No interrupt			
		FSW1	Stopwatch timer			factor is	factor is			
		FTM32	Clock timer 32 H			generated	generated	0	R/W	
		FTM8	Clock timer 8 Hz				1-9			
		FTM2	Clock timer 2 Hz			(W)	(W)			
		FTM1	Clock timer 1 Hz			Reset	No operation			
l	20		C.OCK UIIICI I IIZ	crrupt I						

#### TMD0-TMD7: 00FF41H

The clock timer data can be read out. Each bit of TMD0-TMD7 and frequency correspondence are as follows:

 TMD0:
 128 Hz
 TMD4:
 8 Hz

 TMD1:
 64 Hz
 TMD5:
 4 Hz

 TMD2:
 32 Hz
 TMD6:
 2 Hz

 TMD3:
 16 Hz
 TMD7:
 1 Hz

Since the TMD0-TMD7 is exclusively for reading, the write operation is invalid.

At initial reset, the timer data is set to "00H".

#### TMRST: 00FF40H•D1

Resets the clock timer.

When "1" is written: Clock timer reset When "0" is written: No operation Reading: Always "0"

The clock timer is reset by writing "1" to the TMRST.

When the clock timer is reset in the RUN status, it restarts immediately after resetting. In the case of the STOP status, the reset data "00H" is maintained. No operation results when "0" is written to the TMRST.

Since the TMRST is exclusively for writing, it always becomes "0" during reading.

#### TMRUN: 00FF40H•D0

Controls RUN/STOP of the clock timer.

When "1" is written: RUN When "0" is written: STOP Reading: Valid

The clock timer starts up-counting by writing "1" to the TMRUN and stops by writing "0".

In the STOP status, the count data is maintained until it is reset or set in the next RUN status. Also, when the STOP status changes to the RUN status, the data that was maintained can be used for resuming the count.

At initial reset, the TMRUN is set to "0" (STOP).

#### PTM0, PTM1: 00FF20H•D0, D1

Sets the priority level of the clock timer interrupt. The two bits PTM0 and PTM1 are the interrupt priority register corresponding to the clock timer interrupt. Table 5.9.3.2 shows the interrupt priority level which can be set by this register.

Table 5.9.3.2 Interrupt priority level settings

PTM1	PTM0	Interrupt priority level
1	1	Level 3 (IRQ3)
1	0	Level 2 (IRQ2)
0	1	Level 1 (TRQ1)
0	0	Level 0 (None)

At initial reset, this register is set to "0" (level 0).

#### ETM1, ETM2, ETM8, ETM32: 00FF22H•D0-D3

Enables or disables the generation of an interrupt for the CPU.

When "1" is written: Interrupt enabled When "0" is written: Interrupt disabled

Reading: Valid

The ETM1, ETM2, ETM8 and ETM32 are interrupt enable registers that respectively correspond to the interrupt factors for 1 Hz, 2 Hz, 8 Hz and 32 Hz. Interrupts set to "1" are enabled and interrupts set to "0" are disabled.

At initial reset, this register is set to "0" (interrupt disabled).

#### FTM1, FTM2, FTM8, FTM32: 00FF24H•D0-D3

Indicates the clock timer interrupt generation status.

When "1" is read: Interrupt factor present When "0" is read: Interrupt factor not present

When "1" is written: Resets factor flag

When "0" is written: Invalid

The FTM1, FTM2, FTM8 and FTM32 are interrupt factor flags that respectively correspond to the interrupts for 1 Hz, 2 Hz, 8 Hz and 32 Hz and are set to "1" at the falling edge of each signal. When set in this manner, if the corresponding interrupt enable register is set to "1" and the corresponding interrupt priority register is set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. Regardless of the interrupt enable register and interrupt priority register settings, the interrupt factor flag will be set to "1" by the occurrence of an interrupt generation condition.

To accept the subsequent interrupt after interrupt generation, re-setting of the interrupt flags (set interrupt flag to lower level than the level indicated by the interrupt priority registers, or execute the RETE instruction) and interrupt factor flag reset are necessary. The interrupt factor flag is reset to "0" by writing "1".

At initial reset, this flag is reset to "0".

#### 5.9.4 Programming notes

(1) The clock timer is actually made to RUN/STOP in synchronization with the falling edge of the 256 Hz signal after writing to the TMRUN register. Consequently, when "0" is written to the TMRUN, the timer shifts to STOP status when the counter is incremented "1". The TMRUN maintains "1" for reading until the timer actually shifts to STOP status. Figure 5.9.4.1 shows the timing chart of the RUN/STOP control.

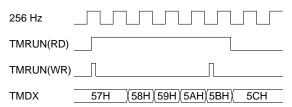


Fig. 5.9.4.1 Timing chart of RUN/STOP control

(2) The SLP instruction is executed when the clock timer is in the RUN status (TMRUN = "1"). The clock timer operation will become unstable when returning from SLEEP status. Therefore, when shifting to SLEEP status, set the clock timer to STOP status (TMRUN = "0") prior to executing the SLP instruction.

#### 5.10 Stopwatch Timer

#### 5.10.1 Configuration of stopwatch timer

The S1C88349 has a built-in 1/100 sec and 1/10 sec stopwatch timer. The stopwatch timer is composed of a 4-bit 2 stage BCD counter (1/100 sec units and 1/10 sec units) that makes the 256 Hz signal that divides the fosc1 the input clock and it can read the count data by software.

Figure 5.10.1.1 shows the configuration of the stopwatch timer.

The stopwatch timer can be used as a timer different from the clock timer and can easily realize stopwatch and other such functions by software.

#### 5.10.2 Count up pattern

The stopwatch timer is respectively composed of the 4-bit BCD counters SWD0-SWD3 and SWD4-SWD7.

Figure 5.10.2.1 shows the count up pattern of the stopwatch timer.

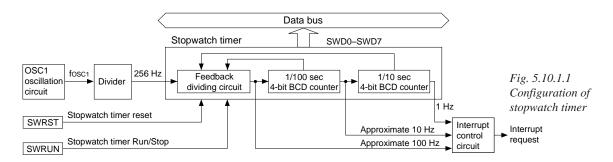
The feedback dividing circuit generates an approximate 100 Hz signal at 2/256 sec and 3/256 sec intervals from a 256 Hz signal divided from fosci.

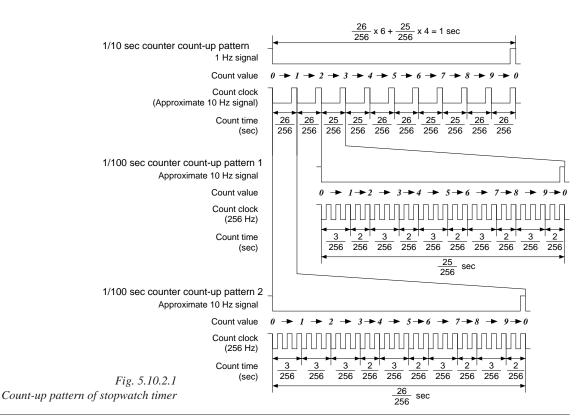
The 1/100 sec counter (SWD0–SWD3) generates an approximate 10 Hz signal at 25/256 sec and 26/256 sec intervals by counting the approximate 100 Hz signal generated by the feedback dividing circuit in 2/256 sec and 3/256 sec intervals. The count-up is made approximately 1/100 sec counting by the 2/256 sec and 3/256 sec intervals.

The 1/10 sec counter (SWD4–SWD7) generates a 1 Hz signal by counting the approximate 10 Hz signal generated by the 1/100 sec counter at 25/256 sec and 26/256 sec intervals in 4:6 ratios.

The count-up is made approximately 1/10 sec counting by 25/256 sec and 26/256 sec intervals.

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### 5.10.3 Interrupt function

The stopwatch timer can generate an interrupt by each of the 100 Hz (approximately 100 Hz), 10 Hz (approximately 10 Hz) and 1 Hz signals. Figure 5.10.3.1 shows the configuration of the stopwatch timer interrupt circuit.

The corresponding factor flags FSW100, FSW10 and FSW1 are respectively set to "1" at the falling edge of the 100 Hz, 10Hz and 1Hz signal and an interrupt is generated. Interrupt can be prohibited by the setting of the interrupt enable registers ESW100, ESW10 and ESW1 corresponding to each interrupt factor flag.

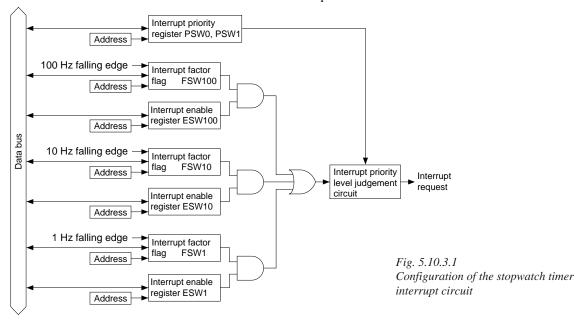
In addition, a priority level of the stopwatch timer interrupt for the CPU can be optionally set at levels 0 to 3 by the interrupt priority registers PSW0 and PSW1.

For details on the above mentioned interrupt control registers and the operation following generation of an interrupt, see "5.17 Interrupt and Standby Status".

The exception processing vector addresses of each interrupt factor are respectively set as shown below.

100 Hz interrupt: 000016H 10 Hz interrupt: 000018H 1 Hz interrupt: 00001AH

Figure 5.10.3.2 shows the timing chart for the stopwatch timer.



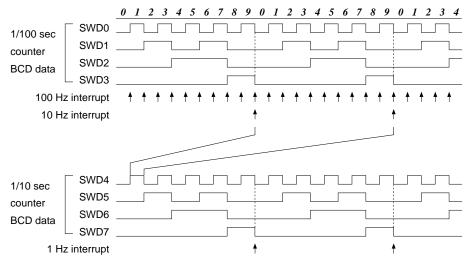


Fig. 5.10.3.2 Stopwatch timer timing chart

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## 5.10.4 Control of stopwatch timer

Table 5.10.4.1 shows the stopwatch timer control bits.

Table 5.10.4.1 Stopwatch timer control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF42	D7	_	-	-	-	_		
	D6	_	_	-	-	-		
	D5	_	_	-	-	-		Constantly "0" when
	D4	_	_	-	-	-		1
	D3	_	_	-	-	-		being read
	D2	_	_	-	-	-		
	D1	SWRST	Stopwatch timer reset	Reset	No operation	-	W	
	D0	SWRUN	Stopwatch timer Run/Stop control	Run	Stop	0	R/W	
00FF43	D7	SWD7	Stopwatch timer data					
	D6	SWD6						
	D5	SWD5	BCD (1/10 sec)					
	D4	SWD4				0	R	
	D3	SWD3	Stopwatch timer data			U	K	
	D2	SWD2						
	D1	SWD1	BCD (1/100 sec)					
	D0	SWD0						
00FF20	D7	PK01	K00 K07 :			0 0 0		
	D6	PK00	K00–K07 interrupt priority register	egister PK01 PK0		0	R/W	
	D5 PSIF1			PSIF1 PSIF0 PSW1 PSW0 Priority		_	R/W	
	D4	PSIF0	Serial interface interrupt priority register	PTM1 PTM		0	K/W	
	D3 PSW1		C44-1-4:i-4	1 1 Level 3	0	R/W		
	D2	PSW0	Stopwatch timer interrupt priority register	1 0 0 1	0 Level 2 1 Level 1 0 Level 0		K/W	
ĺ .	D1	PTM1	Clock times into my to micrity accietor	0 0			R/W	
	D0	PTM0	Clock timer interrupt priority register	<u></u>		0	K/W	
00FF22	D7	_	-	-	-	_		"0" when being read
	D6	ESW100	Stopwatch timer 100 Hz interrupt enable register					
	D5	ESW10	Stopwatch timer 10 Hz interrupt enable register					
	D4	ESW1	Stopwatch timer 1 Hz interrupt enable register	Interrupt	Intonuont			
	D3	ETM32	Clock timer 32 Hz interrupt enable register	•	Interrupt disable	0	R/W	
	D2	ETM8	Clock timer 8 Hz interrupt enable register	enable	disable			
	D1	ETM2	Clock timer 2 Hz interrupt enable register					
	D0	ETM1	Clock timer 1 Hz interrupt enable register					
00FF24	D7	_	-	-	_	_		"0" when being read
	D6	FSW100	Stopwatch timer 100 Hz interrupt factor flag	(R)	(R)			
	D5	FSW10	Stopwatch timer 10 Hz interrupt factor flag	Interrupt	No interrupt			
	D4	FSW1	Stopwatch timer 1 Hz interrupt factor flag	factor is	factor is			
	D3	FTM32	Clock timer 32 Hz interrupt factor flag	generated	generated	0	R/W	
	D2	FTM8	Clock timer 8 Hz interrupt factor flag	(17)	(NE)			
	D1	FTM2	Clock timer 2 Hz interrupt factor flag	(W)	(W)			
	D0	FTM1	Clock timer 1 Hz interrupt factor flag	Reset	No operation			

#### SWD0-SWD7: 00FF43H

The stopwatch timer data can be read out. Higher and lower nibbles and BCD digit correspondence are as follows:

SWD0-SWD3: BCD (1/100 sec) SWD4-SWD7: BCD (1/10 sec)

Since SWD0–SWD7 are exclusively for reading, the write operation is invalid.

At initial reset, the timer data is set to "00H".

#### SWRST: 00FF42H•D1

Resets the stopwatch timer.

When "1" is written: Stopwatch timer reset When "0" is written: No operation

Reading: Always "0"

The stopwatch timer is reset by writing "1" to the SWRST. When the stopwatch timer is reset in the RUN status, it restarts immediately after resetting. In the case of the STOP status, the reset data "00H" is maintained.

No operation results when "0" is written to the SWRST.

Since the SWRST is exclusively for writing, it always becomes "0" during reading.

#### SWRUN: 00FF42H•D0

Controls RUN/STOP of the stopwatch timer.

When "1" is written: RUN When "0" is written: STOP Reading: Valid

The stopwatch timer starts up-counting by writing "1" to the SWRUN and stops by writing "0". In the STOP status, the timer data is maintained until it is reset or set in the next RUN status. Also, when the STOP status changes to the RUN status, the data that was maintained can be used for resuming the count.

At initial reset, the SWRUN is set at "0" (STOP).

#### PSW0, PSW1: 00FF20H•D2, D3

Sets the priority level of the stopwatch timer interrupt.

The two bits PSW0 and PSW1 are the interrupt priority register corresponding to the stopwatch timer interrupt. Table 5.10.4.2 shows the interrupt priority level which can be set by this register.

*Table 5.10.4.2 Interrupt priority level settings* 

	1 1	, 0
PSW1	PSW0	Interrupt priority level
1	1	Level 3 (IRQ3)
1	0	Level 2 (TRQ2)
0	1	Level 1 (TRQ1)
0	0	Level 0 (None)

At initial reset, this register is set to "0" (level 0).

#### ESW1, ESW10, ESW100: 00FF22H•D4, D5, D6

Enables or disables the generation of an interrupt for the CPU.

When "1" is written: Interrupt enabled When "0" is written: Interrupt disabled

Reading: Valid

The ESW1, ESW10 and ESW100 are interrupt enable registers that respectively correspond to the interrupt factors for 1 Hz, 10 Hz and 100 Hz. Interrupts set to "1" are enabled and interrupts set to "0" are disabled.

At initial reset, this register is set to "0" (interrupt disabled).

#### FSW1, FSW10, FSW100: 00FF24H•D4, D5, D6

Indicates the stopwatch timer interrupt generation status.

When "1" is read: Interrupt factor present When "0" is read: Interrupt factor not present

When "1" is written: Resets factor flag

When "0" is written: Invalid

The FSW1, FSW10 and FSW100 are interrupt factor flags that respectively correspond to the interrupts for 1 Hz, 10 Hz and 100 Hz and are set to "1" in synchronization with the falling edge of each signal. When set in this manner, if the corresponding interrupt enable register is set to "1" and the corresponding interrupt priority register is set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. Regardless of the interrupt enable register and interrupt priority register settings, the interrupt factor flag will be set to "1" by the occurrence of an interrupt generation condition.

To accept the subsequent interrupt after interrupt generation, re-setting of the interrupt flags (set interrupt flag to lower level than the level indicated by the interrupt priority registers, or execute the RETE instruction) and interrupt factor flag reset are necessary. The interrupt factor flag is reset to "0" by writing "1".

At initial reset, this flag is reset to "0".

#### 5.10.5 Programming notes

(1) The stopwatch timer is actually made to RUN/STOP in synchronization with the falling edge of the 256 Hz signal after writing to the SWRUN register. Consequently, when "0" is written to the SWRUN, the timer shifts to STOP status when the counter is incremented "1". The SWRUN maintains "1" for reading until the timer actually shifts to STOP status. Figure 5.10.5.1 shows the timing chart of the RUN/STOP control.

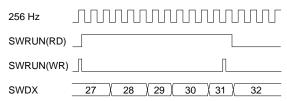


Fig. 5.10.5.1 Timing chart of RUN/STOP control

(2) The SLP instruction is executed when the stopwatch timer is in the RUN status (SWRUN = "1"). The stopwatch timer operation will become unstable when returning from SLEEP status. Therefore, when shifting to SLEEP status, set the clock timer to STOP status (SWRUN = "0") prior to executing the SLP instruction.

## 5.11 Programmable Timer

#### 5.11.1 Configuration of programmable timer

The S1C88349 has two built-in 8-bit programmable timer systems (timer 0 and timer 1).

Timer 0 and timer 1 are composed of 8-bit presettable down counters and they can be used as  $8\text{-bit} \times 2$  channels or  $16\text{-bit} \times 1$  channel programmable timer. They also have an event counter function and a pulse width measurement function using the K10 input port terminal.

Figure 5.11.1.1 shows the configuration of the programmable timer.

Programmable setting of the transfer rate is possible, due to the fact that the programmable timer underflow signal can be used as a synchronous clock for the serial interface.

The underflow divided by 1/2 signal can also be output externally from the R27 output port terminal.

# 5.11.2 Count operation and setting basic mode

Here we will explain the basic operation and setting of the programmable timer.

#### ■ Setting of initial value and counting down The timers 0 and 1 each have a down counter and reload data register.

The reload data registers RLD00–RLD07 (timer 0) and RLD10–RLD17 (timer 1) are registers that set the initial value of the counter.

By writing "1" to the preset control bit PSET0 (timer 0) or PSET1 (timer 1), the down counter loads the initial value set in the reload register RLD.

Therefore, down-counting is executed from the stored initial value according to the input clock.

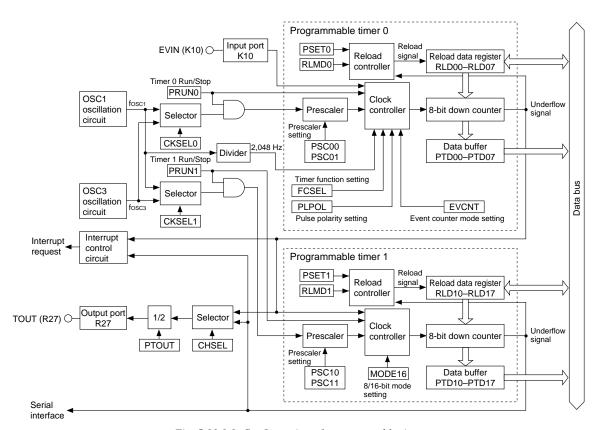


Fig. 5.11.1.1 Configuration of programmable timer

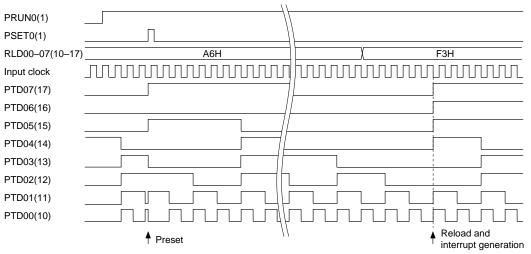


Fig. 5.11.2.1 Basic operation timing of the counter

The registers PRUN0 (timer 0) and PRUN1 (timer 1) are provided to control the RUN/STOP for timers 0 and 1.

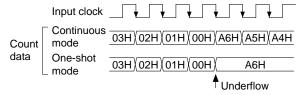
After the reload data has been preset into the counter, down-counting is begun by writing "1" to this register. When "0" is written, the clock input is prohibited and the count stops.

The control of this RUN/STOP has no affect on the counter data. The counter data is maintained even during the stoppage of the counter and it can start the count, continuing from that data.

The reading of the counter data can be done through the data buffers PTD00–PTD07 (timer 0) and PTD10–PTD17 (timer 1) with optional timing. When the down-counting has progressed and an underflow is generated, the counter reloads the initial value set in the reload data register. This underflow signal controls an interrupt generation, pulse (TOUT signal) output and serial interface clocking, in addition to reloading the counter.

#### ■ Continuous/one-shot mode setting

By writing "1" to the continuous/one-shot mode selection registers CONT0 (timer 0) and CONT1 (timer 1), the programmable timer is set to the continuous mode. In the continuous mode, the initial counter value is automatically loaded when an underflow is generated, and counting is continued. This mode is suitable when programmable intervals are necessary (such as an interrupt and a synchronous clock for the serial interface). On the other hand, when writing "0" to the registers CONT0 (timer 0) and CONT1 (timer 1), the programmable timer is set to the one-shot mode. The counter loads an initial value and stops when an underflow is generated. At this time, the RUN/ STOP control register PRUN0 (timer 0) and PRUN1 (timer 1) are automatically reset to "0". After the counter stops, a one-shot count can be performed once again by writing "1" to registers PRUN0 (timer 0) and PRUN1 (timer 1). This mode is suitable for single time measurement, for example.



When "A6H" is set into reload data register RLD.

Fig. 5.11.2.2 Continuous mode and one-shot mode

#### ■ 8/16-bit mode setting

By writing "0" to the 8/16-bit mode selection register MODE16, timer 0 and timer 1 are set as independent timers in 8-bit  $\times$  2 channels. In this mode, timer 0 and timer 1 can be controlled individually and each of them operates independently.

On the other hand, when writing "1" to the register MODE16, timer 0 and timer1 are set as 1 channel 16-bit timer. This is done by setting timer 0 to the lower 8 bits, and timer 1 to the upper 8 bits. The timer is controlled by timer 0's registers. In this case, the control registers for timer 1 are invalid. (PRUN1 is fixed at "0".)

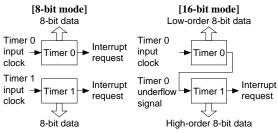


Fig. 5.11.2.3 8/16-bit mode setting and counter configuration

#### 5.11.3 Setting of input clock

Prescalers have been provided for timers 0 and 1. The prescalers generate the input clock for each by dividing the source clock signal from the OSC1 or OSC3 oscillation circuit.

The source clock and the dividing ratio of the prescaler can be selected individually for timer 0 and timer 1 in software.

The input clocks are set by the below sequence.

#### (1) Selection of source clock

Select the source clock (OSC1 or OSC3) for each prescaler. This is done with the source clock selection registers CKSEL0 (timer 0) and CKSEL1 (timer 1): when "0" is written, OSC1 is selected and when "1" is written, OSC3 is selected. When the 16-bit mode is selected, the source clock is selected by register CKSEL0, and the register CKSEL1 setting becomes invalid. When the OSC3 oscillation circuit is made the clock source, it is necessary to turn the OSC3 oscillation ON, prior to using the programmable timer.

From the time the OSC3 oscillation circuit is turning ON until oscillation stabilizes, an interval of several msec to several 10 msec is necessary. Consequently, you should allow an adequate waiting time after turning the OSC3 oscillation circuit ON before starting the count of the programmable timer. (The oscillation start time will vary somewhat depending on the oscillator and on external parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)

At initial reset, OSC3 oscillation circuit is set to OFF status.

#### (2) Selection of prescaler dividing ratio

Select the dividing ratio of each prescaler from among 4 types. This selection is done by the prescaler dividing ratio selection registers PSC00/PSC01 (timer 0) and PSC10/PSC11 (timer 1). Setting value and dividing ratio correspondence are shown in Table 5.11.3.1.

Table 5.11.3.1 Selection of prescaler dividing ratio

PSC11	PSC10	Prescaler dividing ratio				
PSC01	PSC00	Frescaler dividing ratio				
1	1	Source clock / 64				
1	0	Source clock / 16				
0	1	Source clock / 4				
0	0	Source clock / 1				

By writing "1" to the register PRUN0 (timer 0) and PRUN1 (timer 1), the source clock is input to the prescaler. Therefore, the clock with selected dividing ratio is input to the timer and the timer starts counting down.

When the 16-bit mode has been selected, the dividing ratio for the source clock is selected by register PSC00/PSC01 and the setting of register PSC10/PSC11 becomes invalid.

#### 5.11.4 Timer mode

The timer mode counts down using the prescaler output as an input clock. In this mode, the programmable timer operates as a timer that obtains fixed cycles using the OSC1 or OSC3 oscillation circuit as a clock source.

See "5.11.2 Count operation and basic mode setting" for basic operation and control, and "5.11.3 Setting input clock" for the clock source and setting of the prescaler.

#### 5.11.5 Event counter mode

Timer 0 includes an even counter function that counts by inputting an external clock (EVIN) to input port K10. This function is selected by writing "1" to the timer 0 counter mode selection register EVCNT.

When the event counter mode is selected, timer 0 operates as an event counter and timer 1 operates as a normal timer in 8-bit mode. In the 16-bit mode, timer 0 and timer 1 operate as 1 channel 16-bit event counter. In the event counter mode, since the timer 0 is clocked externally, the settings of registers PSC00/PSC01 become invalid.

Count down timing can be controlled by either the falling edge or rising edge selected by the timer 0 pulse polarity selection register PLPOL. When "0" is written to the register PLPOL, the falling edge is selected, and when "1" is written, the rising edge is selected. The timing is shown in Figure 5.11.5.1.

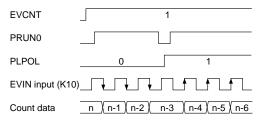


Fig. 5.11.5.1 Timing chart for event counter mode

The event counter also includes a noise rejecter to eliminate noise such as chattering for the external clock (EVIN). This function is selected by writing "1" to the timer 0 function selection register FCSEL.

For a reliable count when "with noise rejecter" is selected, you must allow 0.98 msec or more pulse width for both LOW and HIGH levels. (The noise rejecter allows clocking counter at the second falling edge of the internal 2,048 Hz signal after changing the input level of the K10 input port terminal. Consequently, the pulse width that can reliably be rejected is 0.48 msec.)

Figure 5.11.5.2 shows the count down timing with the noise rejecter selected.

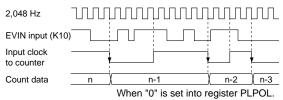


Fig. 5.11.5.2 Count down timing with noise rejecter

The event counter mode is the same as the timer mode except that the clock is external (EVIN). See "5.11.2 Count operation and setting basic mode" for the basic operation and control.

#### 5.11.6 Pulse width measurement timer mode

Timer 0 includes a pulse width measurement function that measures the width of the input signal to the K10 input port terminal. This function is selected by writing "1" to the timer function selection register FCSEL when in the timer mode (EVCNT = "0"). When the pulse width measurement mode is selected, timer 0 operates as an pulse width measurement and timer 1 operates as a normal timer in 8-bit mode. In the 16-bit mode, timer 0 and timer 1 operate as 1 channel 16-bit pulse width measurement. The level of the input signal (EVIN) for measurement can be changed either a LOW or HIGH level by the timer 0 pulse polarity selection register PLPOL. When "0" is written to register PLPOL, a LOW level width is measured and when "1" is written, a HIGH level width is measured. The timing is shown in Figure 5.11.6.1.

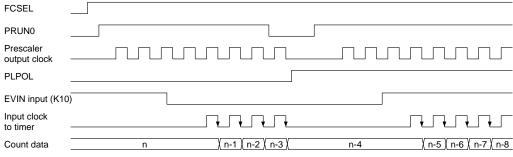


Fig. 5.11.6.1 Timing chart for pulse width measurement timer mode

The pulse width measurement timer mode is the same as the timer mode except that the input clock is controlled by the level of the signal (EVIN) input to the K10 input port terminal.

See "5.11.2 Count operation and setting basic mode" for the basic operation and control.

#### 5.11.7 Interrupt function

The programmable timer can generate an interrupt due to an underflow signal of timer 0 and timer 1. Figure 5.11.7.1 shows the configuration of the programmable timer interrupt circuit.

The respectively corresponding interrupt factor flags FPT0 and FPT1 are set to "1" and an interrupt is generated by an underflow signal of timers 1 and 0. Interrupt can also be prohibited by the setting of the interrupt enable registers EPT0 and EPT1 corresponding to each interrupt flag.

In addition, a priority level of the programmable timer interrupt for the CPU can be optionally set at levels 0 to 3 by the interrupt priority registers PPT0 and PPT1.

For details on the above mentioned interrupt control registers and the operation following generation of an interrupt, see "5.17 Interrupt and Standby Status".

The exception processing vector addresses of each interrupt factor are respectively set as shown below.

Programmable timer 1 interrupt: 000006H Programmable timer 0 interrupt: 000008H

When the 16-bit mode is selected, the interrupt factor flag FPT0 is not set to "1" and a timer 0 interrupt cannot be generated. (In the 16-bit mode, the interrupt factor flag FPT1 is set to "1" by an underflow of the 16-bit counter.)

#### 5.11.8 Setting of TOUT output

The programmable timer can generate a TOUT signal due to an underflow of timer 0 or timer 1. A TOUT signal is the above mentioned underflow divided by 1/2. The timer underflow which is to be used can be selected by the TOUT output channel selection register CHSEL. When writing "0" to register CHSEL, timer 0 is selected and when "1" is written, timer 1 is selected. However, in the 16-bit mode, it is fixed in timer 1 (underflow of the 16-bit timer) and the setting of register CHSEL becomes invalid.

Figure 5.11.8.1 shows the TOUT signal waveform when channel switching.

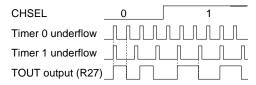


Fig. 5.11.8.1 TOUT signal waveform at channel change

The TOUT signal can be output from the R27 output port terminal and the programmable clock can be supplied to an external device.

The configuration of the output port R27 is shown in Figure 5.11.8.2.

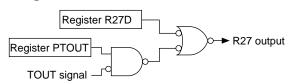
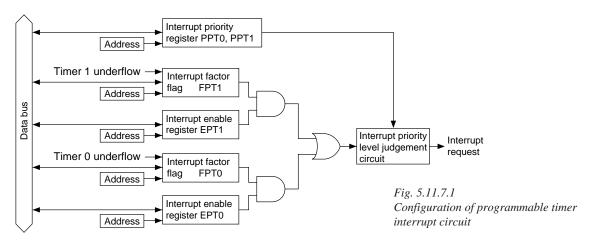


Fig. 5.11.8.2 Configuration of R27



The output control of the TOUT signal is done by register PTOUT. When "1" is set to the PTOUT, the TOUT signal is output from the R27 output port and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set in the data register R27D.

Since the TOUT signal is generated asynchronously from the register PTOUT, when the signal is turned ON or OFF by the register setting, a hazard of a 1/2 cycle or less is generated.

Figure 5.11.8.3 shows the output waveform of TOUT signal.



Fig. 5.11.8.3 Output waveform of the TOUT signal

# 5.11.9 Transmission rate setting of serial interface

The underflow signal of the timer 1 can be used to clock the serial interface.

The transmission rate setting in this case is made in registers PSC1X and RLD1X, and is used to set the count mode to the reload count mode (RLMD1 = "1").

Since the underflow signal of the timer 1 is divided by 1/32 in the serial interface, the value set in register RLD1X which corresponds to the transmission rate is shown in the following expression:

# RLD1X = fosc / (32\*bps\*4PSC1X) - 1

fosc: Oscillation frequency (OSC1/OSC3)

bps: Transmission rate

PSC1X: Setting value to the register PSC1X (0-3)

(00H can be set to RLD1X)

Table 5.11.9.1 shows an example of the transmission rate setting when the OSC3 oscillation circuit is used as a clock source.

Table 5.11.9.1 Example of transmission rate setting

Transfer rate	OSC3 os	scillation fr	equency /	Programm	nable time	settings
Transfer rate	fosc3 = 3	.072 MHz	fosc3 = 4	.608 MHz	fosc3 = 4.	9152 MHz
(bps)	PSC1X	RLD1X	PSC1X	RLD1X	PSC1X	RLD1X
9,600	0 (1/1)	09H	0 (1/1)	0EH	0 (1/1)	0FH
4,800	0 (1/1)	13H	0 (1/1)	1DH	0 (1/1)	1FH
2,400	0 (1/1)	27H	0 (1/1)	3BH	0 (1/1)	3FH
1,200	0 (1/1)	4FH	0 (1/1)	77H	0 (1/1)	7FH
600	0 (1/1)	9FH	0 (1/1)	EFH	0 (1/1)	FFH
300	1 (1/4)	4FH	1 (1/4)	77H	1 (1/4)	7FH
150	1 (1/4)	9FH	1 (1/4)	EFH	1 (1/4)	FFH

# 5.11.10 Control of programmable timer

Table 5.11.10.1 shows the programmable timer control bits.

Table 5.11.10.1(a) Programmable timer control bits

Address	Bit	Name		nction	1	0	SR	R/W	Comment
00FF30	D7	_	_		_	-	_		
	D6	_	_		_	-	_		Constantly "0" when
	D5	_	_		_	-	_		being read
<b> </b>	D4	MODE16	8/16-bit mode selection	on	16-bit x 1	8-bit x 2	0	R/W	
	D3	CHSEL	TOUT output channel	l selection	Timer 1	Timer 0	0	R/W	
	D2	PTOUT	TOUT output control		On	Off	0	R/W	
	D1	CKSEL1	Prescaler 1 source clo	ock selection	fosc3	foscı	0	R/W	
	D0	CKSEL0	Prescaler 0 source clo	ock selection	fosc3	foscı	0	R/W	
00FF31	D7	EVCNT	Timer 0 counter mode	e selection	Event counter	Timer	0	R/W	
	D6	FCSEL	Timer 0	In timer mode	Pulse width	Normal	0	R/W	
			function selection		measurement	mode			
				In event counter mode	With	Without			
					noise rejector	noise rejector			
	D5	PLPOL	Timer 0	Down count timing	Rising edge	Falling edge	0	R/W	
			pulse polarity	in event counter mode		of K10 input			
			selection	In pulse width	High level	Low level measurement			
				measurement mode		for K10 input			
	D4	PSC01	Timer 0 prescaler div	iding ratio selection			0	R/W	
			PSC01 PSC00	Prescaler dividing ratio					
			1 1	Source clock / 64					
	D3	PSC00	1 0	Source clock / 16			0	R/W	
			0 1	Source clock / 4					
			0 0	Source clock / 1					
	D2	CONT0	Timer 0 continuous/o	ne-shot mode selection	Continuous	One-shot	0	R/W	
		PSET0	Timer 0 preset		Preset	No operation	_	W	"0" when being read
	D0	PRUN0	Timer 0 Run/Stop cor	ntrol	Run	Stop	0	R/W	
00FF32	D7	_	_		-	-	_		Constantly "0" when
	D6	_	_		-	-	_		being read
	D5	_	_		-	-	_		being read
	D4	PSC11	Timer 1 prescaler div	iding ratio selection			0	R/W	
			PSC11 PSC10	Prescaler dividing ratio					
			1 1	Source clock / 64					
	D3	PSC10	1 0	Source clock / 16			0	R/W	
			0 1	Source clock / 4					
			0 0	Source clock / 1					
				ne-shot mode selection	Continuous	One-shot	0	R/W	
		PSET1	Timer 1 preset		Preset	No operation	_	W	"0" when being read
		PRUN1	Timer 1 Run/Stop cor		Run	Stop	0	R/W	
00FF33		RLD07	Timer 0 reload data D						
		RLD06	Timer 0 reload data D						
		RLD05	Timer 0 reload data D						
		RLD04	Timer 0 reload data D		High	Low	1	R/W	
		RLD03	Timer 0 reload data D						
		RLD02	Timer 0 reload data D						
<b> </b>	D1	RLD01	Timer 0 reload data D	01					
, <b>i</b>		RLD00	Timer 0 reload data D				1		

100

Table 5.11.10.1(b) Programmable timer control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF34	D7	RLD17	Timer 1 reload data D7 (MSB)					
	D6	RLD16	Timer 1 reload data D6					
	D5	RLD15	Timer 1 reload data D5					
	D4	RLD14	Timer 1 reload data D4		_		D 444	
	D3	RLD13	Timer 1 reload data D3	High	Low	1	R/W	
	D2	RLD12	Timer 1 reload data D2					
	D1	RLD11	Timer 1 reload data D1					
	D0	RLD10	Timer 1 reload data D0 (LSB)					
00FF35	D7	PTD07	Timer 0 counter data D7 (MSB)					
	D6	PTD06	Timer 0 counter data D6					
	D5	PTD05	Timer 0 counter data D5					
	D4	PTD04	Timer 0 counter data D4	*** 1		1		
	D3	PTD03	Timer 0 counter data D3	High	Low	1	R	
	D2	PTD02	Timer 0 counter data D2					
	D1	PTD01	Timer 0 counter data D1					
	D0	PTD00	Timer 0 counter data D0 (LSB)					
00FF36	D7	PTD17	Timer 1 counter data D7 (MSB)					
	D6	PTD16	Timer 1 counter data D6					
	D5	PTD15	Timer 1 counter data D5					
	D4	PTD14	Timer 1 counter data D4		_			
	D3	PTD13	Timer 1 counter data D3	High	Low	1	R	
	D2	PTD12	Timer 1 counter data D2					
	D1	PTD11	Timer 1 counter data D1					
	D0	PTD10	Timer 1 counter data D0 (LSB)					
00FF21	D7	_	_	-	-	_		
	D6	_	_	-	_	_		Constantly "0" when
	D5	_	_	-	_	_		being read
	D4	_	_	-	_	_		
	D3	PPT1	D 11.2 14 4 1 14	PPT1 PPT		0	D/W	
	D2	PPT0	Programmable timer interrupt priority register	$\frac{PK11}{1}$ $\frac{PK1}{1}$	0 level 3	0	R/W	
<u>'</u>	D1	PK11	W10 1W11' / ' ' '	1 0	Level 2	0	D/W	
	D0	PK10	K10 and K11 interrupt priority register	$\begin{array}{ccc} 0 & 1 \\ 0 & 0 \end{array}$	Level 1 Level 0	0	R/W	
00FF23	D7	EPT1	Programmable timer 1 interrupt enable register					
	D6	EPT0	Programmable timer 0 interrupt enable register					
]	D5	EK1	K10 and K11 interrupt enable register					
	D4	EK0H	K04–K07 interrupt enable register	Interrupt	Interrupt	0	D/W	
	D3	EK0L	K00-K03 interrupt enable register	enable	disable	U	R/W	
	D2	ESERR	Serial I/F (error) interrupt enable register					
	D1	ESREC	Serial I/F (receiving) interrupt enable register					
	D0	ESTRA	Serial I/F (transmitting) interrupt enable register					
00FF25	D7	FPT1	Programmable timer 1 interrupt factor flag	(R)	(R)			
	D6	FPT0	Programmable timer 0 interrupt factor flag	Interrupt	No interrupt			
	D5	FK1	K10 and K11 interrupt factor flag	factor is	factor is			
	D4	FK0H	K04–K07 interrupt factor flag	generated	generated	0	D/337	
	D3	FK0L	K00-K03 interrupt factor flag			0	R/W	
	D2	FSERR	Serial I/F (error) interrupt factor flag	(W)	(W)			
	D1	FSREC	Serial I/F (receiving) interrupt factor flag	Reset	No operation			
	D0	FSTRA	Serial I/F (transmitting) interrupt factor flag					

#### MODE16: 00FF30H•D4

Selects the 8/16-bit mode.

When "1" is written: 16 bits  $\times$  1 channel When "0" is written: 8 bits  $\times$  2 channels

Reading: Valid

Select whether timer 0 and timer 1 will be used as 2 channel independent 8-bit timers or as a 1 channel combined 16-bit timer. When "0" is written to MODE16, 8-bit  $\times$  2 channels is selected and when "1" is written, 16-bit  $\times$  1 channel is selected. At initial reset, MODE16 is set to "0" (8-bit  $\times$  2 channels).

# CKSEL0, CKSEL1: 00FF30H•D0, D1

Select the source clock of the prescaler.

When "1" is written: OSC3 clock When "0" is written: OSC1 clock Reading: Valid

Select whether the source clock of prescaler 0 will be set to OSC1 or OSC3. When "0" is written to CKSEL0, OSC1 is selected and when "1" is written, OSC3 is selected.

In the same way, the source clock of prescaler 1 is selected by CKSEL1.

When event counter mode has been selected, the setting of the CKSEL0 becomes invalid. In the same way, the CKSEL1 setting becomes invalid when 16-bit mode has been selected.

At initial reset, this register is set to "0" (OSC1 clock).

# PSC00, PSC01: 00FF31H•D3, D4 PSC10, PSC11: 00FF32H•D3, D4

Select the dividing ratio of the prescaler. Two-bit PSC00 and PSC01 is the prescaler dividing ratio selection registers for timer 0, and the two-bit PSC10 and PSC11 correspond to timer 1. The prescaler dividing ratios that can be set by these registers are shown in Table 5.11.10.2.

Table 5.11.10.2 Selection of prescaler dividing ratio

PSC11	PSC10	Prescaler dividing ratio
PSC01	PSC00	1 rescaler dividing ratio
1	1	Input clock / 64
1	0	Input clock / 16
0	1	Input clock / 4
0	0	Input clock / 1

When event counter mode has been selected, the setting of the PSC00 and PSC01 becomes invalid. In the same way, the PSC10 and PSC11 setting becomes invalid when 16-bit mode has been selected. At initial reset, this register is set to "0" (input clock/1).

# EVCNT: 00FF31H•D7

Selects the counter mode for the timer 0.

When "1" is written: Event counter mode

When "0" is written: Timer mode Reading: Valid

Select whether timer 0 will be used as an event counter or a timer. When "1" is written to EVCNT, the event counter mode is selected and when "0" is written, the timer mode is selected.

At initial reset, EVCNT is set to "0" (timer mode).

#### FCSEL: 00FF31H•D6

Selects the function for each counter mode of timer 0.

#### • In timer mode

When "1" is written: Pulse width measurement

timer mode

When "0" is written: Normal mode

Reading: Valid

In the timer mode, select whether timer 0 will be used as a pulse width measurement timer or a normal timer. When "1" is written to FCSEL, the pulse width measurement mode is selected and the counting is done according to the level of the signal (EVIN) input to the K10 input port terminal. When "0" is written to FCSEL, the normal mode is selected and the counting is not affected by the K10 input port terminal.

#### • In event counter mode

When "1" is written: With noise rejecter When "0" is written: Without noise rejecter

Reading: Valid

In the event counter mode, select whether the noise rejecter for the K10 input port terminal will be selected or not.

When "1" is written to FCSEL, the noise rejecter is selected and counting is done by an external clock (EVIN) with 0.98 msec or more pulse width. (The noise rejecter allows clocking counter at the second falling edge of the internal 2,048 Hz signal after changing the input level of the K10 input port terminal. Consequently, the pulse width that can reliably be rejected is 0.48 msec.)

When "0" is written to FCSEL, the noise rejector is not selected and the counting is done directly by an external clock (EVIN) input to the K10 input port terminal.

At initial reset, FCSEL is set to "0".

#### PLPOL: 00FF31H•D5

Selects the pulse polarity for the K10 input port terminal.

#### • In event counter mode

When "1" is written: Rising edge When "0" is written: Falling edge

Reading: Valid

In the event counter mode, select whether the count timing will be set at the falling edge of the external clock (EVIN) input to the K10 input port terminal or at the rising edge. When "0" is written to PLPOL, the falling edge is selected and when "1" is written, the rising edge is selected.

#### • In pulse width measurement mode

When "1" is written: High level pulse width

measurement

When "0" is written: LOW level pulse width

measurement

Reading: Valid

In the pulse width measurement mode, select whether the LOW level width of the signal (EVIN) input to the K10 input port terminal will be measured or the HIGH level will be measured. When "0" is written to PLPOL, the LOW level width measurement is selected and when "1" is written, the HIGH level width measurement is selected.

In the normal mode (EVCNT = FCSEL = "0"), the setting of PLPOL becomes invalid. At initial reset, PLPOL is set to "0".

#### CONT0, CONT1: 00FF31H•D2, 00FF32H•D2

Select the continuous/one-shot mode.

When "1" is written: Continuous mode When "0" is written: One-shot mode

Reading: Valid

Select whether timer 0 will be used in the continuous mode or in the one-shot mode.

By writing "1" to CONT0, the programmable timer is set to the continuous mode. In the continuous mode, the initial counter value is automatically loaded when an underflow is generated, and counting is continued. On the other hand, when writing "0" to CONT0, the programmable timer is set to the one-shot mode. The counter loads an initial value and stops when an underflow is generated. At this time, PRUN0 is automatically reset to "0".

In the same way, the continuous/one-shot mode for timer 1 is selected by CONT1. (In the one-shot mode for timer 1, PRUN1 is automatically reset to "0" when the counter underflow is generated.) At initial reset, this register is set to "0" (one-shot mode).

# RLD00-RLD07: 00FF33H RLD10-RLD17: 00FF34H

Sets the initial value for the counter.

RLD00-RLD07: Reload data for Timer 0 RLD10-RLD17: Reload data for Timer 1

The reload data set in this register is loaded into the respective counters and is counted down with that as the initial value.

Reload data is loaded to the counter under two conditions, when "1" is written to PSET0 or PSET1 and when the counter underflow automatically loads.

At initial reset, this register is set to "FFH".

## PTD00-PTD07: 00FF35H PTD10-PTD17: 00FF36H

Data of the programmable timer can be read out.

PTD00-PTD07: Timer 0 counter data PTD10-PTD17: Timer 1 counter data

These bits act as a buffer to maintain the counter data during readout, and the data can be read as optional timing. However, in the 16-bit mode, to avoid a read error, (data error when a borrow from timer 0 to timer 1 is generated in the middle of reading PTD00–PTD07 and PTD10–PTD17), PTD10–PTD17 latches the timer 1 counter data according to the reading of PTD00–PTD07. The latched status of PTD10–PTD17 is canceled according to the readout of PTD10–PTD17 or when 0.73–1.22 msec (depends on the readout timing) has elapsed. Therefore, in 16-bit mode, be sure to read the counter data of PTD00–PTD07 and PTD10–PTD17 in order.

Since these bits are exclusively for reading, the write operation is invalid.

At initial reset, these bits are set to "FFH".

#### PSET0, PSET1: 00FF31H•D1, 00FF32H•D1

Presets the reload data to the counter.

When "1" is written: Preset
When "0" is written: No operation
Reading: Always "0"

By writing "1" to PSET0, the reload data in PLD00– PLD07 is preset to the counter of timer 0. When the counter of timer 0 is preset in the RUN status, it restarts immediately after presetting.

In the case of STOP status, the reload data that has been preset is maintained.

No operation results when "0" is written. In the same way, the reload data in PLD10–PLD17 is preset to the counter of timer 1 by PSET1. When the 16-bit mode is selected, writing "1" to PSET1 is invalid.

This bit is exclusively for writing, it always becomes "0" during reading.

#### PRUNO, PRUN1: 00FF31H•D0, 00FF32H•D0

Controls the RUN/STOP of the counter.

When "1" is written: RUN When "0" is written: STOP Reading: Valid

The counter of timer 0 starts down-counting by writing "1" to PRUN0 and stops by writing "0". In the STOP status, the counter data is maintained until it is preset or set in the next RUN status. Also, when the STOP status changes to the RUN status, the data that was maintained can be used for resuming the count.

In the same way, the RUN/STOP of the timer 1 counter is controlled by PRUN1.

When the 16-bit mode is selected, PRUN1 is fixed at "0"

At initial reset and when an underflow is generated in the one-shot mode, this register is set to "0" (STOP).

#### CHSEL: 00FF30H•D3

Selects the channel of the TOUT signal.

When "1" is written: Timer 1 underflow When "0" is written: Timer 0 underflow

Reading: Valid

Select whether the timer 0 underflow will be used for the TOUT signal or the timer 1 underflow will be used. When "0" is written to CHSEL, timer 0 is selected and when "1" is written, timer 1 is selected. When the 16-bit mode has been selected, it is fixed to timer 1 (underflow of the 16-bit timer), and setting of CHSEL becomes invalid.

At initial reset, CHSEL is set to "0" (timer 0 underflow).

#### PTOUT: 00FF30H•D2

Controls the TOUT signal output.

When "1" is written: TOUT signal output
When "0" is written: HIGH level (DC) output

Reading: Valid

PTOUT is the output control register for TOUT signal. When "1" is set, the TOUT signal is output from the output port terminal R27 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R27D. At initial reset, PTOUT is set to "0" (HIGH level output).

# PPT0, PPT1: 00FF21H•D2, D3

Sets the priority level of the programmable timer interrupt.

The two bits PPT0 and PPT1 are the interrupt priority register corresponding to the programmable timer interrupt. Table 5.11.10.3 shows the interrupt priority level which can be set by this register.

Table 5.11.10.3 Interrupt priority level settings

PPT1	PPT0	Interrupt priority level
1	1	Level 3 (IRQ3)
1	0	Level 2 (IRQ2)
0	1	Level 1 (IRQ1)
0	0	Level 0 (None)

At initial reset, this register is set to "0" (level 0).

#### EPT0, EPT1: 00FF23H•D6, D7

Enables or disables the generation of an interrupt for the CPU.

When "1" is written: Interrupt enabled When "0" is written: Interrupt disabled

Reading: Valid

The EPT0 and EPT1 are interrupt enable registers that respectively correspond to the interrupt factors for timer 0 and timer 1. Interrupts set to "1" are enabled and interrupts set to "0" are disabled. When the 16-bit mode is selected, setting of EPT0 becomes invalid.

At initial reset, this register is set to "0" (interrupt disabled).

#### FPT0, FPT1: 00FF25H•D6, D7

Indicates the programmable timer interrupt generation status.

When "1" is read: Interrupt factor present
When "0" is read: Interrupt factor not present

When "1" is written: Resets factor flag

When "0" is written: Invalid

The FPT0 and FPT1 are interrupt factor flags that respectively correspond to the interrupts for timer 0 and timer 1 and are set to "1" in synchronization with the underflow of each counter.

When set in this manner, if the corresponding interrupt enable register is set to "1" and the corresponding interrupt priority register is set to a higher level than the setting of interrupt flags (I0 and I1), an interrupt will be generated to the CPU. Regardless of the interrupt enable register and interrupt priority register settings, the interrupt factor flag will be set to "1" by the occurrence of an interrupt generation condition.

To accept the subsequent interrupt after interrupt generation, re-setting of the interrupt flags (set interrupt flag to lower level than the level indicated by the interrupt priority registers, or execute the RETE instruction) and interrupt factor flag reset are necessary. The interrupt factor flag is reset to "0" by writing "1".

When the 16-bit mode is selected, the interrupt factor flag FPT0 is not set to "1" and a timer 0 interrupt cannot be generated. (In the 16-bit mode, the interrupt factor flag FPT1 is set to "1" by an underflow of the 16-bit counter.)

At initial reset, this flag is reset to "0".

# 5.11.11 Programming notes

(1) The programmable timer is actually made to RUN/STOP in synchronization with the falling edge of the input clock after writing to the PRUN0(1) register. Consequently, when "0" is written to the PRUN0(1), the timer shifts to STOP status when the counter is decremented "1". The PRUN0(1) maintains "1" for reading until the timer actually shifts to STOP status. Figure 5.11.11.1 shows the timing chart of the RUN/STOP control.

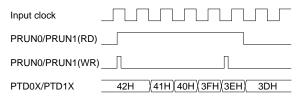


Fig. 5.11.11.1 Timing chart of RUN/STOP control
The event counter mode is excluded from the above note.

- (2) The SLP instruction is executed when the programmable timer is in the RUN status (PRUN0(1) = "1"). The programmable timer operation will become unstable when returning from SLEEP status. Therefore, when shifting to SLEEP status, set the clock timer to STOP status (PRUN0(1) = "0") prior to executing the SLP instruction.
  In the same way, disable the TOUT signal (PTOUT = "0") to avoid an unstable clock output
- (3) Since the TOUT signal is generated asynchronously from the register PTOUT, when the signal is turned ON or OFF by the register setting, a hazard of a 1/2 cycle or less is generated.

to the R27 output port terminal.

(4) When the OSC3 oscillation circuit is made the clock source, it is necessary to turn the OSC3 oscillation ON, prior to using the programmable timer

From the time the OSC3 oscillation circuit is turning ON until oscillation stabilizes, an interval of several msec to several 10 msec is necessary. Consequently, you should allow an adequate waiting time after turning the OSC3 oscillation circuit ON before starting the count of the programmable timer. (The oscillation start time will vary somewhat depending on the oscillator and on external parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)

At initial reset, OSC3 oscillation circuit is set to OFF status.

- (5) When the 16-bit mode has been selected, be sure to read the counter data in the order of PTD00– PTD07 and PTD10–PTD17. Moreover, the time interval between reading PTD00–PTD07 and PTD10–PTD17 should be 0.73 msec or less.
- (6) For the reason below, pay attention to the reload data write timing when changing the interval of the programmable timer interrupts while the programmable timer is running. The programmable timer counts down at the falling edge of the input clock and at the same time it generates an interrupt if the counter underflows. Then it starts loading the reload data to the counter and the counter data is determined at the next rising edge of the input clock (period shown in as ① in the figure).

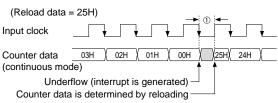


Fig. 5.11.11.2 Reload timing for programmable timer

To avoid improper reloading, do not rewrite the reload data after an interrupt occurs until the counter data is determined including the reloading period ①. Be especially careful when using the OSC1 (low-speed clock) as the clock source of the programmable timer and the CPU is operating with the OSC3 (high-speed clock).

# 5.12 LCD Controller

# 5.12.1 Configuration of LCD controller

The S1C88349 has a built-in dot matrix LCD driver. The S1C88349 allows an LCD panel with a maximum of 1,632 dots (51 segments  $\times$  32 commons). It also has an LCD controller for an external LCD driver (S1D15210 or equivalent).

Figure 5.12.1.1 shows the configuration of the LCD controller and the drive power supply.

Note: Refer to the "S1D15000 Series Technical Manual" for the S1D15210 externaly expanded LCD driver.

# 5.12.2 Mask option

The mask option allows selection of the power source that supplies the LCD drive voltages VD1 to VD5. Either the internal power supply configured with a voltage regulator and a voltage booster or an external power supply can be selected.

```
LCD power source

☐ Internal power supply
☐ External power supply
```

# 5.12.3 Switching drive duty

The S1C88349 supports three types of LCD drive duty settings, 1/8, 1/16 and 1/32, and it can be switched using the LDUTY and DUTY8 registers. When "0" is written to the drive duty selection register LDUTY, 1/32 duty is selected and when "1" is written, 1/16 duty is selected. When "1" is written to DUTY8, the drive duty is fixed at 1/8 and setting of LDUTY becomes invalid.

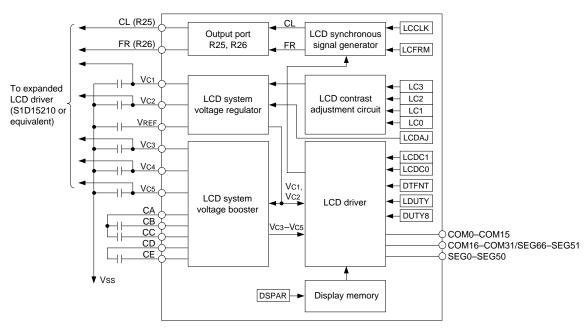


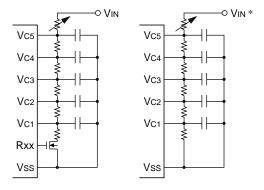
Fig. 5.12.1.1 Configuration of LCD controller and drive power supply

# 5.12.4 LCD power supply

The LCD drive voltages VC1–VC5 can be generated by the internal voltage regulator and voltage booster circuits or can be supplied from an external power supply. The power source is selected by mask option.

The internal power supply generates two types of reference voltage, TYPE A (4.5 V) and TYPE B (5.5 V), and either one can be selected according to the panel characteristics using the LCDAJ register. The internal power supply is designed for a small scale LCD panel and is not suitable for driving a panel that has large size pixels or for driving a large capacity panel using an external expanded LCD driver. In this case, select external power supply and input the regulated voltage from outside of the IC.

Figure 5.12.4.1 shows the circuit examples when using an external power supply.



<sup>\*</sup> Vss level or high impedance when LCD is not driven.

Fig. 5.12.4.1 Circuit examples when using an external power supply

# 5.12.5 LCD driver

The maximum number of dots changes according to the drive duty selection.

When 1/32 duty is selected, the combined common/segment output terminal is switched to the common terminal. An LCD panel with 51 segments × 32 commons (maximum 1,632 dots) can be driven in the S1C88349. When 1/16 duty is selected, the combined common/segment output terminal is switched to the segment terminal. An LCD panel with 67 segments × 16 commons (maximum 1,072 dots) in the S1C88349 can be driven. When 1/8 duty is selected, the combined common/segment output terminal is switched to the segment terminal as when 1/16 duty is selected. An LCD panel with 67 segments × 8 commons (maximum 536 dots) in the S1C88349 can be driven. Furthermore, when 1/ 8 duty is selected, terminals COM8-COM15 become invalid, in that they always output an OFF

Table 5.12.5.1 shows the correspondence between the drive duty and the maximum number of displaying dots. The drive bias is 1/5 (five potentials, VC1–VC5) for any one of the 1/32, 1/16 and 1/8 duties. The respective drive waveforms are shown in Figures 5.12.5.1–5.12.5.3.

Table 5.12.5.1 Correspondence between drive duty and maximum number of displaying dots

	1				3 1 2 0
DUTY8	LDUTY	Duty	Common terminal	Segment terminal	Maximum number of display dots
0	0	1/32	COM0-COM31	SEG0-SEG50	1,632 dots
0	1	1/16	COM0-COM15	SEG0-SEG66	1,072 dots
1	×	1/8	COM0-COM7	SEG0-SEG66	536 dots

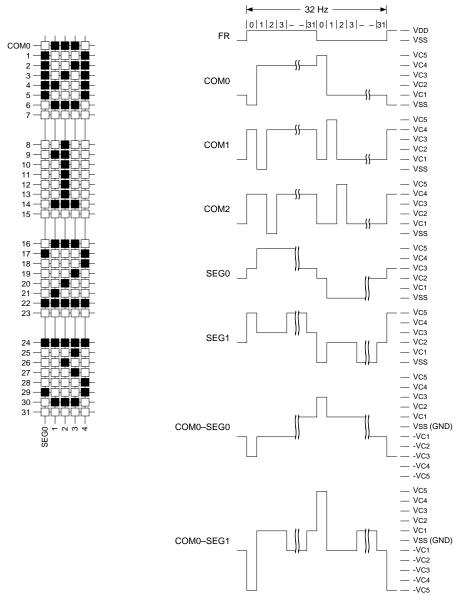


Fig. 5.12.5.1 Drive waveform for 1/32 duty

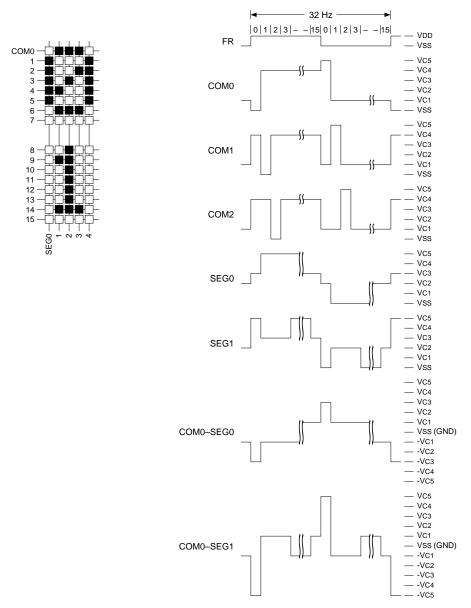
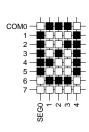


Fig. 5.12.5.2 Drive waveform for 1/16 duty



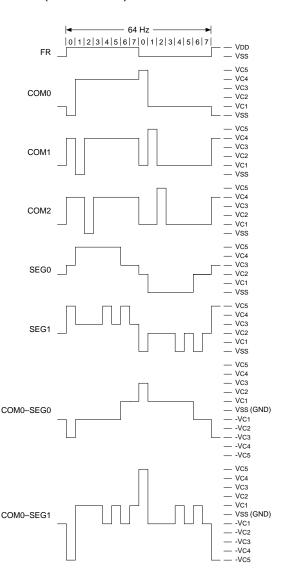


Fig. 5.12.5.3 Drive waveform for 1/8 duty

# 5.12.6 Display memory

The S1C88349 has a built-in 402-byte display memory. The display memory is allocated to address Fx00H-Fx42H (x = 8-DH) and the correspondence between the memory bits and common/segment terminal is changed according to the selection status of the following items.

- (1) Drive duty (1/32, 1/16 or 1/8 duty)
- (2) Dot font  $(5 \times 8 \text{ or } 5 \times 5 \text{ dots})$

When 1/16 or 1/8 duty is selected for drive duty, two-screen memory can be secured, and the two screens can be switched by the display memory area selection register DSPAR. When "0" is written to DSPAR, display area 0 is selected and when "1" is written, display area 1 is selected.

Furthermore, memory allocation for  $5 \times 8$  dots and  $5 \times 5$  dots can be selected in order to easily display  $5 \times 5$ -dot font characters on the LCD panel. This selection can be done by the dot font selection register DTFNT: when "0" is written to DTFNT,  $5 \times 8$  dots is selected and when "1" is written,  $5 \times 5$  dots is selected.

The correspondence between the display memory bits set according to the drive duty and font size, and the common/segment terminals are shown in Figures 5.12.6.1–5.12.6.6.

When "1" is written to the display memory bit corresponding to the dot on the LCD panel, the dot goes ON and when "0" is written, it goes OFF. Since display memory is designed to permit reading/writing, it can be controlled in bit units by logical operation instructions and other means (read, modify and write instructions).

The display area bits which have not been assigned within the 402-byte display memory can be used as general purpose RAM with read/write capabilities. Even when external memory has expanded into the display memory area, this area is not released to external memory. Access to this area is always via display memory.

Address/Data bit	0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2	3 4 5 6 7 8 9 A B C D E F 0 1 2 COM
00F800H D2 D3 D4 D4 D4 D5	Display area	
00F900H DZ	Display area	8 0 0 1 1 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0
00FA00H D2 D3	Display area	16 18 19 20 21 22 23
00FB00H D2 D3 D3 D4 D4 D5 D4 D5	Display area	24 25 26 27 28 29 30 31
00FC00H D2 D3		
00FD00H D2 D3 D4 D42H D45 D5	D0       D1       D2       D4       D6       D6       D7       SEG 0 1   2   3   4   5   6   7   8   9   10   11   12   13   14   15   16   17   18   19   20   21   22   23   24   25   26   27   28   29   30   31   32   33   34   35   39   30   41   42   43   44   45   46   47   48   49   50	

Fig. 5.12.6.1 1/32 duty and  $5 \times 8$  dots display memory map

Address/Data bit	0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 1 2 3 4 5 6 7 8 8 9 A B C D E F 0 1 1 2 3 4 5 6 7 8 8 9 A B C D E F 0 1 1 2 3 4 5 6 7 8 8 9 A B C D E F 0 1 1 2 3 4 5 6 7 8 8 9 A B C D E F 0 1 1 2 3 4 5 6 7 8 8 9 A B C D E F 0 1 1 2 3 4 5 6 7 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8	COM
D1 D2 D4	Display area	0 - 2 8 4
l l l .		
	Display area	6 8 6
200		
	Display area	0
(0)		
00FB00H D2   D3	Display area	19 18 19 20 20
00FC00H D2 D4	Display area	22 23 23 25 25 25 25 25 25 25 25 25 25 25 25 25
	Display area	27 28 29 30 30
2 2 2	1	50
SEG 0	0     2   3   4   5   6   7   8   9   1011 12113 14 19 16 17 18 19 20 21 22 23 24 25 26 27 28 29 30 31 32 33 35 35 35 35 35 35 35 35 35 35 35 35	

Fig. 5.12.6.2 1/32 duty and  $5 \times 5$  dots display memory map

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Address/Data bit 0 1 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2 3 4 5 6 7 8 9 A B C D E	E F 0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1 2
	Display area 0 (when "0" is set into DSPAR)
	8 9 10 11 11 13 13 14 15
	Display area 1 (when "1" is set into DSPAR)
	8 9 10 11 12 13 14 14 15
00FC00H D1 D3 D3 D4 D5 D6 D7 D5 D6 D7 D6 D7 D6 D7	
00FD00H	16 17 18 19 18 19 18 19 19 19 19 19 19 19 19 19 19 19 19 19

Fig. 5.12.6.3  $\,$  1/16 duty and 5  $\times$  8 dots display memory map

Fig. 5.12.6.4 1/16 duty and  $5 \times 5$  dots display memory map

COM	0 - 0 6 4 6 9 6		0 - 2 6 4 2 9 7			Ι
3 4 5 6 7 8 9 A B C D E F 0 1						
2 F 0 1 2 3 4 5 6 7 8 9 A B C D E F 0 1	ยก "0" is set into DSPAR)		Display area 1 (when "1" is set into DSPAR)			
1 E F O 1 2 3 4 5 6 7 8 9 A B C D E F			Display area 1 (wh			
0 1 2 3 4 5 6 7 8 9 A B C D		11 12 13 13 14 14 14 17 17 17 17 17 17 17 17 17 17 17 17 17	100   1   100   10	11 13 13 14 14 14 17 17 17 17 17 17 17 17 17 17 17 17 17	11 10 10 10 10 10 10 10 10 10 10 10 10 1	00 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1
Address/Data bit	00F800H DE	00F900H DD	00FA00H D2 D3 D4 D4 D4 D4 D4 D6 D7	00FB00H DD	00FC00H	00FD00H

Fig. 5.12.6.5 1/8 duty and  $5 \times 8$  dots display memory map

Fig. 5.12.6.6 1/8 duty and  $5 \times 5$  dots display memory map

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# 5.12.7 Display control

The display status of the built-in LCD driver and the contrast adjustment can be controlled with the built-in LCD controller. The LCD display status can be selected by display control registers LCDC0 and LCDC1. Setting the value and display status are shown in Table 5.12.7.1.

Table 5.12.7.1 LCD display control

LCDC1	LCDC0	LCD display				
1	1	All LCDs lit (Static)				
1	0	All LCDs out (Dynamic)				
0	1	Normal display				
0	0	Drive OFF				

All the dots in the LCD display can be turned on or off directly by the drive waveform output from the LCD driver, and data in the display memory is not changed. Also, since the common terminal at this time is set to static drive when all the dots are on and is set to dynamic drive when they are off, this function can be used as follows:

- (1) Since all dots on is binary output (VC5 and Vss) with static drive, the common/segment terminal can be used as a monitor terminal for the OSC1 oscillation frequency adjustment.
- (2) Since all dots off is dynamic drive, you can brink the entire LCD display without changing display memory data.

Selecting LCD drive OFF turns the LCD drive power circuit OFF and all the VC1–VC5 terminals go to VSS level. However, if external power supply has been selected by the mask option, the VC1–VC5 shift to floating status when drive is turned OFF. Furthermore, when the SLP instruction is executed, registers LCDC0 and LCDC1 are automatically reset to "0" (set to drive off) by hardware.

The LCD contrast can be adjusted in 16 stages. This adjustment is done by the contrast adjustment register LC0–LC3, and the setting values correspond to the contrast as shown in Table 5.12.7.2. However, if external power supply has been selected by the mask option, the contrast adjustment register LC0–LC3 is ineffective and contrast adjustment cannot be done.

Table 5.12.7.2 LCD contrast adjustment

LC3	LC2	LC1	LC0	Contrast
1	1	1	1	Dark
1	1	1	0	<b>1</b>
1	1	0	1	
:	:	:	:	
0	0	1	0	
0	0	0	1	$\downarrow$
0	0	0	0	Light

# 5.12.8 CL and FR outputs

In order for the S1C88349 to handle connection to an externally expanded LCD driver, output ports R25 and R26 can be used to output a CL signal (LCD synchronous signal) and FR signal (LCD frame signal), respectively.

The configuration of output ports R25 and R26 are shown in Figure 5.12.8.1.

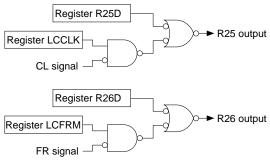


Fig. 5.12.8.1 Configuration of R25 and R26

The output control for the CL signal is done by the register LCCLK. When you set "1" for the LCCLK, the CL signal is output from the output port terminal R25, when "0" is set, the HIGH (VDD) level is output. At this time, "1" must always be set for the data register R25D.

The output control for the FR signal is done by the register LCFRM. When you set "1" for the LCFRM, the FR signal is output from the output port terminal R26, when "0" is set, the HIGH (VDD) level is output. At this time, "1" must always be set for the data register R26D.

The frequencies of each signal are changed as shown in Table 5.12.8.1 according to the drive duty selection.

Table 5.12.8.1 Frequencies of CL and FR signals

Drive duty	CL signal (Hz)	FR signal (Hz)
1/32	2,048	32
1/16	1,024	32
1/8	1,024	64

Since the signals are generated asynchronously from the registers LCCLK and LCFRM, when the signals are turned ON or OFF by the register settings, a hazard of a 1/2 cycle or less is generated. Figure 5.12.8.2 shows the output waveforms of the CL and FR signals.



Fig. 5.12.8.2 Output waveforms of CL and FR signals (when 1/16 duty is selected)

# 5.12.9 Control of LCD controller

Table 5.12.9.1 shows the LCD controller control bits.

Table 5.12.9.1 LCD controller control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF09	D7	-	_	-	-	_		
	D6	_	_	-	-	_		Constantly "0" when
	D5	_	_	-	-	_		being read
	D4	_	_	-	ı	_		
	D3	VCCHG	Reserved	1	0	0	R/W	
	D2	LCDB	Reserved	1	0	0	R/W	
	D1	LCDAJ	Power TYPE A (4.5V)/B (5.5V) switch	TYPE A	TYPE B	0	R/W	
	D0	DUTY8	LCD drive duty switch	1/8 duty	1/16, 1/32	0	R/W	*1
00FF10	D7	_	_	-	_	-		Constantly "0" when
	D6	_	_	-	_	-		being read
	D5	_	_	-	ı	_		being read
	D4	LCCLK	CL output control for expanded LCD driver	On	Off	0	R/W	
	D3	LCFRM	FR output control for expanded LCD driver	On	Off	0	R/W	
	D2	DTFNT	LCD dot font selection	5 x 5 dots	5 x 8 dots	0	R/W	
	D1	LDUTY	LCD drive duty selection	1/16 duty	1/32 duty	0	R/W	*1
	D0	SGOUT	R/W register	1	0	0	R/W	Reserved register
00FF11	D7	_	_	-	_	-		"0" when being read
	D6	DSPAR	LCD display memory area selection	Display area 1	Display area 0	0	R/W	
	D5	LCDC1	LCD display control			0	R/W	
			LCDC1 LCDC0 LCD display					These bits are reset
			1 1 All LCDs lit					to (0, 0) when
	D4	LCDC0	1 0 All LCDs out			0	R/W	SLP instruction
			0 1 Normal display					is executed.
			0 0 Drive off					
	D3	LC3	LCD contrast adjustment			0	R/W	
	D2	LC2	LC3 LC2 LC1 LC0 Contrast Dark			0	R/W	
	D1	LC1	1 1 1 1 Dark 1 1 1 0 :			0	R/W	
	D0	LC0	0 0 0 0 Light			0	R/W	

<sup>\*1</sup> Writing "1" to DUTY8 (FF09•D0) disables 1/16 and 1/32 duty selection using LDUTY (FF10•D1).

#### LCDAJ: 00FF09H•D1

Switches the internal power supply.

When "1" is written: TYPE A When "0" is written: TYPE B Reading: Valid

LCDAJ is the LCD power switching register. When "1" is written, the LCD power supply is set for TYPE A (4.5 V) and when "0" is written the LCD power supply is set for TYPE B (5.5 V). At initial reset, LCDAJ is set to "0" (TYPE B).

#### DUTY8: 00FF09H•D0

Switches the drive duty.

When "1" is written: 1/8 duty

When "0" is written: 1/16 or 1/32 duty

Reading: Valid

DUTY8 is the drive duty switching register. When "1" is written, 1/8 duty is selected and when "0" is written, 1/16 or 1/32 duty is selected.

When "1" is written to DUTY8, switching between 1/16 duty and 1/32 duty using LDUTY becomes invalid.

At initial reset, DUTY8 is set to "0" (1/16 or 1/32 duty).

#### LDUTY: 00FF10H•D1

Selects the drive duty.

When "1" is written: 1/16 duty When "0" is written: 1/32 duty Reading: Valid

When "0" is written to LDUTY, 1/32 duty is selected and the combined common/segment output terminal is switched to the common terminal.

When "1" is written to LDUTY, 1/16 duty is selected and the combined common/segment output terminal is switched to the segment terminal. When "1" is written to DUTY8, the combined common/segment terminals are fixed to the segment terminals and the setting of LDUTY becomes invalid.

The correspondence between the display memory bits set according to the drive duty, and the common/segment terminals are shown in Figures 5.12.6.1–5.12.6.6.

At initial reset, LDUTY is set to "0" (1/32 duty).

# **DTFNT: 00FF10H•D2**

Selects the dot font.

When "1" is written:  $5 \times 5$  dots When "0" is written:  $5 \times 8$  dots Reading: Valid

Select  $5 \times 8$  dots or  $5 \times 5$  dots type for the display memory area.

When "0" is written to DTFNT,  $5 \times 8$  dots is selected and when "1" is written,  $5 \times 5$  dots is selected. The correspondence between the display memory bits set according to the dot font, and the common/segment terminals are shown in Figures 5.12.6.1–5.12.6.6.

At initial reset, DTFNT is set to "0" ( $5 \times 8$  dots).

#### DSPAR: 00FF11H•D6

Selects the display area.

When "1" is written: Display area 1 When "0" is written: Display area 0 Reading: Valid

Selects which display area is secured for two screens in the display memory, will be displayed when 1/16 or 1/8 duty is selected.

When "0" is written to DSPAR, display area 0 is selected and when "1" is written, display area 1 is selected.

When 1/32 duty is selected, since the display area is only for one screen, the setting of DSPAR becomes invalid.

The correspondence between the display memory bits set according to the display area, and the common/segment terminals are shown in Figures 5.12.6.1–5.12.6.6.

At initial reset, DSPAR is set to "0" (display area 0).

# LCDC0, LCDC1: 00FF11H•D4, D5

Controls the LCD display.

Table 5.12.9.2 LCD display control

LCDC1	LCDC0	LCD display
1	1	All LCDs lit (Static)
1	0	All LCDs out (Dynamic)
0	1	Normal display
0	0	Drive OFF

The four settings mentioned above can be made without changing the display memory data. At initial reset and in the SLEEP status, this register is set to "0" (drive off).

#### LC0-LC3: 00FF11H•D0-D3

Adjusts the LCD contrast.

Table 5.12.9.3 LCD contract adjustment

LC3	LC2	LC1	LC0	Contrast
1	1	1	1	Dark
1	1	1	0	1
1	1	0	1	
1	1	0	0	
1	0	1	1	
1	0	1	0	
1	0	0	1	
1	0	0	0	
0	1	1	1	
0	1	1	0	
0	1	0	1	
0	1	0	0	
0	0	1	1	
0	0	1	0	
0	0	0	1	$\downarrow$
0	0	0	0	Light

The contrast can be adjusted in 16 stages as mentioned above. This adjustment changes the drive voltage on terminals VC1-VC5. At initial reset, this register is set to "0".

### LCCLK: 00FF10H•D4

Controls the CL signal output.

When "1" is written: CL signal output When "0" is written: HIGH level (DC) output

Reading: Valid

LCCLK is the output control register for CL signal. When "1" is set, the CL signal is output from the output port terminal R25 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R25D.

At initial reset, LCCLK is set to "0" (HIGH level output).

#### *LCFRM:* 00FF10H•D3

120

Controls the FR signal output.

When "1" is written: FR signal output

When "0" is written: HIGH level (DC) output

Reading: Valid

LCFRM is the output control register for FR signal. When "1" is set, the FR signal is output from the output port terminal R26 and when "0" is set, HIGH (VDD) level is output. At this time, "1" must always be set for the data register R26D.

At initial reset, LCFRM is set to "0" (HIGH level output).

# 5.12.10 Programming notes

- (1) Since the CL and FR signals are generated asynchronously from the output control registers LCCLK and LCFRM, when the signals is turned ON or OFF by setting of the registers LCCLK and LCFRM, a hazard of a 1/2 cycle or less is generated.
- (2) When the SLP instruction is executed, display control registers LCDC0 and LCDC1 are automatically reset to "0" by hardware. Furthermore, in the SLEEP status, HIGH (VDD) level is output for the CL and FR signals. (When registers R25D and R26D are set to "1".)

# 5.13 Sound Generator

# 5.13.1 Configuration of sound generator

The S1C88349 has a built-in sound generator for generating BZ (buzzer) signal.

BZ signals generated from the sound generator can be output from the R50 output port terminal. Aside permitting the respective setting of the buzzer signal frequency and sound level (duty adjustment) to 8 stages, it permits the adding of a digital envelope by means of duty ratio control. It also has a one-shot output function for outputting key operated sounds.

Figure 5.13.1.1 shows the configuration of the sound generator.

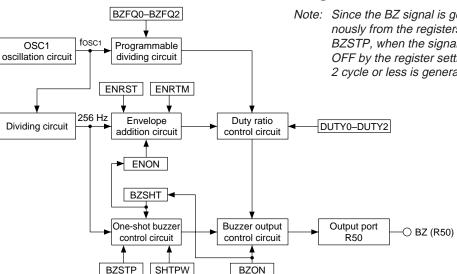


Fig. 5.13.1.1 Configuration of sound generator

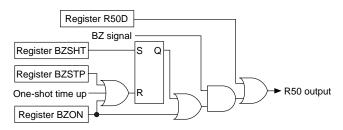


Fig. 5.13.2.1 Configuration of R50



Fig. 5.13.2.2 Output waveform of BZ signal

# 5.13.2 Control of buzzer output

BZ signal can be output from the R50 output port terminal.

The configuration of the output port R50 is shown in Figure 5.13.2.1.

The output control for the BZ signal generated by the sound generator is done by the buzzer output control register BZON, one-shot buzzer trigger bit BZSHT and one-shot buzzer forced stop bit BZSTP. When "1" is set to BZON or BZSHT, the BZ signal is output from the R50 output port terminal and when "0" is set to BZON or "1" is set to BZSTP, the LOW (Vss) level is output. At this time, "0" must always be set for the output data register R50D. Figure 5.13.2.2 shows the output waveform of the BZ signal.

Note: Since the BZ signal is generated asynchronously from the registers BZON, BZSHT and BZSTP, when the signal is turned ON or OFF by the register settings, a hazard of a 1/ 2 cycle or less is generated.

# 5.13.3 Setting of buzzer frequency and sound level

The BZ signal is a divided signal using the OSC1 oscillation circuit (32.768 kHz) as the clock source and 8 frequencies can be selected. This selection is done by the buzzer frequency selection register BZFQ0–BZFQ2. The setting value and buzzer frequency correspondence is shown in Table 5.13.3.1. By selecting the duty ratio of the BZ signal from among 8 types, the buzzer sound level can be adjusted. This selection is made in the duty ratio selection register DUTY0–DUTY2. The setting value and duty ratio correspondence is shown in Table 5.13.3.2.

Table 5.13.3.1 Buzzer signal frequency settings

BZFC	ີ 22	BZFQ1	BZFQ0	Buzzer frequency (Hz)
0		0	0	4096.0
0		0	1	3276.8
0		1	0	2730.7
0		1	1	2340.6
1		0	0	2048.0
1		0	1	1638.4
1		1	0	1365.3
1		1	1	1170.3

Table 5.13.3.2 Duty ratio settings

				Duty	ratio by buzze	er frequencies	(Hz)
Level	DUTY2	DUTY1	DUTY0	4096.0	3276.8	2730.7	2340.6
				2048.0	1638.4	1365.3	1170.3
Level 1 (Max)	0	0	0	8/16	8/20	12/24	12/28
Level 2	0	0	1	7/16	7/20	11/24	11/28
Level 3	0	1	0	6/16	6/20	10/24	10/28
Level 4	0	1	1	5/16	5/20	9/24	9/28
Level 5	1	0	0	4/16	4/20	8/24	8/28
Level 6	1	0	1	3/16	3/20	7/24	7/28
Level 7	1	1	0	2/16	2/20	6/24	6/28
Level 8 (Min)	1	1	1	1/16	1/20	5/24	5/28

Duty ratio refers to the ratio of pulse width to the pulse cycle; given that HIGH level output time is TH, and LOW level output time is TL the BZ signal becomes TH/(TH+TL).

When DUTY0-DUTY2 have all been set to "0", the duty ratio becomes maximum and the sound level also becomes maximum. Conversely, when DUTY0-DUTY2 have all been set to "1", the duty ratio becomes minimum and the sound level also becomes minimum.

Note that the duty ratio setting differ depending on frequency. See Table 5.13.3.2.

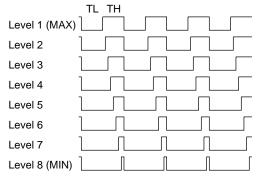


Fig. 5.13.3.1 Duty ratio of buzzer signal waveform

Note: When using the digital envelope, the DUTY0-DUTY2 setting becomes invalid.

# 5.13.4 Digital envelope

A digital envelope with duty control can be added to the BZ signal.

The envelope can be realized by staged changing of the same duty ratio as detailed in Table 5.13.3.2 in the preceding section from level 1 (maximum) to level 8 (minimum).

The addition of an envelope to the buzzer signal can be done by writing "1" to the envelope control register ENON. When "0" is written, the duty ratio is set at the level selected in DUTY0–DUTY2. By writing "1" to ENON to turn the buzzer output ON (writing "1" to BZON), a BZ signal with a level 1 duty ratio is output, and then the duty ratio can be attenuated in stages to level 8. The attenuated envelope can be returned to level 1 by writing "1" to the envelope reset bit ENRST. When attenuated to level 8, the duty level remains at level 8 until the buzzer output is turned OFF (writing "0" to BZON) or writing "1" to ENRST.

The stage changing time for the envelope level can be selected either 125 msec or 62.5 msec by the envelope attenuation time selection register ENRTM. Figure 5.13.4.1 shows the timing chart of the digital envelope.

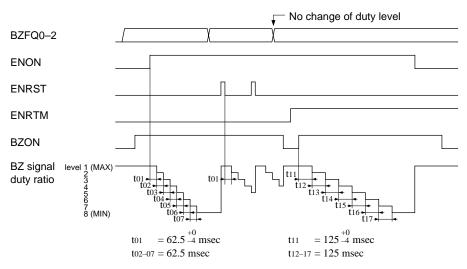


Fig. 5.13.4.1 Timing chart of digital envelope

# 5.13.5 One-shot output

The sound generator has a built-in one-shot output function for outputting a short duration buzzer signal for key operation sounds and similar effects. Either 125 msec or 31.25 msec can be selected by the one-shot buzzer duration selection register SHTPW for buzzer signal output time.

The output control of the one-shot buzzer is done by writing "1" to the one-shot buzzer trigger BZSHT, then the BZ signal is output in synchronization with the internal 256 Hz signal from the R50 output port terminal. Thereafter, when the set time has elapsed, the BZ signal in synchronization with the 256 Hz signal automatically goes OFF in the same manner.

The BZSHT can be read to determine status. When BZSHT is "1", it indicates a BUSY status (during one-shot output) and when BZSHT is "0", it indicates a READY status (during stop).

When you want to turn the BZ signal OFF prior to the elapse of the set time, the BZ signal can be immediately stopped (goes OFF in asynchronization with 256 Hz signal) by writing "1" to the one-shot forced stop bit BZSTP. Since the one-shot output has a short duration, an envelope cannot be added. (When "1" is written to BZSHT, ENON is automatically reset to "0".) Consequently, only the frequency and sound level can be set for one-shot output.

The control for the one-shot output is invalid during normal buzzer output.

Figure 5.13.5.1 shows the timing chart of the one-shot output.

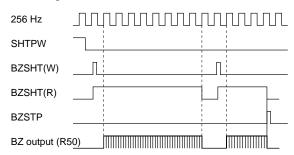


Fig. 5.13.5.1 Timing chart of one-shot output

# 5.13.6 Control of sound generator

Table 5.13.6.1 shows the sound generator control bits.

Table 5.13.6.1 Sound generator control bits

Address	Bit	Name		Fu	nction				1	0	SR	R/W	Comment
00FF44	D7	_	_						-	-	_		Constantly "0" when
	D6	BZSTP	One-shot buz	zzer forc	ibly sto	p			Forcibly stop	No operation	_	W	being read
	D5	BZSHT	One-shot buz	zzer trig	ger/stat	us		R	Busy	Ready	0	R/W	
								W	Trigger	No operation			
	D4	SHTPW	One-shot buz	zzer dura	ation w	idth sele	ection	l	125 msec	31.25 msec	0	R/W	
	D3	ENRTM	Envelope att	enuation	time				1 sec	0.5 sec	0	R/W	
	D2	ENRST	Envelope res	et					Reset	No operation	-	W	"0" when being read
		ENON	Envelope On	Off co	ntrol				On	Off	0	R/W	*1
	D0	BZON	Buzzer outpu	ıt contro	ol				On	Off	0	R/W	
00FF45	D7	_	_						-	-	_		"0" when being read
	D6	DUTY2	Buzzer signa	-							0	R/W	
			DUTY2-1			quency (I 2730.7		_					
			2 1 0	2048.0	1638.4	1365.3	1170	.3				L	
	D5	DUTY1	$\begin{array}{ccccc} 0 & 0 & 0 \\ 0 & 0 & 1 \end{array}$	8/16 7/16		12/24 11/24					0	R/W	
			0 1 0	6/16	6/20	10/24							
			0 1 1	5/16	5/20	9/24	9/28					L	
	D4	DUTY0	1 0 0	4/16	4/20	8/24	8/28				0	R/W	
			$\begin{array}{cccccccccccccccccccccccccccccccccccc$	3/16 2/16	3/20 2/20	7/24 6/24	7/28 6/28						
			1 1 1	1/16	1/20	5/24	5/28						
	D3	_	_						-	-	_		"0" when being read
	D2	BZFQ2	Buzzer frequ	ency sel	lection						0	R/W	
			BZFQ2 BZ	FQ1 B2	ZFQ0	Frequenc	y (Hz	2)					
			0	0	0	4096	5.0	_					
	D1	BZFQ1	-	0	1	3276					0	R/W	
			-	1	0	2730 2340							
			-	0	1	2048							
	D0	BZFQ0	_	0	1	1638					0	R/W	
			1	1	0	1365							
			1	1	1	1170	0.3						

<sup>\*1</sup> Reset to "0" during one-shot output.

#### BZON: 00FF44H•D0

Controls the BZ signal output.

When "1" is written: BZ signal output
When "0" is written: LOW level (DC) output

Reading: Valid

BZON is the output control register for BZ signal. When "1" is set, the BZ signal is output from the output port terminal R50 and when "0" is set, LOW (Vss) level is output. At this time, "0" must always be set for the data register R50D.

At initial reset, BZON is set to "0" (LOW level output).

## BZFQ0-BZFQ2: 00FF45H•D0-D2

Selects the BZ signal frequency.

Table 5.13.6.2 Buzzer frequency settings

BZFQ2	BZFQ1	BZFQ0	Buzzer frequency (Hz)
0	0	0	4096.0
0	0	1	3276.8
0	1	0	2730.7
0	1	1	2340.6
1	0	0	2048.0
1	0	1	1638.4
1	1	0	1365.3
1	1	1	1170.3

The buzzer frequency can be selected from among the above 8 types that have divided the OSC1 clock. At initial reset, this register is set at "0" (4096.0 Hz).

#### DUTY0-DUTY2: 00FF45H•D4-D6

Selects the duty ratio of the BZ signal.

Table 5.13.6.3 Duty ratio settings

				Duty	ratio by buzze	er frequencies	(Hz)
Level	DUTY2	DUTY1	DUTY0	4096.0	3276.8	2730.7	2340.6
				2048.0	1638.4	1365.3	1170.3
Level 1 (Max)	0	0	0	8/16	8/20	12/24	12/28
Level 2	0	0	1	7/16	7/20	11/24	11/28
Level 3	0	1	0	6/16	6/20	10/24	10/28
Level 4	0	1	1	5/16	5/20	9/24	9/28
Level 5	1	0	0	4/16	4/20	8/24	8/28
Level 6	1	0	1	3/16	3/20	7/24	7/28
Level 7	1	1	0	2/16	2/20	6/24	6/28
Level 8 (Min)	1	1	1	1/16	1/20	5/24	5/28

The buzzer sound level can be adjusted by selecting the duty ratio from among the above 8 types. However, when the envelope has been set to ON (ENON = "1"), this setting becomes invalid.

At initial reset, this register is set to "0" (level 1).

# **ENRST:** 00FF44H•D2

Resets the envelope.

When "1" is written: Reset

When "0" is written: No operation

Reading: Always "0"

The envelope is reset by writing "1" to ENRST and the duty ratio returns to level 1 (maximum). Writing "0" to ENRST and writing "1" when an envelope has not been added become invalid. Since ENRST is exclusively for writing, it always becomes "0" during reading.

#### ENON: 00FF44H•D1

Controls the addition of an envelope to the BZ signal.

When "1" is written: ON When "0" is written: OFF Reading: Valid

By writing "1" to ENON, an envelope can be added to BZ signal output. When "0" is written, an envelope is not added and the BZ signal is fixed at the duty ratio selected in DUTY0-DUTY2. At initial reset and when "1" is written to BZSHT.

ENON is set to "0" (OFF).

#### ENRTM: 00FF44H•D3

Selects the envelope attenuation time that is added to the BZ signal.

When "1" is written: 1.0 sec

 $(125 \text{ msec} \times 7 = 875 \text{ msec})$ 

When "0" is written: 0.5 sec

 $(62.5 \text{ msec} \times 7 = 437.5 \text{ msec})$ 

Reading: Valid

The attenuation time of the digital envelope is determined by the time for changing the duty ratio. The duty ratio is changed in 125 msec (8 Hz) units when "1" is written to ENRTM and in 62.5 msec (16 Hz) units, when "0" is written.

This setting becomes invalid when an envelope has been set to OFF (ENON = "0").

At initial reset, ENRTM is set to "0" (0.5 sec).

#### SHTPW: 00FF44H•D4

Selects the output duration width of the one-shot buzzer.

When "1" is written: 125 msec When "0" is written: 31.25 msec Reading: Valid

The one-shot buzzer output duration width is set to 125 msec when "1" is written to SHTPW and 31.25 msec, when "0" is written.

At initial reset, SHTPW is set to "0" (31.25 msec).

#### BZSHT: 00FF44H•D5

Controls the one-shot buzzer output.

When "1" is written: Trigger When "0" is written: No operation

When "1" is read: Busy When "0" is read: Ready

Writing "1" into BZSHT causes the one-shot output circuit to operate and the BZ signal to be output. The buzzer output is automatically turned OFF after the time set by SHTPW has elapsed. At this time, "0" must always be set for the data register R50D.

The one-shot output is only valid when the normal buzzer output is OFF (BZON = "0") status. The trigger is invalid during ON (BZON = "1") status. When a re-trigger is assigned during a one-shot output, the one-shot output time set with SHTPW is measured again from that point. (time extension) The operation status of the one-shot output circuit can be confirmed by reading BZSHT, when the one-shot output is ON (busy), BZSHT reads "1" and when the output is OFF (ready), it reads "0". At initial reset, BZSHT is set to "0" (ready).

#### BZSTP: 00FF44H•D6

Forcibly stops the one-shot buzzer output.

When "1" is written: Forcibly stop When "0" is written: No operation Reading: Constantly "0"

By writing "1" into BZSTP, the one-shot buzzer output can be stopped prior to the elapsing of the time set with SHTPW.

Writing "0" is invalid and writing "1" except during one-shot output is also invalid.

When "1" is written to BZSHT and BZSTP simultaneously, BZSTP takes precedence and one-shot output becomes stop status.

Since BZSTP is for writing only, during readout it is constantly set to "0".

# 5.13.7 Programming notes

- (1) Since the BZ signal is generated asynchronously from the register BZON, when the signal is turned ON or OFF by the register setting, a hazard of a 1/2 cycle or less is generated.
- (2) The SLP instruction has executed when the BZ signal is in the enable status (BZON = "1" or BZSHT = "1"), an unstable clock is output from the R50 output port terminal at the time of return from the SLEEP status. Consequently, when shifting to the SLEEP status, you should set the BZ signal to the disable status (BZON = BZSHT = "0") prior to executing the SLP instruction.
- (3) The one-shot output is only valid when the normal buzzer output is OFF (BZON = "0") status. The trigger is invalid during ON (BZON = "1") status.

# 5.14 Analog Comparator

# 5.14.1 Configuration of analog comparator

The S1C88349 has an MOS input analog comparator built into two channels. The respective analog comparators have two differential input terminals (inverted input terminal CMPMx and non-inverted input terminal CMPPx) that are available for general purpose use.

Figure 5.14.1.1 shows the configuration of the analog comparator.

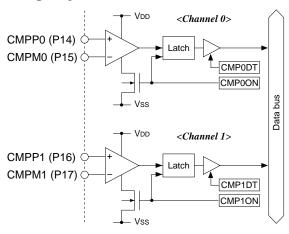


Fig. 5.14.1.1 Configuration of analog comparator

Since the input terminals of the analog comparator CMPP0, CMPM0, CMPP1 and CMPM1 are common to I/O ports P14–P17, when using as the input terminal for the analog comparator, "0" (input mode) must be written to I/O control registers IOC14–IOC17.

Table 5.14.1.1 Input terminal configuration

Terminal	When analog comparator is used
P14	CMPP0
P15	CMPM0
P16	CMPP1
P17	CMPM1

Note: The P14–P17 terminals are shared with the A/D converter input ports. Therefore, do not run the A/D converter when the analog comparator is used.

# 5.14.2 Mask option

Since the input terminals of the analog comparator are common to the I/O ports, the mask option for the I/O port corresponding to the channel to be used must be set to "Gate direct".

	I/O ports pull-up resistor	
	P14 (CMPP0) □ With resistor	Gate direct
	P15 (CMPM0) $\square$ With resistor	☐ Gate direct
	P16 (CMPP1) □ With resistor	☐ Gate direct
	P17 (CMPM1) $\square$ With resistor	Gate direct
ı		

<sup>\* &</sup>quot;\" above shows an example of both channels being used.

# 5.14.3 Analog comparator operation

By writing "1" to the analog comparator control register CMPxON, the analog comparator goes ON, and the analog comparator starts comparing the external voltages that have been input to the two differential input terminals CMPPx and CMPMx. The result can be read from the comparator comparison result detection bit CMPxDT through the latch and when CMPPx (+) > CMPMx (-), it is "1" and when CMPPx (+) < CMPMx (-), it is "0". After the analog comparator has been turned ON, a maximum time of 3 msec is necessary until output stabilizes. Consequently, you should allow an adequate waiting time after turning the analog comparator ON, before reading the comparison result.

When the analog comparator is turned OFF, the comparison result at that point will be latched and the concerned data can be read thereafter, until the analog comparator is turned ON.

You should turn the analog comparator OFF, when it is not necessary, so as to reduce current consumption.

See Chapter 8, "ELECTRICAL CHARACTERISTICS" for the input voltage range.

Note: Since the input terminals of the analog comparator are shared with the I/O ports, the I/O control registers (IOC14–IOC17) corresponding to the channel to be used must be set to the input mode.

# 5.14.4 Control of analog comparator

Table 5.14.4.1 shows the analog comparator control bits.

Table 5.14.4.1 Analog comparator control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF13	D7	-	_	-	-	_		
	D6	_	_	-	-	_		Constantly "0" when
	D5	_	_	-	-	_		being read
	D4	_	_	-	-	_		
	D3	CMP10N	Comparator 1 On/Off control	On	Off	0	R/W	
	D2	CMP0ON	Comparator 0 On/Off control	On	Off	0	R/W	
	D1	CMP1DT	Comparator 1 data	+>-	+<-	0	R	
	D0	CMP0DT	Comparator 0 data	+>-	+<-	0	R	

#### CMP0ON, CMP1ON: 00FF13H•D2, D3

Controls the analog comparator ON/OFF.

When "1" is written: ON When "0" is written: OFF Reading: Valid

The analog comparator 0 goes ON by writing "1" to CMP0ON and goes OFF, when "0" is written. The analog comparator 1 can be controlled with CMP1ON in the same way.

At initial reset, this register is set "0" (OFF).

# CMP0DT, CMP1DT: 00FF13H•D0, D1

The comparison result of the analog comparator can be read out.

When "1" is read: CMPPx (+) > CMPMx (-)When "0" is read: CMPPx (+) < CMPMx (-)

Writing: Invalid

The result of analog comparator 0 can be read from CMP0DT. When the status of external voltage input to differential input terminals CMPP0 and CMPM0 is CMPP0 (+) > CMPM0 (-), CMP0DT becomes "1" and when it is CMPP0 (+) < CMPM0 (-), CMP0DT becomes "0".

As the same way, the comparison result between CMPP1 and CMPM1 can be read from CMP1DT. When the analog comparator is turned OFF, the latched result immediately prior to going OFF is read out.

At initial reset, this bit is set to "1".

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# 5.14.5 Programming notes

- (1) To reduce current consumption, turn the analog comparator OFF (CMP0ON = CMP1ON = "0") when it is not necessary.
- (2) After the analog comparator has been turned ON, a maximum time of 3 msec is necessary until output stabilizes. Consequently, you should allow an adequate waiting time after turning the analog comparator ON, before reading the comparison result.
- (3) Since the input terminals of the analog comparator are common to the I/O ports, the I/O control registers (IOC14–IOC17) corresponding to the channel to be used must be set to the input mode.
- (4) The P14–P17 terminals are shared with the A/D converter input ports. Therefore, do not run the A/D converter when the analog comparator is used.

S1C88349 TECHNICAL MANUAL

# 5.15 A/D Converter

# 5.15.1 Characteristics and configuration of A/D converter

The S1C88349 has a built-in A/D converter with the following characteristics.

• Conversion formula: Successive-approximation

type

• Resolution: 10 bits

Input channel: Maximum 4 channels
 Conversion time: Minimum 22 µsec

(in 1 MHz operation)

- $\bullet$  Setting of analog conversion voltage range is possible with reference voltage terminal (AVREF)
- A/D conversion result is possible to read from 10-bit data register
- Sample & hold circuit built-in
- A/D conversion completion generates an interrupt

Figure 5.15.1.1 shows the configuration of the A/D converter.

# 5.15.2 Terminal configuration of A/D converter

The terminals used with the A/D converter are as follows:

#### ■ AVDD, AVss (power supply input terminal)

The AVDD and AVss terminals are power supply terminals for the A/D converter. The voltage should be input as AVDD  $\leq$  VDD and AVss = Vss.

#### ■ AVREF (reference voltage input terminal)

The AVREF terminal is the reference voltage terminal of the analog block. Input voltage range of the A/D conversion is decided by this input. The voltage should be input as AVREF  $\leq$  AVDD.

#### ■ AD4-AD7 (analog input terminal)

The analog input terminals AD4–AD7 are shared with the I/O port terminals P14–P17. Therefore, it is necessary to set them for the A/D converter by software when using them as analog input terminals. This setting can be done for each terminal. (Refer to Section 5.15.4 for setting.)

At initial reset, all the terminals are set in the  ${\rm I/O}$  port terminal.

Analog voltage value AVIN that can be input is in the range of AVss  $\leq$  AVIN  $\leq$  AVREF.

Note: Since the P14–P17 terminals are shared with the analog comparator, the A/D converter and the analog comparator cannot be used simultaneously. When using the A/D converter, do not run the analog comparator.

# 5.15.3 Mask option

I/O ports pull-up resistors	
P14 (AD4)   With resistor	☐ Gate direct
P15 (AD5)   With resistor	☐ Gate direct
P16 (AD6)   With resistor	☐ Gate direct
P17 (AD7)   With resistor	Gate direct

\* "\" above shows an example of both channels being used.

The input terminals of the A/D converter are shared with the I/O port terminals P14–P17. Therefore, the terminal specification of the A/D converter is decided by setting the I/O port mask option. Select "Gate direct" for the port corresponding to the channel to be used to obtain the conversion precision.

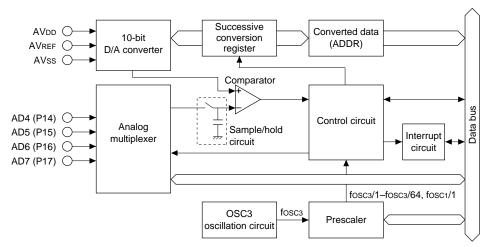


Fig. 5.15.1.1 Configuration of A/D converter

# 5.15.4 A/D conversion

#### ■ Setting the A/D input terminal

When using the A/D converter, it is necessary to set up the terminals used for analog input from the P14-P17 initialized as the I/O port terminals. Four terminals can all be used as analog input terminals.

The PAD (PAD4–PAD7) register is used to set analog input terminals. When the PAD register bits are set to "1", the corresponding terminals function as the analog input terminals.

Table 5.15.4.1 Correspondence between A/D input terminal and PAD register

Terminal	A/D input control register
P14 (AD4)	PAD4
P15 (AD5)	PAD5
P16 (AD6)	PAD6
P17 (AD7)	PAD7

#### ■ Setting the input clock

The A/D conversion clock can be selected from eight types shown in Table 5.15.4.2. The selection is done using the PSAD register.

Table 5.15.4.2 Input clock selection

Sele	ction reg	jister	Division	Output
PSAD2	PSAD1	PSAD0	ratio	control
1	1	1	fosc1/1	PRAD
1	1	0	fosc3/64	register
1	0	1	fosc3/32	
1	0	0	fosc3/16	"1": ON
0	1	1	fosc3/8	"0": OFF
0	1	0	fosc3/4	
0	0	1	fosc3/2	
0	0	0	fosc3/1	

The selected clock is input to the A/D converter by writing "1" to the PRAD register.

Notes: • When the OSC3 oscillation circuit is made the clock source, it is necessary to turn the OSC3 oscillation ON, prior to using the A/D converter.

From the time the OSC3 oscillation circuit is turning ON until oscillation stabilizes, an interval of several msec to several 10 msec is necessary. Consequently, you should allow an adequate waiting time after turning the OSC3 oscillation circuit ON before starting the count of the programmable timer. (The oscillation start time will vary somewhat depending on the oscillator and on external parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)

At initial reset, OSC3 oscillation circuit is set to OFF status.

- The clock division ratio (see Table 5.15.4.2) must be set so that the A/D conversion clock frequency is 1 MHz or less. Furthermore, the A/D conversion clock frequency should be changed according to the voltage to be used. Refer to Chapter 8, "ELECTRICAL CHARACTERISTICS".
- The input clock should be set when the A/D converter stops. Changing in the A/D converter operation may cause a malfunction.
- To prevent malfunction, do not start A/D conversion (writing to the CHS register) when the A/D conversion clock is not being output from the prescaler, and do not turn the prescaler output clock off during A/D conversion.

## ■ Selecting the input signal

The analog signals from the AD4 (P14)–AD7 (P17) terminals are input to the multiplexer, and the analog input channel for A/D conversion is selected by software. This selection can be done using the CHS register as shown in Table 5.15.4.3.

Table 5.15.4.3 Selection of analog input channel

CHS1	CHS0	Input channel
1	1	AD7
1	0	AD6
0	1	AD5
0	0	AD4

#### ■ A/D conversion operation

An A/D conversion starts by writing data to the ADRUN register. For example, when performing A/D conversion using AD7 as the analog input, write "1" (1, 1) to the CHS register (CHS1, CHS0) and then write "1" to the ADRUN register. The A/D input channel is selected and the A/D conversion starts. However, it is necessary that the P17 terminal has been set as an analog input terminal. The built-in sample & hold circuit starts sampling of the analog input specified from tAD after writing. When the sampling is completed, the held analog input voltage is converted into a 10-bit digital value in successive-approximation architecture. The conversion result is loaded into the ADDR (ADDR0-ADDR9) register. ADDR0 is the LSB and ADDR9 is the MSB.

Note: If the CHS register selects an input channel which is not included in the analog input terminals set by the PAD register (the PAD register can select several terminals simultaneously), the A/D conversion does not result in a correct converted value.

#### Example)

Terminal setting:

PAD5 = 1, PAD7 = PAD6 = PAD4 = 0

(AD5 terminal is used)

Selection of input channel:

CHS1 = 0, CHS0 = 0

(AD4 is selected)

In a setting like this, the A/D conversion result will be invalid because the contents of the settings are not matched.

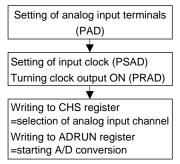


Fig. 5.15.4.1 Flowchart for starting A/D conversion

An A/D conversion is completed when the conversion result is loaded into the ADDR register. At that point, the A/D converter generates an interrupt (explained in the next section).

Figure 5.15.4.2 shows the timing chart of A/D conversion.

# 5.15.5 Interrupt function

to the CPU.

The A/D converter can generate an interrupt when an A/D conversion has completed.

Figure 5.15.5.1 shows the configuration of the A/D converter interrupt circuit.

The A/D converter sets the interrupt factor flag FAD to "1" when it stores the conversion. At this time, if the interrupt enable register EAD is "1" and the interrupt priority register PADC (2 bits) is set to a higher level than the setting of the interrupt flags (I0 and I1), an interrupt is generated

By setting the EAD register to "0", the interrupt to the CPU can also be disabled. However, the interrupt factor flag is set to "1" when an A/D conversion has completed regardless of the interrupt enable register and interrupt priority register settings.

The interrupt factor flag set in "1" is reset to "0" by writing "1".

Refer to Section 5.17, "Interrupt and Standby Status", for details of the interrupt control registers and operations subsequent to interrupt generation. The exception processing vector address for the A/D conversion completion interrupt has been set in 000024H.

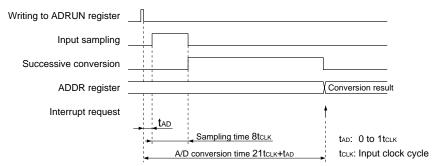


Fig. 5.15.4.2 Timing chart of A/D conversion

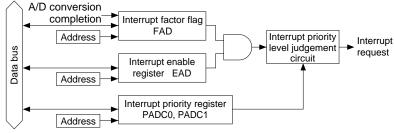


Fig. 5.15.5.1 Configuration of A/D converter interrupt circuit

# 5.15.6 Control of A/D converter

Table 5.15.6.1 shows the A/D converter control bits.

Table 5.15.6.1(a) A/D converter control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF80	D7	_	_	_	_	_		Common
001100	D6		_	_	_			Constantly "0" when
	D5	_	_	_	_	_		being read
	D4		_	_	_			being read
	_	PRAD	A/D converter clock control	On	Off	0	R/W	
		PSAD2	A/D converter division ratio	Oil	OII	0	R/W	
	-	. 0, .22	PSAD2 PSAD1 PSAD0 Division ratio				10	
			$\frac{15702}{1} \frac{15701}{1} \frac{15700}{1} \frac{15700}{1} \frac{157100}{1}$					
	D1	PSAD1	1 1 0 fosc3 / 64			0	R/W	
			1 0 1 fosc3 / 32					
			1 0 0 fosc3 / 16 0 1 1 fosc3 / 8					
	D0	PSAD0	0 1 1 fosc3 / 8 0 1 0 fosc3 / 4			0	R/W	
		. 0, .20	0 0 1 foscs/2				10	
			0 0 0 fosc3 / 1					
00FF81	D7	PAD7	P17 A/D converter input control			0	R/W	
		PAD6	P16 A/D converter input control	A/D		0	R/W	
	_	PAD5	P15 A/D converter input control	converter	I/O port	0	R/W	
		PAD4	P14 A/D converter input control	input		0	R/W	
	D3	_	_	_	_	_		
	D2	_	_	_	_	_		Constantly "0" when
	D1	_	_	_	_	_		being read
	D0	_	_	_	_	_		
00FF82	D7	ADRUN	A/D conversion start control register	Start	Invalid	0	W	
	D6	_	_	_	_	_		
	D5	_	_	_	_	_		Constantly "0" when
	D4	_	_	_	_	_		being read
	D3	_	_	_	_	_		
	D2	_	_	_	_	_		
	D1	CHS1	Analog input channel selection			0	R/W	
			CHS1 CHS0 Input channel					
			1 1 AD7					
	D0	CHS0	1 0 AD6			0	R/W	
			0 1 AD5					
			0 0 AD4					
00FF83	D7	ADDR9	A/D conversion result D9 (MSB)					
	D6	ADDR8	A/D conversion result D8					
	D5	ADDR7	A/D conversion result D7					
	D4	ADDR6	A/D conversion result D6				R	
	D3	ADDR5	A/D conversion result D5			_	K	
	D2	ADDR4	A/D conversion result D4					
	D1	ADDR3	A/D conversion result D3					
	D0	ADDR2	A/D conversion result D2					
00FF84	D7	_	_	-	-	_		
	D6	_	_	-	-	_		
	D5	_	_	-	-	_		Constantly "0" when
	D4	_	_	-	_	_		being read
	D3	_	_	-	-	_		
	D2	_	_	-	-	_		
	D1	ADDR1	A/D conversion result D1				R	
	D0	ADDR0	A/D conversion result D0 (LSB)					

Table 5.15.6.1(b) A/D converter control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF28	D7	PADC1	A/D converter interrupt priority register	1 1	DC1 PADC0 Priority Level 3		R/W	
	D6	PADC0		$\begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}$	Level 2 Level 1 Level 0	0	R/W	
	D5	_	Reserved	Prohibited	-	0		Do not write "1".
	D4	_	Reserved	Prohibited	-	0		
	D3	ı	_	-	-	_		
	D2	_	_	-	-	_		Constantly "0" when
	D1	_	_	-	-	_		being read
	D0	_	_	-	-	_		
00FF2A	D7	EAD	A/D converter interrupt enable register	Enable	Disable	0	R/W	
	D6	ı	Reserved	-	-	0	R/W	
	D5	_	_	-	-	_		
	D4	ı	_	_	-	-		
	D3	-	_	-	-	_		Constantly "0" when
	D2	-	_	-	-	_		being read
	D1	-	_	-	-	_		
	D0	_	_	_	_	_		
00FF2C	D7	FAD	A/D converter interrupt factor flag R)	Generated	Not generated	0	R/W	
			W	Reset	No operation			
	D6	ı	Reserved	_	_	0	R/W	
	D5	_	_	-	-	_		
	D4	_	_	-	-	_		
	D3	-	_	-	-	_		Constantly "0" when
	D2	ı	_	-	-	_		being read
	D1	_	_	-	-	_		
	D0	_	_	-	_	_		

### *PAD4-PAD7: 00FF81H•D4-D7*

Sets the P14–P17 terminals as the analog input terminals for the A/D converter.

When "1" is written: A/D converter input

When "0" is written: I/O port Reading: Valid

When "1" is written to PADn, the P1n terminal is set to the analog input terminal ADn. (n=4-7)

When "0" is written, the terminal is used with the I/O port

At initial reset, this register is set to "0" (I/O port).

### *PSAD0-PSAD2: 00FF80H•D0-D2*

Selects the clock for the A/D converter.

Table 5.15.6.2 Input clock selection

	ction reg		Division	Output
PSAD2	PSAD1	PSAD0	ratio	control
1	1	1	fosc1/1	PRAD
1	1	0	fosc3/64	register
1	0	1	fosc3/32	
1	0	0	fosc3/16	"1": ON
0	1	1	fosc3/8	"0": OFF
0	1	0	fosc3/4	
0	0	1	fosc3/2	
0	0	0	fosc3/1	

This setting controls the division ratio of the prescaler.

At initial reset, this register is set to "0" (fosc3/1).

#### PRAD: 00FF80H•D3

Controls the clock supply to the A/D converter.

When "1" is written: ON When "0" is written: OFF Reading: Invalid

By writing "1" to the PRAD register, the clock selected with the PSAD register is input to the A/D converter.

When "0" is written, the clock is not input to the A/D converter.

At initial reset, this register is set to "0" (OFF).

#### ADRUN: 00FF82H•D7

Starts A/D conversion.

When "1" is written: Start A/D conversion

When "0" is written: Invalid Reading: Always "0"

By writing "1" to this register, the A/D converter starts A/D conversion of the channel selected by the CHS register, and stores the conversion result to the ADDR register.

#### CHS0, CHS1: 00FF82H•D0, D1

Selects an analog input channel.

Table 5.15.6.3 Selection of analog input channel

	<i>y</i> 8 . 1							
CHS1	CHS0	Input channel						
1	1	AD7						
1	0	AD6						
0	1	AD5						
0	0	AD4						

At initial reset, this register is set to "0" (AD4).

#### ADDR0-ADDR9: 00FF84H•D0, D1, 00FF83H

A/D conversion result is stored.
ADDR0 is the LSB and ADDR9 is the MSB.
ADDR0 and ADDR1 are assigned in D0 bit and D1 bit of the address 00FF84H. D2–D7 bits in this address are always "0" when being read.
At initial reset, data is undefined.

#### PADC0, PADC1: 00FF28H•D6, D7

Sets the priority level of the  $\ensuremath{\mathrm{A}/\mathrm{D}}$  conversion completion interrupt.

Table 5.15.6.4 shows the interrupt priority level which can be set by the PADC register.

Table 5.15.6.4 Interrupt priority level settings

PADC1	PADC0	Interrupt priority level
1	1	Level 3 (IRQ3)
1	0	Level 2 (IRQ2)
0	1	Level 1 (IRQ1)
0	0	Level 0 (None)

At initial reset, this register is set to "0" (level 0).

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#### EAD: 00FF2AH•D7

Enables or disables the A/D conversion completion interrupt generation to the CPU.

When "1" is written: Interrupt is enabled When "0" is written: Interrupt is disabled Reading: Valid

The EAD register is the interrupt enable register corresponding to the A/D conversion completion interrupt factor. When this register is set to "1", the interrupt is enabled, and when it is set to "0", the interrupt is disabled.

At initial reset, this register is set to "0" (interrupt is disabled).

#### FAD: 00FF2CH•D7

Indicates the generation of A/D conversion completion interrupt factor.

When "1" is read: Int. factor has generated When "0" is read: Int. factor has not generated

When "1" is written: Factor flag is reset

When "0" is written: Invalid

FAD is the interrupt factor flag corresponding to the A/D conversion completion interrupt. It is set to "1" when an A/D conversion is completed. At this point, if the corresponding interrupt enable register is set to "1" and the corresponding interrupt priority register is set to a higher level than the setting of the interrupt flags (I0 and I1), an interrupt is generated to the CPU.

Regardless of the interrupt enable register and interrupt priority register settings, the interrupt factor flag is set to "1" when the interrupt generation condition is met.

To accept the subsequent interrupt after an interrupt generation, it is necessary to re-set the interrupt flags (set the interrupt flag to a lower level than the level indicated by the interrupt priority registers, or execute the RETE instruction) and to reset the interrupt factor flag. The interrupt factor flag is reset to "0" by writing "1".

At initial reset, the FAD flag is reset to "0".

## 5.15.7 Programming notes

- (1) When the OSC3 oscillation circuit is made the clock source, it is necessary to turn the OSC3 oscillation ON, prior to using the A/D converter.
  - From the time the OSC3 oscillation circuit is turning ON until oscillation stabilizes, an interval of several msec to several 10 msec is necessary. Consequently, you should allow an adequate waiting time after turning the OSC3 oscillation circuit ON before starting the count of the programmable timer. (The oscillation start time will vary somewhat depending on the oscillator and on external parts. Refer to the oscillation start time example indicated in Chapter 8, "ELECTRICAL CHARACTERISTICS".)
  - At initial reset, OSC3 oscillation circuit is set to OFF status.
- (2) When SLEEP mode is set during A/D conversion, correct A/D conversion result cannot be obtained because the OSC3 oscillation circuit stops. Do not set in SLEEP mode during A/D conversion.
- (3) The input clock and analog input terminals should be set when the A/D converter stops. Changing in the A/D converter operation may cause a malfunction.
- (4) The clock division ratio (see Table 5.15.4.2) must be set so that the A/D conversion clock frequency is 1 MHz or less. Furthermore, the A/D conversion clock frequency should be changed according to the voltage to be used. Refer to Chapter 8, "ELECTRICAL CHARACTERISTICS".
- (5) To prevent malfunction, do not start A/D conversion (writing to the CHS register) when the A/D conversion clock is not being output from the prescaler, and do not turn the prescaler output clock off during A/D conversion.
- (6) If the CHS register selects an input channel which is not included in the analog input terminals set by the PAD register (the PAD register can select several terminals simultaneously), the A/D conversion does not result in a correct converted value.
- (7) During A/D conversion, do not operate the P1n terminals which are not used for analog inputs of the A/D converter (for input/output of digital signal). It affects the A/D conversion precision.

- (8) Since the P14–P17 terminals are shared with the analog comparator, the A/D converter and the analog comparator cannot be used simultaneously. When using the A/D converter, do not run the analog comparator.
- (9) Note that writing "1" to the A/D converter control bits (FF28H•D5, D4) may cause a malfunction.

# 5.16 Supply Voltage Detection (SVD) Circuit

## 5.16.1 Configuration of SVD circuit

The S1C88349 has a built-in supply voltage detection (SVD) circuit configured with a 4-bit successive approximation A/D converter.

The SVD circuit has 16 sampling levels (level 0–level 15) for supply voltage, and this can be controlled by software.

In addition, an initial reset signal can be generated when the supply voltage drops to level 0 or less. This is selected by the mask option.

Figure 5.16.1.1 shows the configuration of the SVD circuit.

# 5.16.2 Mask option

Initial reset by SVD circuit	
□ Not Use	
□ Use	

When the SVD circuit detects that supply voltage has dropped below level 0 four successive times (see Chapter 8, "ELECTRICAL CHARACTERISTICS"), it outputs an initial reset signal until the supply voltage has been restored to level 2. You can select whether this function is used or not by mask option.

## 5.16.3 Operation of SVD circuit

#### ■ Sampling control of the SVD circuit

The SVD circuit has two operation modes: continuous sampling and  $1/4~{\rm Hz}$  auto-sampling mode. Operation mode selection is done by the SVD control registers SVDON and SVDSP as shown in Table 5.16.3.1. When both bits of SVDON and SVDSP are set to "1", continuous sampling is selected.

Table 5.16.3.1 Correspondence between control register and operation mode

SVDON	SVDSP	Operating mode	
0	0	SVD circuit OFF	
0	1	1/4 Hz auto-sampling ON	
1	×	Continuous sampling ON	

In both operation modes, reading SVDON can confirm whether the SVD circuit is operating (BUSY) or on standby (READY); "1" indicates BUSY and "0" indicates READY.

When executing an SLP instruction while the SVD circuit is operating, the stop operation of the OSC1 oscillation circuit is kept waiting until the sampling is completed. The two bits of SVDON and SVDSP are automatically reset to "0" by hardware while waiting for completion of sampling.

To reduce current consumption, turn the SVD

#### **■** Detection result

The SVD circuit A/D converts the supply voltage (VDD–VSS) by 4-bit resolution and sets the result thereof into the SVD0–SVD3 register.

circuit OFF when it is not necessary.

The data in SVD0-SVD3 correspond to the detection levels as shown in Table 5.16.3.2 and the detection data is maintained until the next sampling.

For the correspondence between the detection level and the supply voltage, see Chapter 8, "ELECTRICAL CHARACTERISTICS".

An interval of 7.8 msec (fosc1 = 32.768 kHz) is required from the start of supply voltage sampling by the SVD circuit to completion by writing the result into SVD0–SVD3. Therefore, when reading SVD0–SVD3 before sampling is finished, the previous result will be read.

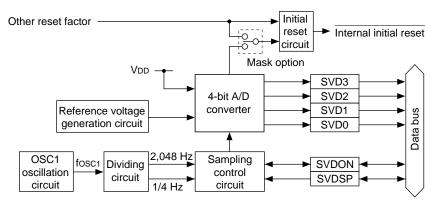


Fig. 5.16.1.1 Configuration of SVD circuit

Table 5.16.3.2 Supply voltage detection results

SVD3	SVD2	SVD1	SVD0	Detection level
1	1	1	1	Level 15
1	1	1	0	Level 14
1	1	0	1	Level 13
1	1	0	0	Level 12
1	0	1	1	Level 11
1	0	1	0	Level 10
1	0	0	1	Level 9
1	0	0	0	Level 8
0	1	1	1	Level 7
0	1	1	0	Level 6
0	1	0	1	Level 5
0	1	0	0	Level 4
0	0	1	1	Level 3
0	0	1	0	Level 2
0	0	0	1	Level 1
0	0	0	0	Level 0

#### ■ Timing of sampling

Next, we will explain the timing for two operation modes.

(1) Continuous sampling mode

This mode is selected when "1" is written to

SVDON and sampling of the supply voltage is
done continuously in 7.8 msec cycles.

The SVD circuit starts operation in synchronization with the internal 2,048 Hz signal and performs one sampling in 16 clock cycles. The sampling is done continuously without setting the standby time and the result is latched to SVD0–SVD3 in every 16 clock cycles. Cancellation of continuous sampling is done by writing "0" to SVDON. The SVD circuit maintains ON status until completion of sampling and then goes OFF.

After writing "0" to SVDON, SVDON reads "1" until the SVD circuit actually goes OFF. Figure 5.16.3.1 shows the timing chart of the continuous sampling.

#### (2) 1/4 Hz auto-sampling mode

This mode is selected when "0" is written to SVDON and "1" is written to SVDSP. In this case, supply voltage sampling is done in every 4 seconds.

The sampling time is 7.8 msec as in continuous sampling, and the result in SVD0–SVD3 is updated every 4 seconds.

Cancellation of 1/4 Hz auto-sampling is done by writing "0" to SVDSP. If the SVD circuit is sampling, SVD circuit waits until completion and then turns OFF. In addition, "1" is read from SVDON while the SVD circuit is sampling. Figure 5.16.3.2 shows the timing chart of the 1/4 Hz auto-sampling.

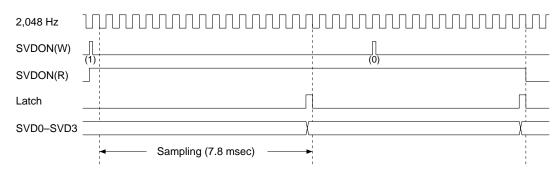


Fig. 5.16.3.1 Timing chart of continuous sampling

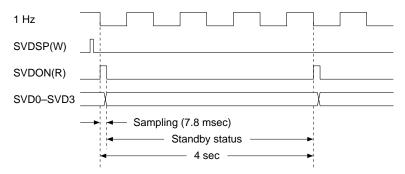


Fig. 5.16.3.2 Timing chart of 1/4 Hz auto-sampling

#### Reset function at low voltage detection

To avoid CPU runaway due to a supply voltage drop, an initial reset function when the supply voltage drops to level 0 or less can be selected by the mask option.

The SVD circuit shifts to continuous sampling status when it detects level 0 (SVD3–SVD0 = 0000B) four successive times. At this time, the internal initial reset signal is generated. The reset status continues until the supply voltage returns to level 2 (SVD3–SVD0 = 0010B) or higher.

When the reset status is canceled by the restoration of the supply voltage, the SVD circuit returns to its previous status. Continuous sampling status continuous in case of the previous status was continuous sampling. Then CPU starts the reset exception processing.

Figure 5.16.3.3 shows the timing chart of the initial reset signal generation. (Example when using 1/4 Hz auto-sampling.)

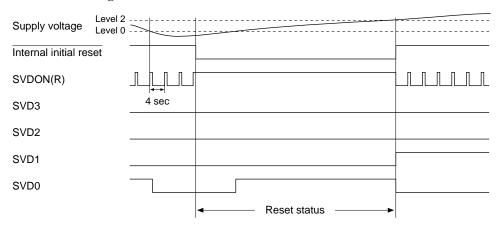


Fig. 5.16.3.3 Timing chart of the initial reset signal generation

# 5.16.4 Control of SVD circuit

Table 5.16.4.1 shows the SVD circuit control bits.

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF12	D7	_	_	ı	ı	_		Constantly "0" when
	D6	_	_	_	_	_		being read
	D5	SVDSP	SVD auto-sampling control	On	Off	0	R/W	These registers are
								reset to "0" when
	D4	SVDON	SVD continuous sampling control/status R	Busy	Ready	1→0*1	R/W	SLP instruction is
			W	On	Off	0		executed.
	D3	SVD3	SVD detection level			X	R	*2
	D2	SVD2	SVD3   SVD2   SVD1   SVD0   Detection level   1   Level 15			X	R	
	D1	SVD1	1 1 1 0 Level 14			X	R	
	D0	SVD0	: : : : : : 0 0 0 0 Level 0			X	R	

Table 5.16.4.1 SVD circuit control bits

<sup>\*1</sup> After initial reset, this status is set "1" until conclusion of hardware first sampling.

<sup>\*2</sup> Initial value is set according to the supply voltage detected at first hardware sampling after an initial reset. Until conclusion of first sampling, SVD0–SVD3 data are undefined.

#### SVDON: 00FF12H•D4

Controls the turning ON/OFF of the continuous sampling mode.

When "1" is written: Continuous sampling ON When "0" is written: Continuous sampling OFF

When "1" is read: BUSY
When "0" is read: READY

The continuous sampling mode goes ON when "1" is written to SVDON and goes OFF, when "0" is written

In the ON status, sampling of the supply voltage is done continuously in 7.8 msec cycles and the detection result is latched to SVD0–SVD3. SVDON can be read, and "1" indicates SVD circuit operation (BUSY) and "0" indicates standby (READY).

At initial reset and in the SLEEP status, SVDON is set to "0" (continuous sampling OFF/READY).

## SVDSP: 00FF12H•D5

Controls the turning ON/OFF of the 1/4 Hz autosampling mode.

When "1" is written: Auto-sampling ON When "0" is written: Auto-sampling OFF

Reading: Valid

The 1/4 Hz auto-sampling mode goes ON when "1" is written to SVDSP and goes OFF, when "0" is written.

In the ON status, sampling is done in every 4 seconds and "1" is read from SVDON during the actual sampling period (7.8 msec).

At initial reset and in the SLEEP status, SVDSP is set to "0" (auto-sampling OFF).

## SVD0-SVD3: 00FF12H•D0-D3

The detection result of the SVD is set.

The reading data correspond to the detection levels as shown in Table 5.16.4.2 and the data is maintained until the next sampling.

Table 5.16.4.2 Supply voltage detection results

SVD3	SVD2	SVD1	SVD0	Detection level
1	1	1	1	Level 15
1	1	1	0	Level 14
1	1	0	1	Level 13
1	1	0	0	Level 12
1	0	1	1	Level 11
1	0	1	0	Level 10
1	0	0	1	Level 9
1	0	0	0	Level 8
0	1	1	1	Level 7
0	1	1	0	Level 6
0	1	0	1	Level 5
0	1	0	0	Level 4
0	0	1	1	Level 3
0	0	1	0	Level 2
0	0	0	1	Level 1
0	0	0	0	Level 0

For the correspondence between the detection level and the supply voltage, see Chapter 8, "ELECTRICAL CHARACTERISTICS".

The initial value at initial reset is set according to the supply voltage detected at first sampling by hardware. Data of this bit is undefined until this sampling is completed.

# 5.16.5 Programming notes

- (1) To reduce current consumption, turn the SVD circuit OFF (SVDON = SVDSP = "0") when it is not necessary.
- (2) When executing an SLP instruction while the SVD circuit is operating, the stop operation of the OSC1 oscillation circuit is kept waiting until the sampling is completed. The two bits of SVDON and SVDSP are automatically reset to "0" by hardware while waiting for completion of sampling.

# 5.17 Interrupt and Standby Status

#### ■ Types of interrupts

7 systems and 16 types of interrupts have been provided for the S1C88349.

#### **External interrupt**

- K00–K07 input interrupt (2 types)
- K10 and K11 input interrupt (1 type)

#### Internal interrupt

- Clock timer interrupt (4 types)
- Stopwatch timer interrupt (3 types)
- Programmable timer interrupt (2 types)
- Serial interface interrupt (3 types)
- •A/D converter interrupt (1 type)

An interrupt factor flag that indicates the generation of an interrupt factor and an interrupt enable register that sets enable/disable for interrupt requests have been provided for each interrupt and interrupt generation can be optionally set for each factor.

In addition, an interrupt priority register has been provided for each system of interrupts and the priority of interrupt processing can be set to 3 levels in each system.

Figure 5.17.1 shows the configuration of the interrupt circuit.

Refer to the explanations of the respective peripheral circuits for details on each interrupt.

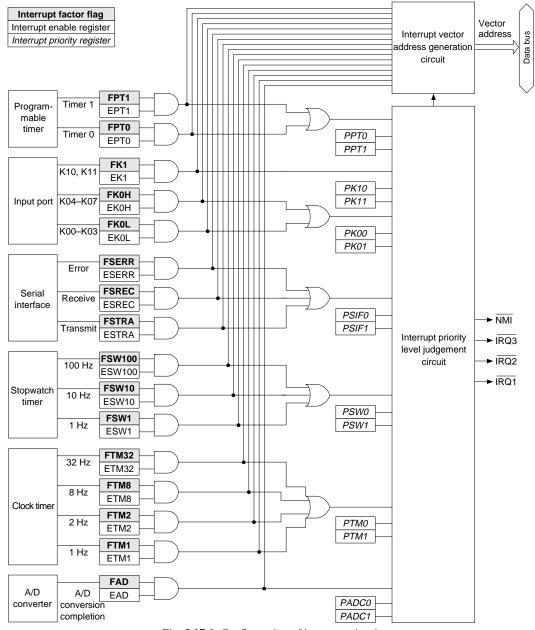


Fig. 5.17.1 Configuration of interrupt circuit

#### ■ HALT status

By executing the program's HALT instruction, the S1C88349 shifts to the HALT status.

Since CPU operation stops in the HALT status, power consumption can be reduced with only peripheral circuit operation.

Cancellation of the HALT status is done by initial reset or an optional interrupt request, and the CPU restarts program execution from an exception processing routine.

See the "S1C88 Core CPU Manual" for the HALT status and reactivation sequence.

#### **■ SLEEP status**

By executing the program's SLP instruction, the S1C88349 shifts to the SLEEP status.

Since the operation of the CPU and peripheral circuits stop completely in the SLEEP status, power consumption can be reduced even more than in the HALT status. Cancellation of the SLEEP status is done by initial reset or an input interrupt from the input port. The CPU reactivates after waiting 8,192/fosc1 seconds of oscillation stabilization time. At this time, the CPU restarts program execution from an exception processing routine (input interrupt routine).

Note: Since oscillation is unstable for a short time after reactivation from the SLEEP status, the wait time is not always 250 msec even when using the 32.768 kHz crystal oscillator for the OSC1 oscillation circuit.

# 5.17.1 Interrupt generation conditions

The interrupt factor flags that indicate the generation of their respective interrupt factors are provided for the previously indicated 7 systems and 16 types of interrupts and they will be set to "1" by the generation of a factor.

In addition, interrupt enable registers with a 1 to 1 correspondence to each of the interrupt factor flags are provided. An interrupt is enabled when "1" is written and interrupt is disabled when "0" is written.

The CPU manages the enable/disable of interrupt requests at the interrupt priority level. An interrupt priority register that sets the priority level is provided for each of the interrupts of the 7 systems and the CPU accepts only interrupts above the level that has been indicated with the interrupt flags (I0 and I1).

Consequently, the following three conditions are necessary for the CPU to accept the interrupt.

- (1) The interrupt factor flag has been set to "1" by generation of an interrupt factor.
- (2) The interrupt enable register corresponding to the above has been set to "1".
- (3) The interrupt priority register corresponding to the above has been set to a priority level higher than the interrupt flag (I0 and I1) setting.

The CPU initially samples the interrupt for the first op-code fetch cycle of each instruction. Thereupon, the CPU shifts to the exception processing when the above mentioned conditions have been established. See the "S1C88 Core CPU Manual" for the exception processing sequence.

# 5.17.2 Interrupt factor flag

Table 5.17.2.1 shows the correspondence between the factors generating an interrupt and the interrupt factor flags.

The corresponding interrupt factor flags are set to "1" by generation of the respective interrupt factors. The corresponding interrupt factor can be confirmed by reading the flags through software.

Table 5.17.2.1 Interrupt factors

Interrupt factor	Interrup	ot factor flag
Programmable timer 1 underflow	FPT1	(00FF25 D7)
Programmable timer 0 underflow	FPT0	(00FF25 D6)
Non matching of the K10 and K11 inputs and the input comparison registers KCP10 and KCP11	FK1	(00FF25 D5)
Non matching of the K04–K07 inputs and the input comparison registers KCP04–KCP07	FK0H	(00FF25 D4)
Non matching of the K00–K03 inputs and the input comparison registers KCP00–KCP03	FK0L	(00FF25 D3)
Serial interface receiving error (in asynchronous mode)	FSERR	(00FF25 D2)
Serial interface receiving completion	FSREC	(00FF25 D1)
Serial interface transmitting completion	FSTRA	(00FF25 D0)
Falling edge of the stopwatch timer 100 Hz signal	FSW100	(00FF24 D6)
Falling edge of the stopwatch timer 10 Hz signal	FSW10	(00FF24 D5)
Falling edge of the stopwatch timer 1 Hz signal	FSW1	(00FF24 D4)
Rising edge of the clock timer 32 Hz signal	FTM32	(00FF24 D3)
Rising edge of the clock timer 8 Hz signal	FTM8	(00FF24 D2)
Rising edge of the clock timer 2 Hz signal	FTM2	(00FF24 D1)
Rising edge of the clock timer 1 Hz signal	FTM1	(00FF24 D0)
A/D conversion completion	FAD	(00FF2C D7)

Interrupt factor flag that has been set to "1" is reset to "0" by writing "1".

At initial reset, the interrupt factor flags are reset to "0".

Note: When executing the RETE instruction without resetting the interrupt factor flag after an interrupt has been generated, the same interrupt will be generated. Consequently, the interrupt factor flag corresponding to that routine must be reset (writing "1") in the interrupt processing routine.

# 5.17.3 Interrupt enable register

The interrupt enable register has a 1 to 1 correspondence with each interrupt factor flag and enable/disable of interrupt requests can be set.

When "1" is written to the interrupt enable register, an interrupt request is enabled, and is disabled when "0" is written. This register also permits reading, thus making it possible to confirm that a status has been set.

At initial reset, the interrupt enable registers are set to "0" and shifts to the interrupt disable status. Table 5.17.3.1 shows the correspondence between the interrupt enable registers and the interrupt factor flags.

# 5.17.4 Interrupt priority register and interrupt priority level

The interrupt priority registers shown in Table 5.17.4.1 are set to each system of interrupts and the interrupt priority levels for the CPU can be set to the optional priority level (0–3). As a result, it is possible to have multiple interrupts that match the system's interrupt processing priority levels.

The interrupt priority level between each system can optionally be set to three levels by the interrupt priority register. However, when more than one system is set to the same priority level, they are processed according to the default priority level.

Table 5.17.4.2 Setting of interrupt priority level

P*1	P*0	Interrupt priority level		
1	1	Level 3 ( <del>IRQ3</del> )		
1	0	Level 2 ( <del>IRQ2</del> )		
0	1	Level 1 ( <del>IRQ1</del> )		
0	0	Level 0 (non)		

Table 5.17.3.1 Interrupt enable registers and interrupt factor flags

Interrupt	Interrupt factor flag		nable register	
Programmable timer 1	FPT1	(00FF25 D7)	EPT1	(00FF23 D7)
Programmable timer 0	FPT0	(00FF25 D6)	EPT0	(00FF23 D6)
K10 and K11 input	FK1	(00FF25 D5)	EK1	(00FF23 D5)
K04–K07 input	FK0H	(00FF25 D4)	EK0H	(00FF23 D4)
K00–K03 input	FK0L	(00FF25 D3)	EK0L	(00FF23 D3)
Serial interface receiving error	FSERR	(00FF25 D2)	ESERR	(00FF23 D2)
Serial interface receiving completion	FSREC	(00FF25 D1)	ESREC	(00FF23 D1)
Serial interface transmitting completion	FSTRA	(00FF25 D0)	ESTRA	(00FF23 D0)
Stopwatch timer 100 Hz	FSW100	(00FF24 D6)	ESW100	(00FF22 D6)
Stopwatch timer 10 Hz	FSW10	(00FF24 D5)	ESW10	(00FF22 D5)
Stopwatch timer 1 Hz	FSW1	(00FF24 D4)	ESW1	(00FF22 D4)
Clock timer 32 Hz	FTM32	(00FF24 D3)	ETM32	(00FF22 D3)
Clock timer 8 Hz	FTM8	(00FF24 D2)	ETM8	(00FF22 D2)
Clock timer 2 Hz	FTM2	(00FF24 D1)	ETM2	(00FF22 D1)
Clock timer 1 Hz	FTM1	(00FF24 D0)	ETM1	(00FF22 D0)
A/D conversion completion	FAD	(00FF2C D7)	EAD	(00FF2A D7)

*Table 5.17.4.1 Interrupt priority register* 

Interrupt	Interrupt priority register
Programmable timer interrupt	PPT0, PPT1 (00FF21 D2, D3)
K10 and K11 input interrupt	PK10, PK11 (00FF21 D0, D1)
K00-K07 input interrupt	PK00, PK01 (00FF20 D6, D7)
Serial interface interrupt	PSIF0, PSIF1 (00FF20 D4, D5)
Stopwatch timer interrupt	PSW0, PSW1 (00FF20 D2, D3)
Clock timer interrupt	PTM0, PTM1 (00FF20 D0, D1)
A/D converter interrupt	PADC0, PADC1 (00FF28 D6, D7)

At initial reset, the interrupt priority registers are all set to "0" and each interrupt is set to level 0. Furthermore, the priority levels in each system have been previously decided and they cannot be changed.

The CPU can mask each interrupt by setting the interrupt flags (I0 and I1). The relation between the interrupt priority level of each system and interrupt flags is shown in Table 5.17.4.3, and the CPU accepts only interrupts above the level indicated by the interrupt flags.

The  $\overline{\text{NMI}}$  (watchdog timer) that has level 4 priority, is always accepted regardless of the setting of the interrupt flags.

Table 5.17.4.3 Interrupt mask setting of CPU

l1	10	Acceptable interrupt
1	1	Level 4 (NMI)
1	0	Level 4, Level 3 (IRQ3)
0	1	Level 4, Level 3, Level 2 (IRQ2)
0	0	Level 4, Level 3, Level 2, Level 1 ( $\overline{IRQ1}$ )

After an interrupt has been accepted, the interrupt flags are written to the level of that interrupt. However, interrupt flags after an  $\overline{\text{NMI}}$  has been accepted are written to level 3 (I0 = I1 = "1").

Table 5.17.4.4 Interrupt flags after acceptance of interrupt

Accepted interrupt priority level		I1	10
Level 4	$(\overline{\mathrm{NMI}})$	1	1
Level 3	$(\overline{IRQ3})$	1	1
Level 2	$(\overline{IRQ2})$	1	0
Level 1	$(\overline{IRQ1})$	0	1

The set interrupt flags are reset to their original value on return from the interrupt processing routine. Consequently, multiple interrupts up to 3 levels can be controlled by the initial settings of the interrupt priority registers alone. Additional multiplexing can be realized by rewriting the interrupt flags and interrupt enable register in the interrupt processing routine.

Note: Beware. If the interrupt flags have been rewritten (set to lower priority) prior to resetting an interrupt factor flag after an interrupt has been generated, the same interrupt will be generated again.

## 5.17.5 Exception processing vectors

When the CPU accepts an interrupt request, it starts exception processing following completion of the instruction being executed. In exception processing, the following operations branch the program.

- In the minimum mode, the program counter (PC) and system condition flag (SC) are moved to stack and in the maximum mode, the code bank register (CB), PC and SC are moved.
- (2) The branch destination address is read from the exception processing vector corresponding to each exception processing (interrupt) factor and is placed in the PC.

An exception vector is 2 bytes of data in which the top address of each exception (interrupt) processing routine has been stored and the vector addresses correspond to the exception processing factors as shown in Table 5.17.5.1.

Table 5.17.5.1 Vector address and exception processing correspondence

Vector address	Exception processing factor	Priority
000000Н	Reset	High
000002H	Zero division	1
000004H	Watchdog timer (NMI)	
000006Н	Programmable timer 1 interrupt	
000008H	Programmable timer 0 interrupt	
00000AH	K10, K11 input interrupt	
00000CH	K04–K07 input interrupt	
00000EH	K00-K03 input interrupt	
000010H	Serial I/F error interrupt	
000012H	Serial I/F receiving complete interrupt	
000014H	Serial I/F transmitting complete interrupt	
000016Н	Stopwatch timer 100 Hz interrupt	
000018H	Stopwatch timer 10 Hz interrupt	
00001AH	Stopwatch timer 1 Hz interrupt	
00001CH	Clock timer 32 Hz interrupt	
00001EH	Clock timer 8 Hz interrupt	
000020H	Clock timer 2 Hz interrupt	
000022H	Clock timer 1 Hz interrupt	$\downarrow$
000024H	A/D converter interrupt	Low
000026Н	System reserved (cannot be used)	No
000028H		priority
:	Software interrupt	rating
0000FEH		raung

Note: An exception processing vector is fixed at 2 bytes, so it cannot specify a branch destination bank address. Consequently, to branch from multiple banks to a common exception processing routine, the top portion of an exception processing routine must be described within the common area (000000H–007FFFH).

# 5.17.6 Control of interrupt

Table 5.17.6.1 shows the interrupt control bits.

Table 5.17.6.1(a) Interrupt control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment	
00FF20	D7	PK01							
	D6	PK00	K00–K07 interrupt priority register	PK01 PK00		0-K07 interrupt priority register PK01 PK00		R/W	
	_	PSIF1		PSIF1 PSII	F0				
		PSIF0	Serial interface interrupt priority register	PSW1 PSW0 Priority PTM1 PTM0 level		0	R/W		
		PSW1		$\frac{11M1}{1}$ $\frac{11N}{1}$	Level 3				
		PSW0	Stopwatch timer interrupt priority register	1 0	1 0 Level 2 0 1 Level 1	0	R/W		
		PTM1		0 0	Level 0				
		PTM0	Clock timer interrupt priority register			0	R/W		
00FF21	D7	_	_	_	_	_			
001121	D6	_	_			_		Constantly "0" when	
	D5	_	_	_	_	_		being read	
	D4	_	_			_		being read	
		PPT1		PPT1 PPT	O Priority				
		PPT0	Programmable timer interrupt priority register	<u>PK11</u> <u>PK1</u>	0 level	0	R/W		
		PK11		$\begin{array}{cccc} 1 & 1 \\ 1 & 0 \end{array}$	Level 3 Level 2				
		PK10	K10 and K11 interrupt priority register	0 1	Level 1	0	R/W		
00FF22	D7	_		0 0	Level 0			"0" when being read	
001122	-	ESW/100	Stopwatch timer 100 Hz interrupt enable register	_	_	_		0 when being read	
		ESW100	Stopwatch timer 10 Hz interrupt enable register						
		ESW10	Stopwatch timer 1 Hz interrupt enable register						
		ETM32	Clock timer 32 Hz interrupt enable register	Interrupt	Interrupt	0	R/W		
		ETM8		enable	disable	0	K/W		
		ETM2	Clock timer 8 Hz interrupt enable register Clock timer 2 Hz interrupt enable register						
		ETM1							
00FF23	-	EPT1	Clock timer 1 Hz interrupt enable register						
00FF23		EPT0	Programmable timer 1 interrupt enable register						
	_	EK1	Programmable timer 0 interrupt enable register						
		EK0H	K10 and K11 interrupt enable register	<b>T</b>					
		EK0L	K04–K07 interrupt enable register	Interrupt	Interrupt	0	R/W		
		ESERR	K00–K03 interrupt enable register	enable	disable				
		ESREC	Serial I/F (error) interrupt enable register						
			Serial I/F (receiving) interrupt enable register						
005504		ESTRA	Serial I/F (transmitting) interrupt enable register						
00FF24	D7	-		-	-	_		"0" when being read	
			Stopwatch timer 100 Hz interrupt factor flag	(R)	(R)				
		FSW10	Stopwatch timer 10 Hz interrupt factor flag	Interrupt	No interrupt				
		FSW1	Stopwatch timer 1 Hz interrupt factor flag	factor is	factor is		D /337		
		FTM32	Clock timer 32 Hz interrupt factor flag	generated	generated	0	R/W		
		FTM8	Clock timer 8 Hz interrupt factor flag	(W)	(W)				
		FTM2	Clock timer 2 Hz interrupt factor flag	Reset	No operation				
005505		FTM1	Clock timer 1 Hz interrupt factor flag	(F)	(F)				
00FF25		FPT1	Programmable timer 1 interrupt factor flag	(R)	(R)				
		FPT0	Programmable timer 0 interrupt factor flag	Interrupt	No interrupt				
		FK1	K10 and K11 interrupt factor flag	factor is	factor is				
		FK0H	K04–K07 interrupt factor flag	generated	generated	0	R/W		
		FK0L	K00–K03 interrupt factor flag						
		FSERR	Serial I/F (error) interrupt factor flag	(W)	(W)				
		FSREC	Serial I/F (receiving) interrupt factor flag	Reset	No operation				
	D0	FSTRA	Serial I/F (transmitting) interrupt factor flag						

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Table 5.17.6.1(b) Interrupt control bits

Address	Bit	Name	Function	1	0	SR	R/W	Comment
00FF28	D7	PADC1	A/D converter interrupt priority register	PADC1 PAD	Level 3	0	R/W	
	D6	PADC0		$\begin{bmatrix} 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix}$	Level 2 Level 1 Level 0	0	R/W	
	D5	-	Reserved	Prohibited	-	0		Do not write "1".
	D4	_	Reserved	Prohibited	-	0		
	D3	_	_	-	-	_		
	D2	_	_	-	-	_		Constantly "0" when
	D1	_	_	-	-	_		being read
	D0	_	_	-	-	-		
00FF2A	D7	EAD	A/D converter interrupt enable register	Enable	Disable	0	R/W	
	D6	_	Reserved	-	-	0	R/W	
	D5	_	_	-	-	-		
	D4	_	_	-	_	-		
	D3	_	_	_	_	-		Constantly "0" when
	D2	_	_	_	-	_		being read
	D1	_	_	-	_	-		
	D0	_	_	-	_	-		
00FF2C	D7	FAD	A/D converter interrupt factor flag R)	Generated	Not generated	0	R/W	
			W	Reset	No operation			
	D6	_	Reserved	-	_	0	R/W	
	D5	-	_	-	-	_		
	D4	-	_	-	-	_		
	D3	<u>-                                    </u>	_	-	-	_		Constantly "0" when
	D2	_	_	-	_	_		being read
	D1	-	-	-	-	-		
	D0	_	_	-	_	_		

Refer to the explanations on the respective peripheral circuits for the setting content and control method for each bit.

## 5.17.7 Programming notes

- (1) When executing the RETE instruction without resetting the interrupt factor flag after an interrupt has been generated, the same interrupt will be generated. Consequently, the interrupt factor flag corresponding to that routine must be reset (writing "1") in the interrupt processing routine.
- (2) Beware. If the interrupt flags (I0 and I1) have been rewritten (set to lower priority) prior to resetting an interrupt factor flag after an interrupt has been generated, the same interrupt will be generated again.
- (3) An exception processing vector is fixed at 2 bytes, so it cannot specify a branch destination bank address. Consequently, to branch from multiple banks to a common exception processing routine, the front portion of an exception processing routine must be described within the common area (000000H–007FFFH).
- (4) Do not execute the SLP instruction for 2 msec after a  $\overline{NMI}$  interrupt has occurred (when fosc1 is 32.768 kHz).

# 6 SUMMARY OF NOTES

# 6.1 Notes for Low Current Consumption

The S1C88349 can turn circuits, which consume a large amount of power, ON or OFF by control registers.

You can reduce power consumption by creating a program that operates the minimum necessary circuits using these control registers.

Next, which circuit systems' operation can be controlled and their control registers (instructions) are explained. You should refer to these when programming.

See Chapter 8, "ELECTRICAL CHARACTERISTICS" for the current consumption.

Refer to "Programming notes" in each peripheral section for precautions of each peripheral circuit.

S1C88349 TECHNICAL MANUAL

Table 6.1.1 Circuit systems and control registers

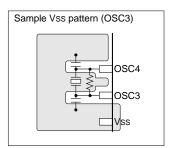
Circuit type	Control register (Instruction)	Status at time of initial resetting
CPU	HALT and SLP instructions	Operation status
Oscillation circuit	CLKCHG, OSCC	OSC1 clock (CLKCHG = "0")
		OSC3 oscillation OFF (OSCC = "0")
Operating mode	VDC0, VDC1	Normal mode (VDC0 = VDC1 = "0")
LCD controller	LCDC0, LCDC1	Drive OFF (LCDC0 = LCDC1 = "0")
SVD circuit	SVDON, SVDSP	OFF status (SVDON = SVDSP = "0")
Analog comparator	CMP0ON, CMP1ON	OFF status (CMP0ON = CMP1ON = "0")
A/D converter	PRAD, ADRUN	OFF status (PRAD = ADRUN = "0")

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# 6.2 Precautions on Mounting

#### <Oscillation Circuit>

- Oscillation characteristics change depending on conditions (board pattern, components used, etc.).
  - In particular, when a ceramic or crystal oscillator is used, use the oscillator manufacturer's recommended values for constants such as capacitance and resistance.
- Disturbances of the oscillation clock due to noise may cause a malfunction. Consider the following points to prevent this:
  - (1) Components which are connected to the OSC1, OSC2, OSC3 and OSC4 terminals, such as oscillators, resistors and capacitors, should be connected in the shortest line.
  - (2) As shown in the right hand figure, make a Vss pattern as large as possible at circumscription of the OSC1, OSC2, OSC3 and OSC4 terminals and the components connected to these terminals. Furthermore, do not use this Vss pattern for any purpose other than the oscillation system.



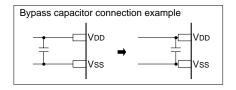
 In order to prevent unstable operation of the oscillation circuit due to current leak between OSC1/OSC3 and VDD, please keep enough distance between OSC1/OSC3 and VDD or other signals on the board pattern.

#### <Reset Circuit>

- The power-on reset signal which is input to the RESET terminal changes depending on conditions (power rise time, components used, board pattern, etc.).
  - Decide the time constant of the capacitor and resistor after enough tests have been completed with the application product.
  - When the built-in pull-up resistor is added to the RESET terminal by mask option, take into consideration dispersion of the resistance for setting the constant.
- In order to prevent any occurrences of unnecessary resetting caused by noise during operating, components such as capacitors and resistors should be connected to the RESET terminal in the shortest line.

## <Power Supply Circuit>

- Sudden power supply variation due to noise may cause malfunction. Consider the following points to prevent this:
  - The power supply should be connected to the VDD, Vss, AVDD, AVss and AVREF terminal with patterns as short and large as possible.
     In particular, the power supply for AVDD,
    - AVSS and AVREF affects A/D conversion precision.
  - (2) When connecting between the VDD and Vss terminals with a bypass capacitor, the terminals should be connected as short as possible.



(3) Components which are connected to the VD1, VC1-VC5 terminals, such as capacitors and resistors, should be connected in the shortest line. In particular, the VC1-VC5 voltages affect the display quality.

#### <A/D Converter>

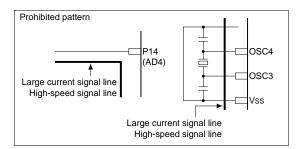
 When the A/D converter is not used, the power supply terminals for the analog system should be connected as shown below.

 $\begin{array}{ccc} \text{AVDD} & \rightarrow & \text{VDD} \\ \text{AVSS} & \rightarrow & \text{VSS} \\ \text{AVREF} & \rightarrow & \text{VSS} \end{array}$ 

input unit.

## <Arrangement of Signal Lines>

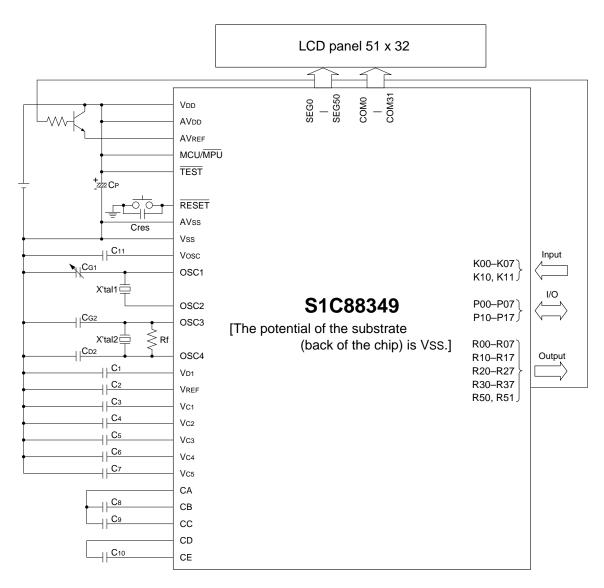
- In order to prevent generation of electromagnetic induction noise caused by mutual inductance, do not arrange a large current signal line near the circuits that are sensitive to noise such as the oscillation unit and analog input unit.
- When a signal line is parallel with a high-speed line in long distance or intersects a high-speed line, noise may generated by mutual interference between the signals and it may cause a malfunction.
   Do not arrange a high-speed signal line especially near circuits that are sensitive to noise such as the oscillation unit and analog



# <Precautions for Visible Radiation (when bare chip is mounted)>

- Visible radiation causes semiconductor devices to change the electrical characteristics. It may cause this IC to malfunction. When developing products which use this IC, consider the following precautions to prevent malfunctions caused by visible radiations.
  - Design the product and implement the IC on the board so that it is shielded from visible radiation in actual use.
  - (2) The inspection process of the product needs an environment that shields the IC from visible radiation.
  - (3) As well as the face of the IC, shield the back and side too.

# 7 BASIC EXTERNAL WIRING DIAGRAM



#### Recommended values for external parts

Symbol	Name	Recommended value
X'tal1	Crystal oscillator	32.768 kHz,
		CI (Max.) = $35 \text{ k}\Omega$
X'tal2	Crystal oscillator	4.9152 MHz
Rf	Feedback resistor	1 ΜΩ
CG1	Trimmer capacitor	5–25 pF
CG2	Gate capacitor	15-30 pF
CD2	Drain capacitor	15-30 pF
C1	Capacitor between Vss and VD1	0.1 μF
C2	Capacitor between Vss and VREF	0.1 μF

Symbol	Name	Recommended value
C3	Capacitor between Vss and Vc1	0.1 μF
C4	Capacitor between Vss and Vc2	0.1 μF
C5	Capacitor between Vss and Vc3	0.1 μF
C6	Capacitor between Vss and Vc4	0.1 μF
<b>C</b> 7	Capacitor between Vss and Vcs	0.1 μF
C8-C10	Booster capacitors	0.1 μF
C11	Capacitor between Vss and Vosc	0.1 μF
СР	Capacitor for power supply	3.3 μF
Cres	Capacitor for RESET terminal	0.47 μF

Note: The above table is simply an example, and is not guaranteed to work.

# 8 ELECTRICAL CHARACTERISTICS

# 8.1 Absolute Maximum Rating

(Mag		Λ	17
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				( * 55 -	- U V)
Item	Symbol	Condition	Rated value	Unit	Note
Power voltage	VDD		-0.3 to +7.0	V	
Liquid crystal power voltage	Vc5		-0.3 to +7.0	V	
Input voltage	VI		-0.3 to VDD + 0.3	V	
Output voltage	Vo		-0.3 to VDD + 0.3	V	1
High level output current	Іон	1 terminal	-5	mA	
		Total of all terminals	-20	mA	
Low level output current	Iol	1 terminal	5	mA	
		Total of all terminals	20	mA	
Permitted loss	PD		200	mW	2
Operating temperature	Topr		-40 to +85	°C	
Storage temperature	Tstg		-65 to +150	°C	
Soldering temperature / time	Tsol		260°C, 10sec (lead section)	_	

Note) 1 Case that to Nch open drain output by the mask option is included.

# 8.2 Recommended Operating Conditions

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Operating power voltage (Normal mode)	Vdd		2.4		5.5	V	
Operating power voltage (Low power mode)	Vdd		1.8		3.5	V	
Operating power voltage (High speed mode)	Vdd		3.5		5.5	V	
Operating frequency (Normal mode)	fosc1	$V_{DD} = 2.4 \text{ to } 5.5 \text{ V}$	30.000	32.768	80.000	kHz	1
	fosc3		0.03		4.2	MHz	1
Operating frequency (Low power mode)	fosc1	$V_{DD} = 1.8 \text{ to } 3.5 \text{ V}$	30.000	32.768	80.000	kHz	1
Operating frequency (High speed mode)	fosc1	$V_{DD} = 3.5 \text{ to } 5.5 \text{ V}$	30.000	32.768	80.000	kHz	1
	fosc3		0.03		8.2	MHz	1
Liquid crystal power voltage	VC5	$V_{C5} \ge V_{C4} \ge V_{C3} \ge V_{C2} \ge V_{C1} \ge V_{SS}$			7.0	V	2
Capacitor between VD1 and Vss	C1			0.1		μF	
Capacitor between VREF and VSS	C2			0.1		μF	3
Capacitor between VC1 and Vss	C3			0.1		μF	3
Capacitor between Vc2 and Vss	C4			0.1		μF	3
Capacitor between Vc3 and Vss	C5			0.1		μF	3
Capacitor between VC4 and Vss	C6			0.1		μF	3
Capacitor between Vc5 and Vss	<b>C</b> 7			0.1		μF	3
Capacitor between CA and CB	C8			0.1		μF	3
Capacitor between CA and CC	<b>C</b> 9			0.1		μF	3
Capacitor between CD and CE	C10			0.1		μF	3
Capacitor between Vosc and Vss	C11			0.1		μF	

Note) 1 When an external clock is input from the OSC1 terminal by the mask option, leave the OSC2 terminal open, and when an external clock is input from the OSC3 terminal, leave the OSC4 terminal open.

<sup>2</sup> In case of plastic package.

<sup>2</sup> When external power supply is selected by the mask option.

<sup>3</sup> When LCD drive power is not used, the capacitor is not necessary. In this case, leave the VC1 to VC5 and CA to CE terminals open.

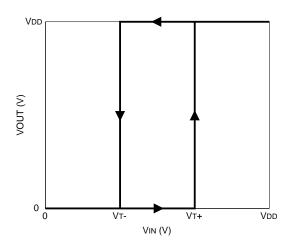
# 8.3 DC Characteristics

Unless otherwise specified: VDD = 1.8 to 5.5 V, Vss = 0 V, Ta = -40 to  $85^{\circ}C$ 

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
High level input voltage (1)	VIH1	Kxx, Pxx, MCU/MPU	0.8Vdd		Vdd	V	
Low level input voltage (1)	VIL1	Kxx, Pxx, MCU/MPU	0		0.2Vdd	V	
High level input voltage (2)	VIH2	OSC1, OSC3	1.6		Vdd	V	1
(Normal mode)							
High level input voltage (2)	VIH2	OSC1	1.0		Vdd	V	1
(Low power mode)							
High level input voltage (2)	VIH2	OSC1, OSC3	2.4		Vdd	V	1
(High speed mode)							
Low level input voltage (2)	VIL2	OSC1, OSC3	0		0.6	V	1
(Normal mode)							
Low level input voltage (2)	VIL2	OSC1	0		0.3	V	1
(Low power mode)							
Low level input voltage (2)	VIL2	OSC1, OSC3	0		0.9	V	1
(High speed mode)							
High level schmitt input voltage	V <sub>T+</sub>	RESET	0.5Vdd		0.9Vdd	V	
Low level schmitt input voltage	VT-	RESET	0.1Vdd		0.5Vdd	V	
High level output current	Іон	Pxx, Rxx, Voh = 0.9 Vdd			-0.5	mA	
Low level output current	Iol	Pxx, $Rxx$ , $Vol = 0.1 Vdd$	0.5			mA	
Input leak current	Ili	$Kxx, Pxx, \overline{RESET}, MCU/\overline{MPU}$	-1		1	μΑ	
Output leak current	ILO	Pxx, Rxx	-1		1	μΑ	
Input pull-up resistance	RIN	Kxx, Pxx, RESET, MCU/MPU	100		500	kΩ	2
Input terminal capacitance	CIN	Kxx, Pxx			15	pF	
		Vin = 0 V, $f = 1 MHz$ , $Ta = 25$ °C					
Segment/Common output current	Isegh	SEGxx, COMxx, Vsegh = Vc5-0.1 V			-5	μA	
	ISEGL	SEGxx, COMxx, Vsegl = 0.1 V	5			μA	

Note) 1 When external clock is selected by mask option.

<sup>2</sup> When addition of pull-up resistor is selected by mask option.



# 8.4 Analog Circuit Characteristics

## ■ LCD drive circuit

Unless otherwise specified: VDD = 1.8 to 5.5 V, Vss = 0 V, Ta = 25 °C, C1–C11 = 0.1  $\mu F$ 

Item	Symbol	Conditi	on	Min.	Тур.	Max.	Unit	Note
LCD drive voltage	Vc1	*1		0.19Vc5		0.23Vc5	V	
	VC2	*2		0.39Vc5		0.43Vc5	V	
	Vc3	*3		0.59Vc5		0.62Vc5	V	
	VC4	*4		0.78Vc5		0.82Vc5	V	
	VC5	*5	LCX = 0H		3.89		V	
	TYPE A		LCX = 1H		3.96		V	
	(4.5V)		LCX = 2H		4.04		V	
			LCX = 3H		4.11		V	
			LCX = 4H		4.18		V	
			LCX = 5H		4.26		V	
			LCX = 6H		4.34		V	
			LCX = 7H	Typ×0.94	4.42	Typ×1.06	V	
			LCX = 8H		4.50		V	
			LCX = 9H		4.58		V	
			LCX = AH		4.66		V	
			LCX = BH		4.74		V	
			LCX = CH		4.82		V	
			LCX = DH		4.90		V	
			LCX = EH		4.99		V	
			LCX = FH 5.08		V			
	VC5	*5	LCX = 0H		4.73		V	
	TYPE B		LCX = 1H		4.83		V	
	(5.5V)		LCX = 2H		4.92		V	
			LCX = 3H		5.02		V	
			LCX = 4H		5.11		V	
			LCX = 5H		5.21		V	
			LCX = 6H		5.30		V	
			LCX = 7H	Typ×0.94	5.40	Typ×1.06	V	
			LCX = 8H		5.50		V	
			LCX = 9H		5.60		V	
			LCX = AH	] [	5.70	] [	V	
			LCX = BH	] [	5.81	] [	V	
			LCX = CH	] [	5.93	] [	V	
			LCX = DH	] [	6.05	] [	V	
			LCX = EH	] [	6.17	] [	V	
			LCX = FH		6.29		V	

<sup>\*1</sup> Connects 1 MΩ load resistor between VDD and VC1 (VC1 has no capability to drive the load connected to Vss due to its property).

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<sup>\*2</sup> Connects 1 M $\Omega$  load resistor between Vss and Vc2.

<sup>\*3</sup> Connects 1 M $\Omega$  load resistor between Vss and Vc3.

<sup>\*4</sup> Connects 1 M $\Omega$  load resistor between Vss and Vc4.

<sup>\*5</sup> Connects 1 MΩ load resistor between Vss and Vcs.

#### ■ SVD circuit

Unless otherwise specified: VDD = 1.8 to 5.5 V, Vss = 0 V, Ta = 25°C

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
SVD voltage	Vsvd	Level 1 → Level 0		1.83		V	1
		Level 2 → Level 1		2.00		V	1
		Level 3 → Level 2		2.17			V
		Level 4 → Level 3		2.33		V	1
		Level 5 → Level 4		2.50		V	2
		Level 6 → Level 5		2.67	Typ×1.08	V	2
		Level 7 → Level 6		2.83		V	2
		Level 8 → Level 7	Typ×0.92	3.00		V	2
		Level 9 → Level 8		3.17		V	2
		Level 10 → Level 9		3.33	]	V	2
		Level 11 → Level 10		3.50	1	V	3
		Level 12 → Level 11		3.67	]	V	3
		Level 13 → Level 12		3.83		V	3
		Level 14 → Level 13		4.00		V	3
		Level 15 → Level 14		4.17		V	3

 $\overline{V_{SVD} (Level \ 0)} < \overline{V_{SVD} (Level \ 1)} < \overline{V_{SVD} (Level \ 2)} < \overline{V_{SVD} (Level \ 3)} < \overline{V_{SVD} (Level \ 4)} < \overline{V_{SVD} (Level \ 5)} < \overline{V_{SVD} (Level \ 6)} < \overline{V_{SVD} (Level \ 7)} < \overline{V_{SVD} (Level \$ 

 $< V_{SVD \, (Level \, 13)} < V_{SVD \, (Level \, 14)} < V_{SVD \, (Level \, 10)} < V_{SVD \, (Level \, 11)} < V_{SVD \, (Level \, 12)} < V_{SVD \, (Level \, 13)} < V_{SVD \, (Level \, 14)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 14)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 14)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 14)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \, 14)} < V_{SVD \, (Level \, 15)} < V_{SVD \, (Level \,$ 

- Note) 1 Low power operating mode only
  - 2 Low power operating mode or Normal operating mode
  - 3 Normal operating mode or High speed operating mode

#### ■ Analog comparator circuit

Unless otherwise specified: VDD = 1.8 to 5.5 V, Vss = 0 V, Ta = 25°C

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Analog comparator	VCMIP	Non-inverted input (CMPP)	0.7		VDD - 0.7	V	1
operating voltage input range	VCMIM	Inverted input (CMPM)	0.7		VDD - 0.7	V	1
Analog comparator offset voltage	VCMOF	$V_{CMIP} = 0.7 \text{ V}$ to $V_{DD} - 0.7 \text{ V}$			20	mV	1
		$V_{CMIM} = 0.7 \text{ V}$ to $V_{DD} - 0.7 \text{ V}$					
Analog comparator stability time	tcmp1				1	ms	2
Analog comparator response time	tcmp2	$V_{CMIP} = 0.7 \text{ V}$ to $V_{DD} - 0.7 \text{ V}$			2	ms	1
		$V_{CMIM} = 0.7 \text{ V}$ to $V_{DD} - 0.7 \text{ V}$					3
		$V_{CMIP} = V_{CMIM} \pm 0.025 V$					

- Note) 1 When "no pull-up resistor" (for comparator input) is selected by mask option.
  - 2 Stability time is the time from turning the circuit ON until the circuit is stabilized.
  - 3 Response time is the time that the output result responds to the input signal.

## ■ A/D converter circuit

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Zero-scale error	Ezs	VDD = AVDD = AVREF = 2.4  to  5.5  V,	-1.5		+1.5	LSB	
Full-scale error	Efs	ADCLK = 1 MHz, Ta = 25°C	-1.5		+1.5	LSB	
Non-linearity error	El		-5		+5	LSB	
Total error	Et		-5		+5	LSB	
A/D converter	IAD	VDD = AVDD = AVREF = 3.0 V,		70	130	μΑ	1
current consumption		ADCLK = 1 MHz, Ta = 25°C					
		AVREF and ADCLK divider current					
		not included					
		$V_{DD} = AV_{DD} = AV_{REF} = 5.0 \text{ V},$		250	350	μΑ	1
		ADCLK = 1 MHz, Ta = 25°C					
		AVREF and ADCLK divider current					
		not included					
A/D conversion clock	f	VDD = AVDD = AVREF = 2.4  to  5.5  V,	0.03		1	MHz	2
		Ta = 25°C					

\* Zero-scale error: Ezs = deviation from the ideal value at zero point

\* Full-scale error: Efs = deviation from the ideal value at the full scale point

\* Non-linearity error: El = deviation of the real conversion curve from the end point line

\* Total error: Et = max (Ezs, Efs, Eabs), Eabs = deviation from the ideal line (including quantization error)

Note) 1 Average current during conversion

2 See Table 5.15.4.2 in "5.15 A/D Converter" for setting the A/D conversion clock.

# 8.5 Power Current Consumption

Unless otherwise specified: VDD = Within the operating voltage in each operating mode, Vss = 0 V, Ta = 25°C,

OSC1 = 32.768 kHz crystal oscillation, CG = 25 pF, C1-C11 = 0.1  $\mu F$ , No panel load

Item	Symbol	Condition	Mir	n.	Тур.	Max.	Unit	Note
Power current	Iddi	In SLEEP status *	1		0.3	1.5	μΑ	
(Normal mode)	IDD2	In HALT status *	2		1.5	4	μΑ	
	IDD3	CPU is in operating (32.768 kHz) *	3		9	15	μΑ	
	IDD4	CPU is in operating (4 MHz) *	4		1.1	1.4	mA	
Power current	Iddi	In SLEEP status *	1		0.2	1	μΑ	
(Low power mode)	IDD2	In HALT status *	2		1	3	μΑ	
	IDD3	CPU is in operating (32.768 kHz) *	3		5	7	μΑ	
Power current	Iddi	In SLEEP status *	1		0.6	3	μΑ	
(High speed mode)	IDD2	In HALT status *	2		2	6	μΑ	
	IDD3	CPU is in operating (32.768 kHz) *	3		13	20	μΑ	
	IDD4	CPU is in operating (8 MHz) *	4		3.7	4.8	mA	
LCD drive circuit current	ILCDN				3	5	μΑ	
SVD circuit current	Isvdn	$V_{DD} = 3 V$			30	60	μΑ	1
Analog comparator circuit current	ICMP1	CMPXDT = "1"			35	60	μΑ	
	ICMP2	CMPXDT = "0"			2	4	μΑ	

<sup>\*1</sup> OSC1: Stop, OSC3: Stop,

Clock timer: Stop, Others: Stop status Clock timer: Operating, Others: Stop status

CPU, ROM, RAM: SLEEP status, CPU, ROM, RAM: HALT status,

<sup>\*2</sup> OSC1: Oscillating, OSC3: Stop, \*3 OSC1: Oscillating, OSC3: Stop,

CPU, ROM, RAM: Operating in 32.768 kHz, Clock timer: Operating, Others: Stop status

<sup>\*4</sup> OSC1: Oscillating, OSC3: Oscillating, CPU, ROM, RAM: Operating in 1 MHz, Clock timer: Operating, Others: Stop status

Note) 1 The value in x V can be found by the following expression: ISVDN (VDD = x V) = ( $x \times 60$ ) - 150 (Max. value)

# 8.6 AC Characteristics

# ■ Operating range

Condition: VDD = Within the operating voltage in each operating mode, Vss = 0 V, Ta = -40 to 85°C

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Operating frequency (Normal mode)	fosc1	VDD = 2.4 to 5.5 V	30.000	32.768	80.000	kHz	
	fosc3		0.03		4.2	MHz	
Operating frequency (Low power mode)	fosc1	VDD = 1.8 to 3.5 V	30.000	32.768	80.000	kHz	
Operating frequency (High speed mode)	fosc1	V <sub>DD</sub> = 3.5 to 5.5 V	30.000	32.768	80.000	kHz	
	fosc3		0.03		8.2	MHz	
Instruction execution time	tcy	1-cycle instruction	25	61	67	μs	
(during operation with OSC1 clock)		2-cycle instruction	50	122	133	μs	
		3-cycle instruction	75	183	200	μs	
		4-cycle instruction	100	244	267	μs	
		5-cycle instruction	125	305	333	μs	
		6-cycle instruction	150 366 400		μs		
Instruction execution time	tcy	1-cycle instruction	0.5		66.7	μs	
Normal mode		2-cycle instruction	1.0		133.3	μs	
(during operation with OSC3 clock)		3-cycle instruction	1.4		200.0	μs	
		4-cycle instruction	1.9		266.7	μs	
		5-cycle instruction	2.4		333.3	μs	
		6-cycle instruction	2.9		400.0	μs	
Instruction execution time	tcy	1-cycle instruction	0.2		66.7	μs	
High speed mode		2-cycle instruction	0.5		133.3	μs	
(during operation with OSC3 clock)		3-cycle instruction	0.7		200.0	μs	
		4-cycle instruction	1.0		266.7	μs	
		5-cycle instruction	1.2		333.3	μs	
		6-cycle instruction	1.5		400.0	μs	

#### ■ External memory access

#### Read cycle (Normal operating mode)

Condition:  $V_{DD} = 2.4 \text{ to } 5.5 \text{ V}, V_{SS} = 0 \text{ V}, T_{a} = 25^{\circ}\text{C}, V_{IH1} = 0.8 V_{DD}, V_{IL1} = 0.2 V_{DD}, V_{IH2} = 1.6 \text{ V}, V_{IL2} = 0.6 \text{ V}, V_{IL2} = 0.6 \text{ V}, V_{IL3} = 0.2 V_{DD}, V_{IL4} = 0.2 V_{DD}, V_{IL4}$ 

VOH = 0.8VDD, VOL = 0.2VDD, CL = 100 pF (load capacitance)

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Address set-up time in read cycle	tras	tc+tl-100+n•tc/2			ns	1
Address hold time in read cycle	trah	th-80			ns	
Read signal pulse width	trp	tc-20+n•tc/2			ns	1
Data input set-up time in read cycle	trds	300			ns	2
Data input hold time in read cycle	trdh	0			ns	

Note) 1 Substitute the number of states for wait insertion in n.

2 Insert a wait cycle if the connected device does not satisfy the trds condition. (See Section 3.6.5, "WAIT control".)

#### · Read cycle (High speed operating mode)

Condition:  $V_{DD} = 3.5 \text{ to } 5.5 \text{ V}, V_{SS} = 0 \text{ V}, T_{a} = 25^{\circ}\text{C}, V_{IH1} = 0.8 V_{DD}, V_{IL1} = 0.2 V_{DD}, V_{IH2} = 2.4 \text{ V}, V_{IL2} = 0.9 \text{ V}, V_{IL2} = 0.9 \text{ V}, V_{IL3} = 0.9 \text{ V}, V_{IL4} = 0.9 \text{ V}, V_{IL5} = 0.9 \text{ V}, V_{IL5}$ 

VOH = 0.8VDD, VOL = 0.2VDD, CL = 100 pF (load capacitance)

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Address set-up time in read cycle	tras	tc+tl-50+n•tc/2			ns	1
Address hold time in read cycle	trah	th-40			ns	
Read signal pulse width	trp	tc-10+n•tc/2			ns	1
Data input set-up time in read cycle	trds	150			ns	2
Data input hold time in read cycle	trdh	0			ns	

Note) 1 Substitute the number of states for wait insertion in n.

#### Read cycle (Low power operating mode)

Condition: VDD = 1.8 to 3.5 V, VSS = 0 V,  $Ta = 25^{\circ}\text{C}$ , VIH1 = 0.8 VDD, VIL1 = 0.2 VDD, VIH2 = 1.0 V, VIL2 = 0.3 V,

 $V_{OH} = 0.8V_{DD}$ ,  $V_{OL} = 0.2V_{DD}$ ,  $C_L = 100 pF$  (load capacitance)

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Address set-up time in read cycle	tras	15			μs	
Address hold time in read cycle	trah	5			μs	
Read signal pulse width	trp	10			μs	
Data input set-up time in read cycle	trds	10			μs	
Data input hold time in read cycle	trdh	0			μs	

#### Write cycle (Normal operating mode)

 $Condition: V \text{DD} = 2.4 \text{ to } 5.5 \text{ V}, \text{ Vss} = 0 \text{ V}, \text{ Ta} = 25^{\circ}\text{C}, \text{ Vihi} = 0.8 \text{ Vdd}, \text{ Vill} = 0.2 \text{ Vdd}, \text{ Vihi} = 1.6 \text{ V}, \text{ Vill} = 0.6 \text{ V}, \text{ Vill} = 0.6 \text{ Vdd}, \text{ Vill} = 0.6 \text{ Vd$ 

VOH = 0.8VDD, VOL = 0.2VDD, CL = 100 pF (load capacitance)

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Address set-up time in write cycle	twas	tc-180			ns	
Address hold time in write cycle	twah	th-80			ns	
Write signal pulse width	twp	tl-40+n•tc/2			ns	1
Data output set-up time in write cycle	twds	tc-180+n•tc/2			ns	1
Data output hold time in write cycle	twdh	th-80		th+80	ns	

Note) 1 Substitute the number of states for wait insertion in n.

#### Write cycle (High speed operating mode)

 $Condition: V_{DD} = 3.5 \text{ to } 5.5 \text{ V}, V_{SS} = 0 \text{ V}, T_{a} = 25^{\circ}\text{C}, V_{IH1} = 0.8 V_{DD}, V_{IL1} = 0.2 V_{DD}, V_{IH2} = 2.4 \text{ V}, V_{IL2} = 0.9 \text{ V}, V_{IL3} = 0.0 \text{ V}, V_{IL4} = 0.0 \text{ V}, V_{IL5} = 0.0 \text{ V}, V_{IL5}$ 

VOH = 0.8VDD, VOL = 0.2VDD, CL = 100 pF (load capacitance)

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Address set-up time in write cycle	twas	tc-90			ns	
Address hold time in write cycle	twah	th-40			ns	
Write signal pulse width	twp	tl-20+n•tc/2			ns	1
Data output set-up time in write cycle	twds	tc-90+n•tc/2			ns	1
Data output hold time in write cycle	twdh	th-40		th+40	ns	

Note) 1 Substitute the number of states for wait insertion in n.

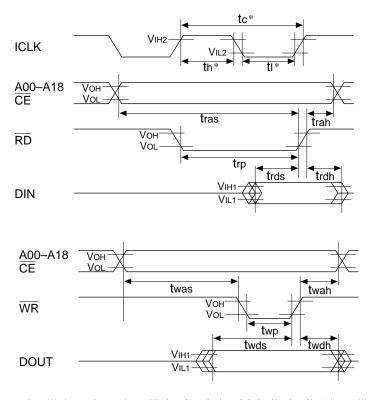
<sup>2</sup> Insert a wait cycle if the connected device does not satisfy the trds condition. (See Section 3.6.5, "WAIT control".)

## • Write cycle (Low power operating mode)

Condition:  $V_{DD} = 1.8 \text{ to } 3.5 \text{ V}, V_{SS} = 0 \text{ V}, T_a = 25^{\circ}\text{C}, V_{IH1} = 0.8 V_{DD}, V_{IL1} = 0.2 V_{DD}, V_{IH2} = 1.0 \text{ V}, V_{IL2} = 0.3 \text{ V}, V_{IL2} = 0.3 \text{ V}, V_{IL3} = 0.2 V_{DD}, V_{IH4} = 1.0 \text{ V}, V_{IL3} = 0.0 \text{ V}, V_{IL4} = 0.0 \text{ V}, V_{IL5} =$ 

VOH = 0.8VDD, VOL = 0.2VDD, CL = 100 pF (load capacitance)

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Address set-up time in write cycle	twas	10			μs	
Address hold time in write cycle	twah	5			μs	
Write signal pulse width	twp	5			μs	
Data output set-up time in write cycle	twds	10			μs	
Data output hold time in write cycle	twdh	5		20	μs	



- \* In the case of crystal oscillation and ceramic oscillation:  $th = 0.5tc \pm 0.05tc$ , tl = tc th (1/tc: oscillation frequency)
- \* In the case of CR oscillation: th = 0.5tc  $\pm 0.10$ tc, tl = tc th (1/tc: oscillation frequency)

#### ■ Serial interface

#### • Clock synchronous master mode (Normal operating mode)

 $Condition: \ V \text{DD} = 2.4 \text{ to } 5.5 \text{ V}, \ V \text{SS} = 0 \text{ V}, \ T \text{a} = 25 ^{\circ} \text{C}, \ V \text{IH} \\ \text{I} = 0.8 \text{VDD}, \ V \text{IL} \\ \text{I} = 0.2 \text{VDD}, \ V \text{OH} = 0.8 \text{VDD}, \ V \text{OL} = 0.2 \text{VDD}, \ V \text{OH} = 0.8 \text{$ 

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Transmitting data output delay time	tsmd			200	ns	
Receiving data input set-up time	tsms	500			ns	
Receiving data input hold time	tsmh	200			ns	

#### • Clock synchronous master mode (High speed operating mode)

 $Condition: V \text{DD} = 3.5 \text{ to } 5.5 \text{ V}, \text{Vss} = 0 \text{ V}, \text{Ta} = 25^{\circ}\text{C}, \text{Vihi} = 0.8 \text{Vdd}, \text{Vili} = 0.2 \text{Vdd}, \text{Voh} = 0.8 \text{Vdd}, \text{Vol} = 0.2 \text{Vdd}$ 

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Transmitting data output delay time	tsmd			100	ns	
Receiving data input set-up time	tsms	250			ns	
Receiving data input hold time	tsmh	100			ns	

## • Clock synchronous master mode (Low power operating mode)

 $Condition: V \text{DD} = 1.8 \text{ to } 3.5 \text{ V}, \text{Vss} = 0 \text{ V}, \text{Ta} = 25^{\circ}\text{C}, \text{Vihi} = 0.8 \text{Vdd}, \text{Vili} = 0.2 \text{Vdd}, \text{Voh} = 0.8 \text{Vdd}, \text{Vol} = 0.2 \text{Vdd}$ 

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Transmitting data output delay time	tsmd			5	μs	
Receiving data input set-up time	tsms	10			μs	
Receiving data input hold time	tsmh	5			μs	

#### • Clock synchronous slave mode (Normal operating mode)

 $Condition: \ V \text{DD} = 2.4 \text{ to } 5.5 \text{ V}, \ V \text{SS} = 0 \text{ V}, \ T \text{a} = 25 ^{\circ} \text{C}, \ V \text{IH} \\ 1 = 0.8 \text{VDD}, \ V \text{IL} \\ 1 = 0.2 \text{VDD}, \ V \text{OH} \\ = 0.8 \text{VDD}$ 

	/		,	,		
Item	Symbol	Min.	Тур.	Max.	Unit	Note
Transmitting data output delay time	tssd			500	ns	
Receiving data input set-up time	tsss	200			ns	
Receiving data input hold time	tssh	200			ns	

#### Clock synchronous slave mode (High speed operating mode)

Condition: VDD = 3.5 to 5.5 V, Vss = 0 V, Ta = 25°C, VIH1 = 0.8VDD, VIL1 = 0.2VDD, VOH = 0.8VDD, VOL = 0.2VDD

condition (BB Cic to Cic 1, 185 C 1, 18	20 0, 1111	. 0.0 . DD, . 121	0.2 . DD, . 011 0.0	DD, 102 012 1 DD		
Item	Symbol	Min.	Тур.	Max.	Unit	Note
Transmitting data output delay time	tssd			250	ns	
Receiving data input set-up time	tsss	100			ns	
Receiving data input hold time	tssh	100			ns	

#### • Clock synchronous slave mode (Low power operating mode)

Condition: VDD = 1.8 to 3.5 V, Vss = 0 V, Ta = 25°C, VIH1 = 0.8 VDD, VIL1 = 0.2 VDD, VOH = 0.8 VDD, VOL = 0.2 VDD

condition (BB 110 to 515 ), (BB 0 ), (Ta 1	-c, , ,	· O.O.BB, TEL	0.2 . DB, . 011 010 .	DD, 10E 012 1 DD		
Item	Symbol	Min.	Тур.	Max.	Unit	Note
Transmitting data output delay time	tssd			10	μs	
Receiving data input set-up time	tsss	5			μs	
Receiving data input hold time	tssh	5			μs	

#### • Asynchronous system (All operating mode)

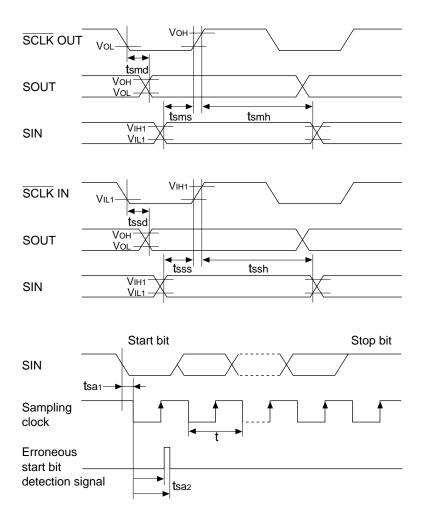
Condition:  $VDD = 1.8 \text{ to } 5.5 \text{ V}, Vss = 0 \text{ V}, Ta = 25^{\circ}\text{C}$ 

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Start bit detection error time	tsaı	0		t/16	s	1
Erroneous start bit detection range time	tsa2	9t/16		10t/16	s	2

- Note) 1 Start bit detection error time is a logical delay time from inputting the start bit until internal sampling begins operating.

  (Time as far as AC is excluded.)
  - 2 Erroneous start bit detection range time is a logical range to detect whether a LOW level (start bit) has been input again after a start bit has been detected and the internal sampling clock has started.

When a HIGH level is detected, the start bit detection circuit is reset and goes into a wait status until the next start bit. (Time as far as AC is excluded.)



## ■ Input clock

# • OSC1, OSC3 external clock (Normal operating mode)

Condition: VDD = 2.4 to 5.5 V, VSS = 0 V,  $Ta = -40 \text{ to } 85^{\circ}\text{C}$ , VIH2 = 1.6 V, VIL2 = 0.6 V

Item		Symbol	Min.	Тур.	Max.	Unit	Note
OSC1 input clock time	Cycle time	toicy	20		32	μs	
	"H" pulse width	toth	10		16	μs	
	"L" pulse width	toil	10		16	μs	
OSC3 input clock time	Cycle time	toscy	250		32,000	ns	
	"H" pulse width	to3h	125		16,000	ns	
	"L" pulse width	to3l	125		16,000	ns	
Input clock rising time		tosr			25	ns	
Input clock falling time		tosf			25	ns	

## • OSC1, OSC3 external clock (High speed operating mode)

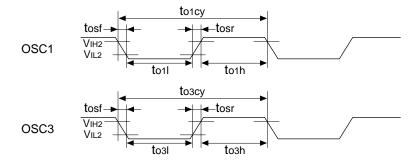
Condition:  $VDD = 3.5 \text{ to } 5.5 \text{ V}, VSS = 0 \text{ V}, Ta = -40 \text{ to } 85^{\circ}\text{C}, VIH2 = 2.4 \text{ V}, VIL2 = 0.9 \text{ V}$ 

Item		Symbol	Min.	Тур.	Max.	Unit	Note
OSC1 input clock time	Cycle time	toicy	20		32	μs	
	"H" pulse width	toth	10		16	μs	
	"L" pulse width	toil	10		16	μs	
OSC3 input clock time	Cycle time	toscy	125		32,000	ns	
	"H" pulse width	to3h	62.5		16,000	ns	
	"L" pulse width	to3l	62.5		16,000	ns	
Input clock rising time		tosr			25	ns	
Input clock falling time		tosf			25	ns	

## • OSC1, OSC3 external clock (Low power operating mode)

Condition: VDD = 1.8 to 3.5 V, Vss = 0 V, Ta = -40 to 85°C, VIH2 = 1.0 V, VIL2 = 0.3 V

Item		Symbol	Min.	Тур.	Max.	Unit	Note
OSC1 input clock time	Cycle time	toicy	20		32	μs	
	"H" pulse width	toth	10		16	μs	
	"L" pulse width	toil	10		16	μs	
Input clock rising time		tosr			25	ns	
Input clock falling time		tosf			25	ns	



# • SCLK, EVIN input clock (Normal operating mode)

Condition: VDD = 2.4 to 5.5 V, Vss = 0 V, Ta = 25°C, VIHI = 0.8VDD, VILI = 0.2VDD

Item	Item		Min.	Тур.	Max.	Unit	Note
SCLK input clock time	Cycle time	tsccy	4			μs	
	"H" pulse width	tsch	2			μs	
	"L" pulse width	tscl	2			μs	
EVIN input clock time	Cycle time	tevcy	64 / fosc1			s	
(With noise rejector)	"H" pulse width	tevh	32 / fosc1			S	
	"L" pulse width	tevl	32 / fosc1			S	
EVIN input clock time	Cycle time	tevcy	4			μs	
(Without noise rejector)	"H" pulse width	tevh	2			μs	
	"L" pulse width	tevl	2			μs	
Input clock rising time		tckr			25	ns	
Input clock falling time		tckf			25	ns	

# • SCLK, EVIN input clock (High speed operating mode)

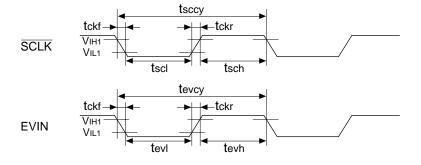
Condition: VDD = 3.5 to 5.5 V, Vss = 0 V, Ta = 25°C, VIHI = 0.8VDD, VILI = 0.2VDD

Item		Symbol	Min.	Тур.	Max.	Unit	Note
SCLK input clock time	Cycle time	tsccy	2			μs	
	"H" pulse width	tsch	1			μs	
	"L" pulse width	tscl	1			μs	
EVIN input clock time	Cycle time	tevcy	64 / fosc1			s	
(With noise rejector)	"H" pulse width	tevh	32 / fosc1			s	
	"L" pulse width	tevl	32 / fosc1			s	
EVIN input clock time	Cycle time	tevcy	2			μs	
(Without noise rejector)	"H" pulse width	tevh	1			μs	
	"L" pulse width	tevl	1			μs	
Input clock rising time		tckr			25	ns	
Input clock falling time		tckf			25	ns	

# • SCLK, EVIN input clock (Low power operating mode)

Condition: VDD = 1.8 to 3.5 V, VSS = 0 V, Ta = 25°C, VIHI = 0.8VDD, VILI = 0.2VDD

Item		Symbol	Min.	Тур.	Max.	Unit	Note
SCLK input clock time	Cycle time	tsccy	100			μs	
	"H" pulse width	tsch	50			μs	
	"L" pulse width	tscl	50			μs	
EVIN input clock time	Cycle time	tevcy	64 / fosc1			s	
(With noise rejector)	"H" pulse width	tevh	32 / fosc1			s	
	"L" pulse width	tevl	32 / fosc1			s	
EVIN input clock time	Cycle time	tevcy	100			μs	
(Without noise rejector)	"H" pulse width	tevh	50			μs	
	"L" pulse width	tevl	50			μs	
Input clock rising time		tckr			25	ns	
Input clock falling time		tckf			25	ns	



## • RESET input clock (All operating mode)

Condition: VDD = 1.8 to 5.5 V, VSS = 0 V,  $Ta = 25^{\circ}\text{C}$ , VIH = 0.5 VDD, VIL = 0.1 VDD

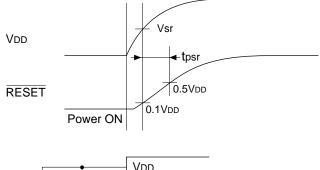
Item	Symbol	Min.	Тур.	Max.	Unit	Note
RESET input time	tsr	100			μs	

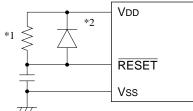


#### ■ Power ON reset

Condition:  $VDD = 1.8 \text{ to } 5.5 \text{ V}, Vss = 0 \text{ V}, Ta = 25^{\circ}C$ 

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Operating power voltage	Vsr	2.4			V	
RESET input time	tpsr	10			ms	





- \*1 When the built-in pull up resistor is not used.
- \*2 Because the potential of the RESET terminal not reached VDD level or higher.

#### Operating mode switching

Condition:  $VDD = 1.8 \text{ to } 5.5 \text{ V}, VSS = 0 \text{ V}, Ta = 25^{\circ}\text{C}$ 

Item	Symbol	Min.	Тур.	Max.	Unit	Note
Stabilization time	tvdc	5			ms	1

Note) 1 Stabilization time is the time from switching on the operating mode until operating mode is stabilized. For example, when turning the OSC3 oscillation circuit on, stabilization time is needed after the operating mode is switched on.

# 8.7 Oscillation Characteristics

Oscillation characteristics change depending on conditions (board pattern, components used, etc.). Use the following characteristics as reference values. In particular, when a ceramic oscillator is used for OSC3, use the oscillator manufacturer's recommended values for constants such as capacitance and resistance. The oscillation start time is important because it becomes the wait time when OSC3 clock is used. (If OSC3 is used as CPU clock before oscillation stabilizes, the CPU may malfunction.)

#### ■ OSC1 (Crystal)

Unless otherwise specified: VDD = Within the operating voltage in each operating mode, Vss = 0 V, Ta = 25°C,

Crystal oscillator = Q12C2\*, CG1 = 25 pF, CD1 = Built-in

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Oscillation start time	tsta				3	s	
External gate capacitance	CG1	Including board capacitance	5		25	pF	1
Built-in gate capacitance	CG1	In case of the chip		17		pF	2
Built-in drain capacitance	CD1	In case of the chip		10		pF	
Frequency/IC deviation	∂f/∂IC	V <sub>DD</sub> = constant	-10		10	ppm	
Frequency/power voltage deviation	∂f/∂V				1	ppm/V	
Frequency adjustment range	∂f/∂Cg	V <sub>DD</sub> = constant, C <sub>G</sub> = 5 to 25pF	25			ppm	

<sup>\*</sup> Q12C2 Made by Seiko Epson corporation

Note) 1 When crystal oscillation is selected by the mask option.

2 When crystal oscillation (gate capacitor built-in) is selected by the mask option.

#### ■ OSC1 (CR)

Unless otherwise specified: VDD = 2.4 to 5.5 V, Vss = 0 V, Ta = 25°C

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Oscillation start time	tsta				100	μs	
Frequency/IC deviation	∂f/∂IC	Rcr = constant	-25		25	%	

#### OSC3 (Crystal)

Unless otherwise specified: VDD = Within the operating voltage in each operating mode, Vss = 0 V, Ta = 25°C,

Crystal oscillator = Q21CA301xxx\*, RF =  $1M\Omega$ , CG2 = CD2 = 15pF

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Oscillation start time (Normal mode)	tsta	4.0 MHz crystal oscillator			10	ms	1
Oscillation start time (High speed mode)	tsta	8.0 MHz crystal oscillator			10	ms	1

<sup>\*</sup> Q21CA301xxx Made by Seiko Epson corporation

Note) 1 The crystal oscillation start time changes by the crystal oscillator to be used, CG2 and CD2.

## ■ OSC3 (Ceramic)

Unless otherwise specified: VDD = Within the operating voltage in each operating mode, Vss = 0 V, Ta = 25°C,

Ceramic oscillator = CSA4.00MG / CSA8.00MTZ\*, RF =  $1M\Omega$ , CG2 = CD2 = 30pF

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Oscillation start time (Normal mode)	tsta	4.0 MHz ceramic oscillator			1	ms	1
Oscillation start time (High speed mode)	tsta	8.0 MHz ceramic oscillator			1	ms	1

<sup>\*</sup> CSA4.00MG / CSA8.00MTZ Made by Murata Mfg. corporation

Note) 1 The ceramic oscillation start time changes by the ceramic oscillator to be used, CG2 and CD2.

#### ■ OSC3 (CR)

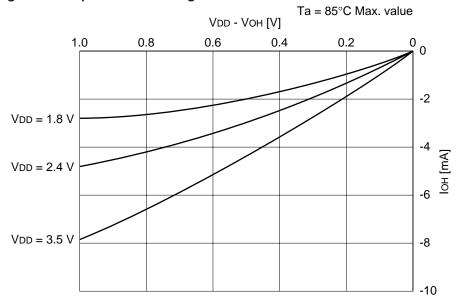
Unless otherwise specified: VDD = Within the operating voltage in each operating mode, Vss = 0 V, Ta = 25°C

Item	Symbol	Condition	Min.	Тур.	Max.	Unit	Note
Oscillation start time (Normal mode)	tsta				100	μs	
Oscillation start time (High speed mode)	tsta				100	μs	
Frequency/IC deviation (Normal mode)	∂f/∂IC	Rcr = constant	-25		25	%	
Frequency/IC deviation (High speed mode)	∂f/∂IC	RCR = constant	-25		25	%	

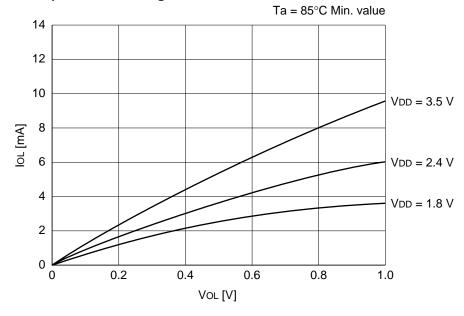
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# 8.8 Characteristics Curves (reference value)

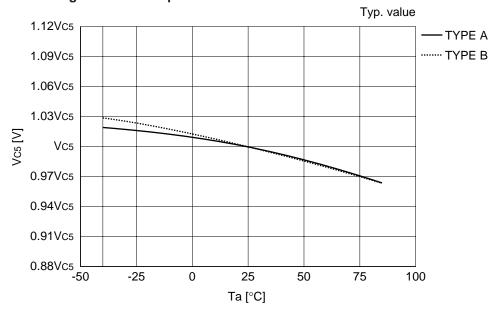
# ■ High level output current-voltage characteristic



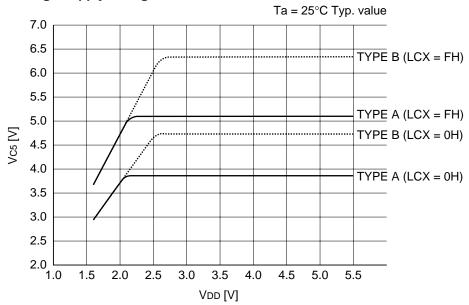
# ■ Low level output current-voltage characteristic



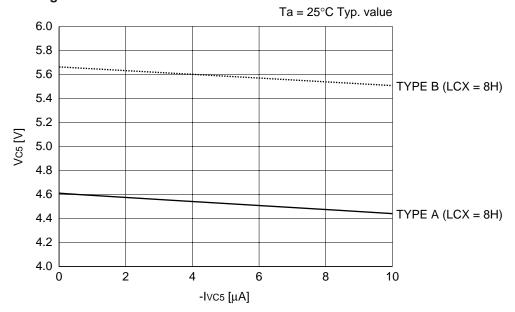
# ■ LCD drive voltage-ambient temperature characteristic



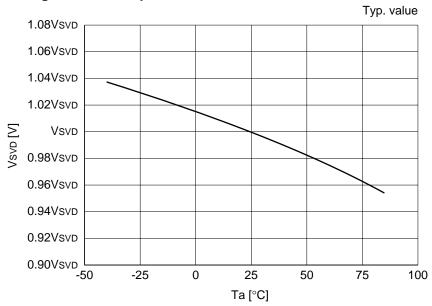
# ■ LCD drive voltage-supply voltage characteristic



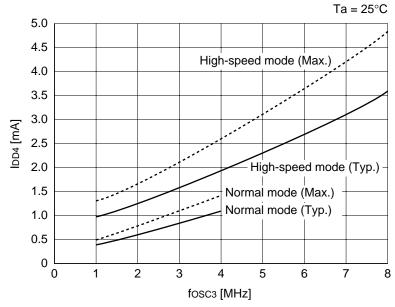
# ■ LCD drive voltage-load characteristic



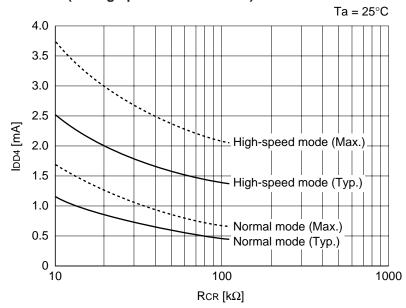
# ■ SVD voltage-ambient temperature characteristic



# ■ Power current (During operation with OSC3) <Crystal/ceramic oscillation>

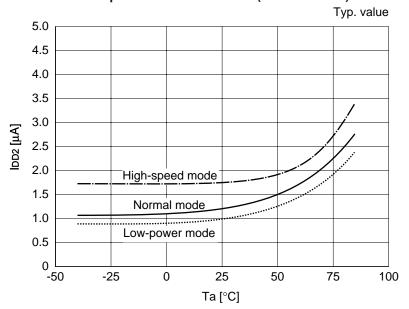


# ■ Power current (During operation with OSC3) <CR oscillation>

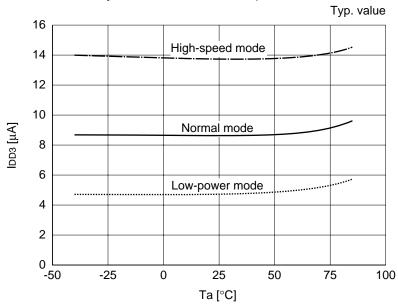


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## ■ Power current-ambient temperature characteristic (In HALT status)



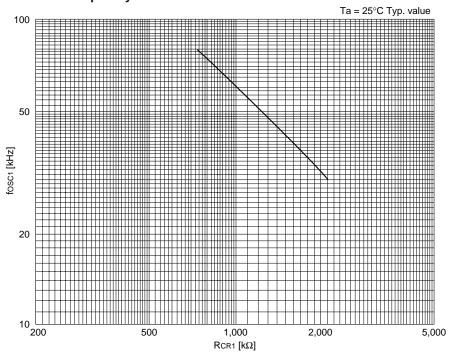
# ■ Power current-ambient temperature characteristic (CPU is under 32.768 kHz operation)



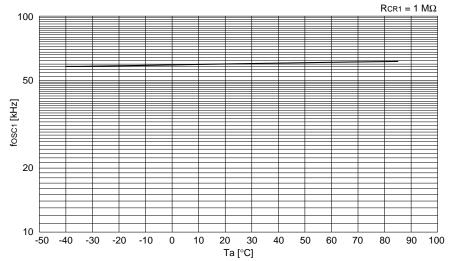
# ■ CR oscillation frequency characteristic

Note: Oscillation frequency changes depending on the conditions (components used, board pattern, etc.). In particular, the OSC3 oscillation frequency changes extensively depending on the product form (chip, plastic package or ceramic package) and board capacitance. Therefore, use the following charts for reference only and select the resistance value after evaluating the actual product. (The resistance value should be set to  $RCR \ge 15 \text{ k}\Omega$ .)

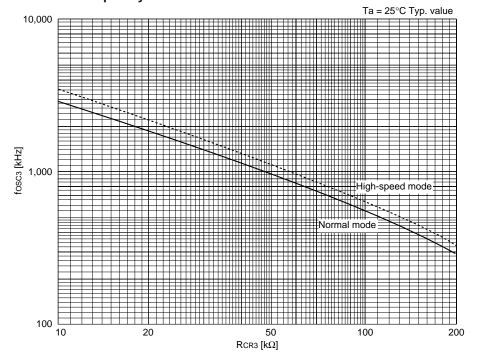
# OSC1 oscillation frequency-resistor characteristic



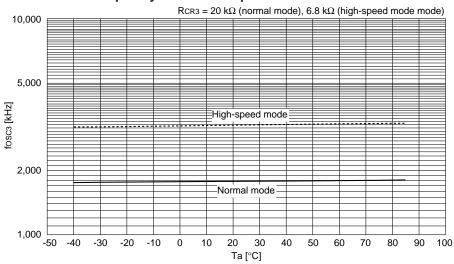
## OSC1 CR oscillation frequency-ambient temperature characteristic



## • OSC3 oscillation frequency-resistor characteristic



### • OSC3 CR oscillation frequency-ambient temperature characteristic

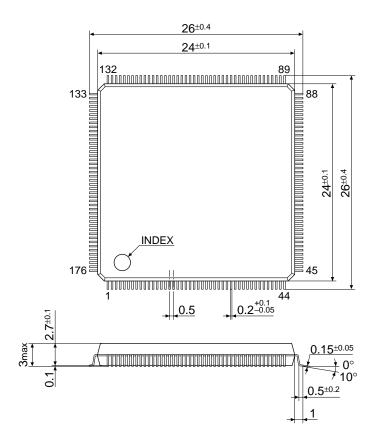


## 9 PACKAGE

## 9.1 Plastic Package

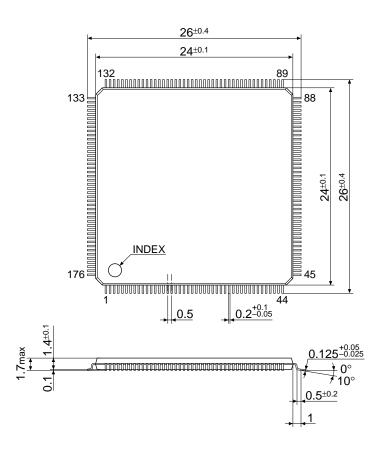
QFP18-176pin

(Unit: mm)

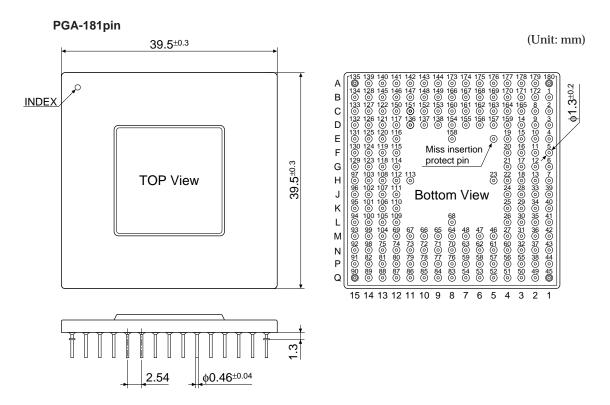


QFP21-176pin

(Unit: mm)



## 9.2 Ceramic Package for Test Samples



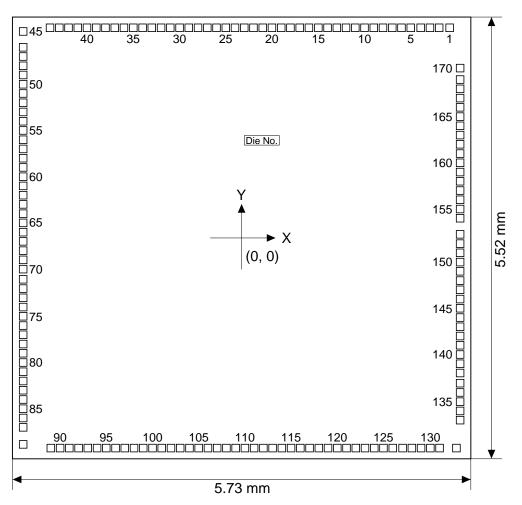
Pin No.	Pin name	Pin No.	Pin name	Pin No.	Pin name	Pin No.	Pin name	Pin No.	Pin name
1	N.C.	37	SEG37	73	CD	109	P14/CMPP0/AD4	145	R22/A18
2	SEG2	38	SEG38	74	CC	110	P13/SRDY	146	R23/RD
3	SEG3	39	SEG39	75	CB	111	P12/SCLK	147	R24/WR
4	SEG4	40	SEG40	76	CA	112	P11/SOUT	148	R25/CL
5	SEG5	41	SEG41	77	Vc5	113	P10/SIN	149	R26/FR
6	SEG6	42	SEG42	78	VC4	114	AVDD	150	R27/TOUT
7	SEG7	43	SEG43	79	Vc3	115	AVss	151	R30/CE0
8	SEG8	44	SEG44	80	Vc2	116	AVREF	152	R31/CE1
9	SEG9	45	SEG45	81	Vcı	117	Vdd	153	R32/CE2
10	SEG10	46	N.C.	82	OSC3	118	P07/D7	154	R33/CE3
11	SEG11	47	N.C.	83	OSC4	119	P06/D6	155	R34/FOUT
12	SEG12	48	N.C.	84	Vdi	120	P05/D5	156	R35
13	SEG13	49	N.C.	85	Vdd	121	P04/D4	157	R36
14	SEG14	50	SEG46	86	Vss	122	P03/D3	158	R37
15	SEG15	51	SEG47	87	Vosc	123	P02/D2	159	Vss
16	SEG16	52	SEG48	88	N.C.	124	P01/D1	160	R50/BZ
17	SEG17	53	SEG49	89	N.C.	125	P00/D0	161	R51/BACK
18	SEG18	54	SEG50	90	N.C.	126	R00/A0	162	COM0
19	SEG19	55	COM31/SEG51	91	OSC1	127	R01/A1	163	COM1
20	SEG20	56	COM30/SEG52	92	OSC2	128	R02/A2	164	COM2
21	SEG21	57	COM29/SEG53	93	TEST	129	R03/A3	165	COM3
22	SEG22	58	COM28/SEG54	94	RESET	130	R04/A4	166	COM4
23	SEG23	59	COM27/SEG55	95	MCU/MPU	131	R05/A5	167	COM5
24	SEG24	60	COM26/SEG56	96	K11/BREQ	132	R06/A6	168	COM6
25	SEG25	61	COM25/SEG57	97	K10/EVIN	133	R07/A7	169	COM7
26	SEG26	62	COM24/SEG58	98	K07	134	R10/A8	170	COM8
27	SEG27	63	COM23/SEG59	99	K06	135	N.C.	171	COM9
28	SEG28	64	COM22/SEG60	100	K05	136	R11/A9	172	COM10
29	SEG29	65	COM21/SEG61	101	K04	137	R12/A10	173	COM11
30	SEG30	66	COM20/SEG62	102	K03	138	R13/A11	174	COM12
31	SEG31	67	COM19/SEG63	103	K02	139	R14/A12	175	COM13
32	SEG32	68	COM18/SEG64	104	K01	140	R15/A13	176	COM14
33	SEG33	69	COM17/SEG65	105	K00	141	R16/A14	177	COM15
34	SEG34	70	COM16/SEG66	106	P17/CMPM1/AD7	142	R17/A15	178	SEG0
35	SEG35	71	Vref	107	P16/CMPP1/AD6	143	R20/A16	179	SEG1
36	SEG36	72	CE	108	P15/CMPM0/AD5	144	R21/A17	180	N.C.

N.C.: No Connection

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## 10 PAD LAYOUT

## 10.1 Diagram of Pad Layout



Chip thickness: 400 µm Pad opening: 100 µm

## 10.2 Pad Coordinates

Table 10.2.1 Pad coordinates

(Unit: µm)

	Pad	Coord	linates		Pad	Coord	linates		Pad		linates
No		X	Y	No	1	X	Y	No			Y
No.	Name			No.	Name			No.	Name	X (10)	
1	OSC1	2,600	2,626	58	R26/FR/TOUT	-2,731	996	115	SEG28	618	-2,626
2	OSC2	2,469	2,626	59	R27/TOUT	-2,731	881	116	SEG29	734	-2,626
3	TEST	2,354	2,626	60	R30/CE0	-2,731	765	117	SEG30	849	-2,626
4	RESET	2,238	2,626	61	R31/CE1	-2,731	650	118	SEG31	965	-2,626
5	MCU/MPU	2,114	2,626	62	R32/CE2	-2,731	534	119	SEG32	1,080	-2,626
6	K11/BREQ	1,998	2,626	63	R33/CE3	-2,731	419	120	SEG33	1,196	-2,626
7	K10/EVIN	1,883	2,626	64	R34/FOUT	-2,731	303	121	SEG34	1,311	-2,626
8	K07	1,767	2,626	65	R35	-2,731	188	122	SEG35	1,427	-2,626
9	K06	1,652	2,626	66	R36	-2,731	72	123	SEG36	1,542	-2,626
10	K05	1,536	2,626	67	R37	-2,731	-43	124	SEG37	1,658	-2,626
11	K04	1,421	2,626	68	Vss	-2,731	-159	125	SEG38	1,773	-2,626
12	K03	1,305	2,626	69	R50/BZ	-2,731	-274	126	SEG39	1,889	-2,626
13	K02	1,190	2,626	70	R51/BACK	-2,731	-390	127	SEG40	2,012	-2,626
14	K01	1,074	2,626	71	COM0	-2,731	-517	128	SEG41	2,128	-2,626
15	K00	959	2,626	72	COM1	-2,731	-633	129	SEG42	2,243	-2,626
16	P17/CMPM1/AD7	841	2,626	73	COM2	-2,731	-748	130	SEG43	2,359	-2,626
17	P16/CMPP1/AD6	726	2,626	74	COM3	-2,731	-864	131	SEG44	2,474	-2,626
18	P15/CMPM0/AD5	610	2,626	75	COM4	-2,731	-979	132	SEG45	2,685	-2,626
19	P14/CMPP0/AD4	495	2,626	76	COM5	-2,731	-1,095	133	SEG46	2,731	-2,277
20	P13/SRDY	379	2,626	77	COM6	-2,731	-1,210	134	SEG47	2,731	-2,162
21	P12/SCLK	264	2,626	78	COM7	-2,731	-1,326	135	SEG47 SEG48	2,731	-2,102
22	P11/SOUT	148	2,626	79	COM8	-2,731	-1,441	136	SEG49	2,731	-1,931
23	P10/SIN	33	2,626	80	COM9	-2,731	-1,557	137	SEG50	2,731	-1,815
24		-83	2,626	81	COM10		-1,672	138			
	AVDD					-2,731			COM31/SEG51	2,731	-1,688
25	AVss	-198	2,626	82	COM11	-2,731	-1,788	139	COM30/SEG52	2,731	-1,572
26	AVREF	-314	2,626	83	COM12	-2,731	-1,903	140	COM29/SEG53	2,731	-1,457
27	VDD	-429	2,626	84	COM13	-2,731	-2,019	141	COM28/SEG54	2,731	-1,341
28	P07/D7	-545	2,626	85	COM14	-2,731	-2,134	142	COM27/SEG55	2,731	-1,226
29	P06/D6	-661	2,626	86	COM15	-2,731	-2,250	143	COM26/SEG56	2,731	-1,110
30	P05/D5	-776	2,626	87	SEG0	-2,731	-2,367	144	COM25/SEG57	2,731	-995
31	P04/D4	-892	2,626	88	SEG1	-2,731	-2,580	145	COM24/SEG58	2,731	-879
32	P03/D3	-1,007	2,626	89	SEG2	-2,385	-2,626	146	COM23/SEG59	2,731	-764
33	P02/D2	-1,123	2,626	90	SEG3	-2,269	-2,626	147	COM22/SEG60	2,731	-648
34	P01/D1	-1,238	2,626	91	SEG4	-2,154	-2,626	148	COM21/SEG61	2,731	-533
35	P00/D0	-1,354	2,626	92	SEG5	-2,038	-2,626	149	COM20/SEG62	2,731	-417
36	R00/A0	-1,471	2,626	93	SEG6	-1,923	-2,626	150	COM19/SEG63	2,731	-302
37	R01/A1	-1,586	2,626	94	SEG7	-1,807	-2,626	151	COM18/SEG64	2,731	-186
38	R02/A2	-1,702	2,626	95	SEG8	-1,692	-2,626	152	COM17/SEG65	2,731	-71
39	R03/A3	-1,817	2,626	96	SEG9	-1,576	-2,626	153	COM16/SEG66	2,731	45
40	R04/A4	-1,933	2,626	97	SEG10	-1,461	-2,626	154	VREF	2,731	246
41	R05/A5	-2,048	2,626	98	SEG11	-1,345	-2,626	155	CE	2,731	361
42	R06/A6	-2,164	2,626	99	SEG12	-1,230	-2,626	156	CD	2,731	477
43	R07/A7	-2,279	2,626	100	SEG13	-1,114	-2,626	157	CC	2,731	592
44	R10/A8	-2,395	2,626	101	SEG14	-999	-2,626	158	СВ	2,731	708
45	R11/A9	-2,731	2,580	102	SEG15	-883	-2,626	159	CA	2,731	823
46	R12/A10	-2,731	2,382	103	SEG16	-768	-2,626	160	VC5	2,731	939
47	R13/A11	-2,731	2,267	103	SEG17	-652	-2,626	161	VC3	2,731	1,054
48	R14/A12	-2,731	2,151	105	SEG17 SEG18	-537	-2,626	162	VC4 VC3	2,731	1,170
49	R15/A13	-2,731	2,036	105	SEG19	-421	-2,626	163	VC3 VC2	2,731	1,170
	R15/A15 R16/A14	-2,731			SEG19 SEG20				VC1		1,401
50			1,920	107		-306	-2,626	164		2,731	
51	R17/A15	-2,731	1,805	108	SEG21	-190	-2,626	165	OSC3	2,731	1,516
52	R20/A16	-2,731	1,689	109	SEG22	-75	-2,626	166	OSC4	2,731	1,632
53	R21/A17	-2,731	1,574	110	SEG23	41	-2,626	167	VDI	2,731	1,747
54	R22/A18	-2,731	1,458	111	SEG24	156	-2,626	168	VDD	2,731	1,863
55	R23/RD	-2,731	1,343	112	SEG25	272	-2,626	169	Vss	2,731	1,979
56	R24/WR	-2,731	1,227	113	SEG26	387	-2,626	170	Vosc	2,731	2,212
57	R25/CL	-2,731	1,112	114	SEG27	503	-2,626	-	_	-	-

# APPENDIX S5U1C88000P1&S5U1C88816P2 MANUAL (Peripheral Circuit Board for S1C88349)

This manual describes how to use the Peripheral Circuit Board for S1C88349 (S5U1C88000P1&S5U1C88816P2). This circuit board is used to provide emulation functions when it is installed in the ICE (S5U1C88000H5), a debugging tool for the 8-bit Single Chip Microcomputer S1C88 Family.

The explanation assumes that the S1C88349/F360/816 circuit data has been downloaded into the S1C88 Family Peripheral Circuit Board (S5U1C88000P1).

For how to download circuit data into the S5U1C88000P1 and specifications of the boards, refer to Sections A.4 and A.6, respectively. For details on ICE functions and how to operate the debugger, refer to the separately prepared manuals.

## A.1 Names and Functions of Each Part

The following explains the names and functions of each part of the S5U1C88000P1&S5U1C88816P2.

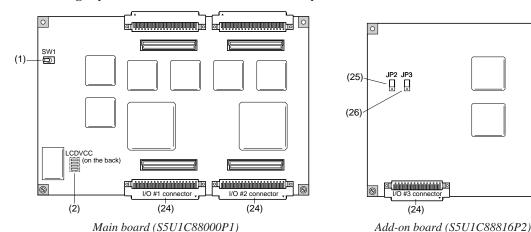


Fig. A.1.1 Board layout

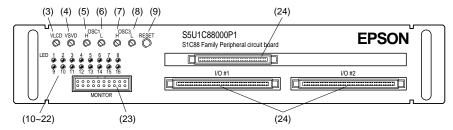


Fig. A.1.2 Panel layout (S5U1C88000P1)

#### (1) SW1

When downloading circuit data, set this switch to the "3" position. Otherwise, set to position "1".

#### (2) LCDVCC (on the back of the S5U1C88000P1 board)

The internal power voltage (Vc5) for the LCD driver can be varied using the DIP switch as shown in Table A.1.1. Be aware that the Vc5 voltage level on this board is different from that of the actual IC.

## (3) VLCD control Unused.

#### (4) VSVD control

This control is used for varying the power supply voltage to confirm the supply voltage detection (SVD) function. (Refer to Section A.5.2, "Differences from actual IC".)

Table A.1.1 Setting LCDVCC

	LCD	VCC	Setting	
1	2	3	4	Setting
ON	OFF	OFF	OFF	$V_{C5} = 6 \text{ V}$
OFF	ON	OFF	OFF	$V_{C5} = 5.75 \text{ V}$
OFF	OFF	ON	OFF	$V_{C5} = 5.5 \text{ V}$
OFF	OFF	OFF	ON	$V_{C5} = 5 \text{ V}$
Ot	her cor	nbinati	Not allowed	

\* The voltage value assumes that the LCD contrast adjustment register LC0-LC3 is 0FH. There is a need to allow for a maximum ±6% of error due to the characteristics of the parts used on this board.

#### (5) OSC1 H control

This control is used for coarse adjustment of the OSC1 CR oscillation frequency.

#### (6) OSC1 L control

This control is used for fine adjustment of the OSC1 CR oscillation frequency.

#### (7) OSC3 H control

This control is used for coarse adjustment of the OSC3 CR oscillation frequency.

#### (8) OSC3 L control

This control is used for fine adjustment of the OSC3 CR oscillation frequency.

#### (9) RESET switch

This switch initializes the internal circuits of this board and feeds a reset signal to the ICE.

#### (10) LED 1 (VDC0), LED 2 (VDC1), LED 3 (VDC2)

These LEDs go ON/OFF according to the below combinations, depending on the internal operating voltage that has been set.

Table A.1.2 VDC0, VDC1 and VDC2 LED indicators

Internal operating		LED	
voltage	VDC0	VDC1	VDC2
1.2 V	ON	OFF	OFF
2.2 V	OFF	ON	OFF
3.1 V	OFF	OFF	ON

#### (11) LED 4 (OSCC)

This LED goes ON when the OSCC register is set to "1" and OFF when the register is set to "0". (lit while the OSC3 is oscillating)

#### (12) LED 5 (LCDC)

This LED goes ON when the LCDC0 or LCDC1 register is set to "1" and OFF when the registers are both set to "0".

#### (13) LED 6 (SVDON)

This LED goes ON when the SVDON register is set to "1" also during auto sampling operation, and OFF when the register is set to "0".

#### (14) LED 7 (HVLD)

This LED goes ON when the OSCC register or BZON register is set to "1" and OFF when the register is set to "0". Furthermore, the LED goes ON during one-shot buzzer output when heavy load protection mode is selected by mask option. (Note)

#### (15) LED 8 (MODE)

This LED goes ON when the S1C88 Core CPU operates in maximum mode and OFF when it operates in minimum mode.

#### (16) LED 9 (LDTYPA)

This LED goes ON when the LCD power supply TYPE A (4.5 V) is used.

#### (17) LED 10 (LDTYPB)

This LED goes ON when the LCD power supply TYPE B (5.5 V) is used.

#### (18) LED 11 (EXLCD)

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This LED goes ON when an external LCD power supply is used.

#### (19) LED 12 (CMP0ON)

This LED goes ON when comparator 0 is

#### (20) LED 13 (CMP1ON)

This LED goes ON when comparator 1 is

#### (21) LEDs 14 to 15 (Reserved)

Unused.

#### (22) LED 16 (FPGA configuration)

If the FPGA on the S5U1C88000P1 includes circuit data, this LED lights when the power is turned on. If this LED does not light at powerup, a circuit data must be written to the FPGA before debugging can be started (turn the power on again after writing data).

#### (23) LED signal monitor connector

This connector provides the signals that drive the LEDs shown above for monitoring. The signals listed below are output from the connector pins. The signal level is high when the LED is lit and is low when the LED is not lit.

19 17 15 13 11 9 7 5 3 1 0 0 0 0 0 0 0 0 0 0000000000

20 18 16 14 12 10 8 6 4

Fig. A.1.3 LED signal monitor connector

Pin 1: LED 1 (Power control: low power mode) Pin 2: LED 2 (Power control: normal mode)

Pin 3: LED 3 (Power control: high speed mode)

Pin 4: LED 4 (OSC3 oscillation status)

Pin 5: LED 5 (LCD circuit status)

Pin 6: LED 6 (SVD circuit status)

Pin 7: LED 7 (Heavy load protection mode) (Note)

Pin 8: LED 8 (Bus mode) Pin 9: LED 9 (LCD power supply TYPE A (4.5 V))

Pin 10: LED 10 (LCD power supply TYPE B (5.5 V))

Pin 11: LED 11 (External LCD power supply)

Pin 12: LED 12 (Comparator 0 operating status)

Pin 13: LED 13 (Comparator 1 operating status)

Pin 18: OSC1 CR oscillation frequency monitor pin Pin 19: OSC3 CR oscillation frequency monitor pin

#### Pins 14 to 17 and 20 are not used.

The OSC3 CR oscillation clock is connected to pins 18 and 19. (The CR oscillation circuit on this board always operates even if crystal oscillation is selected by mask option and regardless of the OSCC register status.) These pins can be used to monitor CR oscillation when adjusting the oscillation frequency.

#### (24) I/O #1, I/O #2, I/O #3 connectors

These are the connectors for connecting the I/ O and LCD. The I/O cables  $(80\text{-pin}/40\text{-pin} \times 2)$ flat type, 60-pin/30-pin  $\times 2$  flat type) are used to connect to the target system.

#### (25) JP2

Install a Jumper on the J2 side.

Install a Jumper on the J3 side.

(Note) The S1C88349 does not support heavy load protection mode.

### A.2 Installation

## A.2.1 Installing S5U1C88816P2 to S5U1C88000P1

Aim the I/O connectors on the add-on board (S5U1C88816P2) at the front panel of the main board (S5U1C88000P1) and insert the four connectors on the back of the S5U1C88816P2 board into the corresponding connectors on the S5U1C88000P1 board.

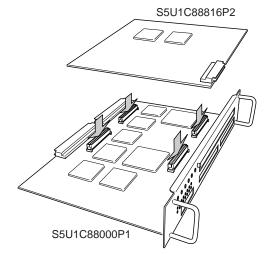


Fig. A.2.1.1 Installing S5U1C88816P2 to S5U1C88000P1

## A2.2 Installing into the ICE (S5U1C88000H5)

Insert the S5U1C88000P1 along by the lower guide rail of the ICE (S5U1C88000H5), until the connectors fit into the ICE back-panel connectors.

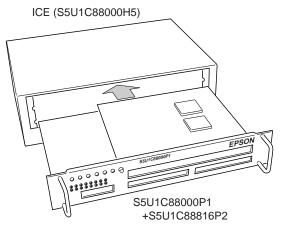


Fig. A.2.2.1 Installing into the ICE (S5U1C88000H5)

Note: The S5U1C88000P1 and S5U1C88816P2 may fail to operate if they are not adequately mounted, so be sure to mount them securely.

## A.3 Connecting to the Target System

This section explains how to connect the S5U1C88000P1&S5U1C88816P2 to the target system.

Note: Turn the power of all equipment off before connecting or disconnecting cables.

Use the I/O cables (80-pin/40-pin  $\times$  2 flat type, 60-pin/30-pin  $\times$  2 flat type) to connect between the I/O #1 to I/O #3 connectors of the front panel and the target system.

Connect the 80-pin and 60-pin cable connectors to the I/O #1 to I/O #3 connectors, and the 40-pin  $\times$  2 and 30-pin  $\times$  2 connectors to the target system. Be careful as power (VDD) is supplied to I/O #1, I/O #2 and I/O #3 connectors.

The following shows the clock frequencies generated from the on-board crystal oscillation circuits:

OSC1 crystal oscillation circuit: 32.768 kHz OSC3 crystal oscillation circuit: 4.9152 MHz When CR oscillation is selected, the oscillation frequency can be adjusted using the controls on the front panel (OSC1H and OSC1L for adjusting OSC1, OSC3H and OSC3L for adjusting OSC3). Use a frequency counter or other equipment to be connected to the OSC1 CR oscillation frequency monitor pin (pin 18) on the monitor connector or OSC3 CR oscillation frequency monitor pin (pin 19) for monitoring the frequency during adjustment. Be sure of the frequency when using this monitor pin because the CR oscillation frequency is initially undefined.

When using an external clock, choose the external clock input option for OSC1 and/or OSC3 and input the clock(s) to pin 7 and/or pin 8 of the I/O connector CN2-1 using oscillator(s).

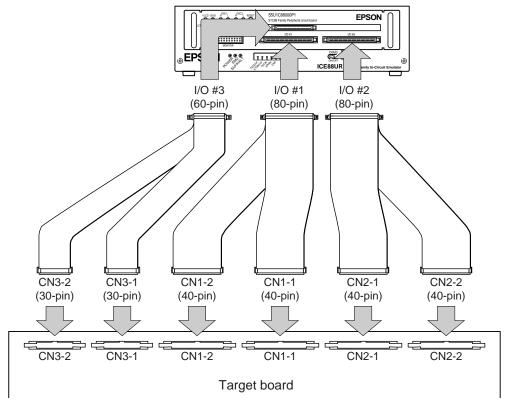


Fig. A.3.1 Connecting to the target system

## I/O connector pin assignment

Table A.3.1 I/O #1 connector

Table A.3.2 I/O #2 connector

Table A.3.1 I/O #1 connector				
	40pin CN1-1	40pin CN1-2		
No.	Pin name	No.	Pin name	
1	VDD (3.3V)	1	R12/A10	
2	VDD (3.3V)	2	R13/A11	
3	Vss	3	R14/A12	
4	Vss	4	R15/A13	
5	K00	5	R16/A14	
6	K01	6	R17/A15	
7	K02	7	R20/A16	
8	K03	8	R21/A17	
9	K04	9	R22/A18	
10	K05	10	R23/RD	
11	K06	11	R24/WR	
12	K07	12	R25/CL	
13	K10/EVIN	13	R26/FR	
14	K11/BREQ	14	R27/TOUT	
15	P00/D0	15	R30/CEO	
16	P01/D1	16	R31/CE1	
17	P02/D2	17	R32/CE2	
18	P03/D3	18	R33/CE3	
19	P04/D4	19	R34/FOUT	
20	P05/D5	20	R35	
21	P06/D6	21	R36	
22	P07/D7	22	R37	
23	P10/SIN	23	R50/BZ	
24	P11/SOUT	24	R51/BACK	
25	P12/SCLK	25	COM0	
26	P13/SRDY	26	COM1	
27	P14/CMPP0	27	COM2	
28	P15/CMPM0	28	COM3	
29	P16/CMPP1	29	COM4	
30	P17/CMPM1	30	COM5	
31	R00/A0	31	COM6	
32	R01/A1	32	COM7	
33	R02/A2	33	COM8	
34	R03/A3	34	COM9	
35	R04/A4	35	COM10	
36	R05/A5	36	COM11	
37	R06/A6	37	COM12	
38	R07/A7	38	COM13	
39	R10/A8	39	COM14	
40	R11/A9	40	COM15	

	40pin CN2-1	40pin CN2-2		
No.	Pin name	No.	Pin name	
1	VDD (3.3V)	1	SEG27	
2	VDD (3.3V)	2	SEG28	
3	Vss	3	SEG29	
4	Vss	4	SEG30	
5	RESET	5	SEG31	
6	MCU/MPU	6	SEG32	
7	OSC1	7	SEG33	
8	OSC3	8	SEG34	
9	Vc1	9	SEG35	
10	Vc2	10	SEG36	
11	Vc3	11	SEG37	
12	Vc4	12	SEG38	
13	Vc5	13	SEG39	
14	SEG0	14	SEG40	
15	SEG1	15	SEG41	
16	SEG2	16	SEG42	
17	SEG3	17	SEG43	
18	SEG4	18	SEG44	
19	SEG5	19	SEG45	
20	SEG6	20	SEG46	
21	SEG7	21	SEG47	
22	SEG8	22	SEG48	
23	SEG9	23	SEG49	
24	SEG10	24	SEG50	
25	SEG11	25	SEG51/COM31	
26	SEG12	26	SEG52/COM30	
27	SEG13	27	SEG53/COM29	
28	SEG14	28	SEG54/COM28	
29	SEG15	29	SEG55/COM27	
30	SEG16	30	SEG56/COM26	
31	SEG17	31	SEG57/COM25	
32	SEG18	32	SEG58/COM24	
33	SEG19	33	SEG59/COM23	
34	SEG20	34	SEG60/COM22	
35	SEG21	35	SEG61/COM21	
36	SEG22	36	SEG62/COM20	
37	SEG23	37	SEG63/COM19	
38	SEG24	38	SEG64/COM18	
39	SEG25	39	SEG65/COM17	
40	SEG26	40	SEG66/COM16	

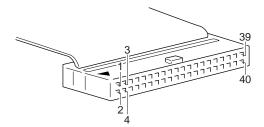


Fig. A.3.2 CN1-1/CN1-2 and CN2-1/CN2-2 pin layout

Table A.3.3 I/O #3 connector

	30-pin CN3-1	30-pin CN3-2		
No.	Pin name	No.	Pin name	
1	N.C.	1	N.C.	
2	N.C.	2	N.C.	
3	N.C.	3	N.C.	
4	N.C.	4	N.C.	
5	N.C.	5	N.C.	
6	N.C.	6	N.C.	
7	N.C.	7	N.C.	
8	N.C.	8	N.C.	
9	N.C.	9	N.C.	
10	N.C.	10	N.C.	
11	N.C.	11	N.C.	
12	N.C.	12	N.C.	
13	N.C.	13	N.C.	
14	N.C.	14	N.C.	
15	N.C.	15	N.C.	
16	N.C.	16	N.C.	
17	N.C.	17	N.C.	
18	N.C.	18	N.C.	
19	N.C.	19	N.C.	
20	N.C.	20	N.C.	
21	N.C.	21	N.C.	
22	N.C.	22	N.C.	
23	N.C.	23	AD4	
24	N.C.	24	AD5	
25	N.C.	25	AD6	
26	N.C.	26	AD7	
27	N.C.	27	N.C.	
28	N.C.	28	N.C.	
29	N.C.	29	N.C.	
30	N.C.	30	N.C.	

Note: The AVREF, AVDD and VDD voltages are fixed at 3.3 V.

The OSC1 and OSC3 pins are used to input external clocks. The Vc1–Vc5 pins are used to supply external LCD voltages.

N.C. means "No Connection". (Cannot be connected.)

## A.4 Downloading Circuit Data to the S5U1C88000P1

This board (S5U1C88000P1) comes with the FPGA that contains factory inspection data, therefore the circuit data for the model to be used should be downloaded. The following explains the downloading procedure.

- Set the switch "SW1"\*1 on this board to the "3" position.
- 2) Install this board to the ICE (S5U1C88000H5) as shown in Section A.2.2.
- 3) Connect the ICE to the host PC. Then turn the host PC and ICE on.
- 4) Invoke the debugger included in the ICE or assembler package. For how to use the ICE and debugger, refer to the manuals supplied with the ICE and assembler package.
- Download the circuit data file (.mcs) corresponding to the model by entering the following commands in the command window.

>XFER		(erase all)
>XFWR	<file name=""></file>	(download the specified
		file)*2
>XFCP	<file name=""></file>	(compare the specified file
		and downloaded data)

- Terminate the debugger and then turn the ICE off.
- 7) Remove this board from the ICE and set the switch "SW1" on the board to the "1" position.
- 8) Install this board to the ICE again.
- 9) Turn the ICE on and invoke the debugger again. Debugging can be started here.
- \*1 See Figure A.1.1, "Board layout", for the location of SW1.
- \*2 The downloading takes about 5 minutes.

## A.5 Precautions

Take the following precautions when using the S5U1C88000P1&S5U1C88816P2.

### A.5.1 Precaution for operation

- (1) Turn the power of all equipment off before connecting or disconnecting cables.
- (2) Turn the VSVD control on the front panel clockwise until it clicks before starting the debugger.
- (3) The mask option data must be loaded before debugging can be started.

#### A.5.2 Differences from actual IC

Caution is called for due to the following function and property related differences with the actual IC. If these precautions are overlooked, it may not operate on the actual IC, even if it operates on the ICE in which the S5U1C88000P1&S5U1C88816P2 has been installed.

#### (1) I/O differences

#### Interface power voltage

This board and target system interface voltage is set to +3.3 V. To obtain the same interface voltage as in the actual IC, attach a level shifter or similar circuit on the target system side to accommodate the required interface voltage.

#### Drive capability of each output port

The drive capability of each output port on this board is higher than that of the actual IC. When designing the application system and software, refer to Chapter 8, "ELECTRICAL CHARACTERISTICS" to confirm the drive capability of each output port.

#### (2) Differences in current consumption

The amount of current consumed by this board differs significantly from that of the actual IC. Inspecting the LEDs on the S5U1C88000P1 front panel may help keep track of approximate current consumption. The following factors/components greatly affect device current consumption:

## Those which can be verified by LEDs and monitor pins

- a) Power supply control: low power mode (LED 1/monitor pin 1)
- b) Power supply control: normal mode (LED 2/monitor pin 2)
- c) Power supply control: high speed mode (LED 3/monitor pin 3)
- d) OSC3 oscillation on/off control (LED 4/monitor pin 4)
- e) LCD drive control (LED 5/monitor pin 5)
- f) SVD circuit on/off control (LED 6/monitor pin 6)

- g) Maximum mode (LED 8/monitor pin 8)
- h) LCD power supply TYPE A (LED 9/monitor pin 9)
- i) LCD power supply TYPE B (LED 10/monitor pin 10)
- j) External LCD power supply (LED 11/monitor pin 11)
- k) Comparator 0 on/off control (LED 12/monitor pin 12)
- l) Comparator 1 on/off control (LED 13/monitor pin 13)

#### (3) Functional precautions

#### LCD circuit

- Pay attention to the output drive capability and output voltage of the LCD terminals (SEG, COM), since they are different from those of the actual IC. When an internal LCD power option is selected using winfog, this board generates the same LCD drive voltage regardless of the selected voltage (4.5 V or 5.5 V). However, the selected voltage level is indicated with the LED. The system and the software should be designed in order to adjust the LCD contrast. The S5U1C88000P1 board allows switching of the LCD drive voltage with its switch on the back side. (Refer to Section A.1, "Names and Functions of Each Part".)
- When the LCDC0 and LCDC1 registers are both set to "0" (LCD power control circuit is off), the SEG and COM terminal outputs of the actual IC are fixed at Vss level. Note, however, that the COM outputs are fixed at Vc4 level and the SEG outputs are fixed at Vc3 (or Vc2 for 1/4 bias) level in this board.

#### SVD circuit

- The SVD function is realized by artificially varying the power supply voltage using the VSVD control on the front panel of the S5U1C88000P1.
- There is a finite delay time from when the power to the SVD circuit turns on until actual detection of the voltage. The delay time on this board differs from that of the actual IC. Refer to Chapter 8, "ELECTRICAL CHARACTERIS-TICS" when setting the appropriate wait time for the actual IC.

#### **Oscillation circuit**

- The OSC1 crystal oscillation frequency is fixed at 32.768 kHz.
- The OSC1 CR oscillation frequency can be adjusted in the range of approx. 20 kHz to 500 kHz using the control on the S5U1C88000P1 front panel. Note that the actual IC does not operate with all of these frequencies; refer to Chapter 8, "ELECTRICAL CHARACTERIS-TICS" to select the appropriate operating frequency.
- The OSC3 crystal oscillation frequency is fixed at 4.9152 MHz.

- The OSC3 CR oscillation frequency can be adjusted in the range of approx. 100 kHz to 8 MHz using the control on the S5U1C88000P1 front panel. Note that the actual IC does not operate with all of these frequencies; refer to Chapter 8, "ELECTRICAL CHARACTERISTICS" to select the appropriate operating frequency.
- The S5U1C88000P1&S5U1C88816P2 does not include the OSC3 ceramic oscillation circuit. When ceramic oscillation circuit is selected by mask option, this board uses the on-board CR oscillation circuit. Therefore, adjust the oscillation frequency using the control on the S5U1C88000P1 front panel before debugging.
- When using an external clock, adjust the external clock (amplitude:  $3.3~V~\pm5\%$ , duty:  $50\%~\pm10\%$ ) and input to the OSC1 or OSC3 terminal with Vss as GND.
- This board can operate normally even when the CPU clock is switched to OSC3 (CLKCHG = "1") immediately after the OSC3 oscillation control circuit is turned on (OSCC = "1") without a wait time inserted. In the actual IC, an oscillation stability wait time is required before switching the CPU clock after the OSC3 oscillation is turned on. Refer to Chapter 8, "ELECTRICAL CHARACTERISTICS" when setting the appropriate wait time for the actual IC.
- Use separate instructions to switch the clock from OSC3 to OSC1 and to turn off the OSC3 oscillation circuit. If executed simultaneously with a single instruction, these operations, although good with this board, may not function properly with the actual IC.
- This board contains oscillation circuits for OSC1 and OSC3. Keep in mind that even though the actual IC may not have a resonator connected to its OSC3, this board can operate with the OSC3 circuit.
- Because the logic level of the oscillation circuit is high, the timing at which the oscillation starts on this board differs from that of the actual IC. Be aware that the CR oscillation does not stop even if the SLP instruction is executed and the monitor pin keeps outputting the oscillation clock.

#### Access to undefined address space

If any undefined space in the S1C88349's internal ROM/RAM or I/O is accessed for data read or write operations, the read/written value is indeterminate. Additionally, it is important to remain aware that the indeterminate state differs between this board and the actual IC.

#### Reset circuit

Keep in mind that the operation sequence from when the ICE with this board installed is powered on until the time at which the program starts running differs from the sequence of the actual IC. This is because this board becomes capable of operating as a debugging system after the user program and optional data are downloaded.

#### Internal power supply circuit

- Since this board does not actually change the internal operating voltage level (the voltage level is fixed at the value that enables operation for all modes), it can operate normally if the software specifies an invalid voltage level. The actual IC needs to set the internal operating voltage to the valid level according to the operating mode. Be sure to refer to Chapter 8, "ELECTRICAL CHARACTERISTICS" when setting the correct voltage. Also, when switching the control voltages, consult the technical manual to determine the appropriate wait time to be inserted.
- Since the usable operating frequency range depends on the device's internal operating voltage, consult Chapter 8, "ELECTRICAL CHARACTERISTICS" to ensure that the device will not be operated with an inappropriate combination of operating frequency and internal power supply.
- The LCD drive voltage on this board is different from that on the actual IC.

#### (4) Notes on model support

#### Parameter file

The ROM, RAM and I/O spaces in the ICE with this board installed are configured when the debugger on the personal computer starts up using the parameter file (88349.par) provided for each model.

The parameter file allows the user to modify its contents according to the ROM and RAM spaces actually used. However, do not configure areas other than below.

ROM area: 0000H to BFFFH RAM area: F000H to F7FFH Stack area: F000H to F7FFH

#### Access disable area

When using this board for development of an S1C88349 application, be sure not to read and write from/to I/O memory addresses FF46H and FF47H.

### (5) Precautions on replacement S5U1C88816P1 with S5U1C88000P1+S5U1C88816P2

- In S5U1C88816P1, the AD4–AD7 analog input pins are shared with the P14–P17 I/O port pins. In S5U1C88000P1+S5U1C88816P2, the AD4–AD7 pins are assigned in the I/O #3 connector and the P14–P17 pins are assigned in the I/O #1 connector. Pay attention to the connection to the target board.
- The I/O voltage is different; it is 5 V in S5U1C88816P1 and 3.3 V in S5U1C88000P1+S5U1C88816P2.

## A.6 Product Specifications

## A.6.1 S5U1C88000P1 specifications

#### S5U1C88000P1

Dimensions (mm):

247.5 (wide)  $\times$  165 (depth)  $\times$  44.6 (height)

Weight:

Approx. 500 g

Power supply:

DC 5 V  $\pm$  5%, less than 1 A

(supplied from ICE main unit)

#### I/O connection cable (80-pin/40-pin x 2, 2 cables)

S5U1C88000P1 connector (80-pin):

KEL 8830E-080-170L, or equivalent

Cable connector (80-pin):

KEL 8822E-080-171 × 1

Cable connector (40-pin):

3M 7940-6500SC  $\times 2$ 

Cable:

40-pin flat cable  $\times 2$ 

Interface:

CMOS interface (3.3 V)

Length:

Approx. 40 cm

#### Monitor signal cable

S5U1C88000P1 connector:

3M 7610-5002SC, or equivalent

Cable connector (10-pin):

3M 7910-6500SC × 1

Interface:

CMOS interface (3.3 V)

Length:

Approx. 40 cm

#### Accessories

40-pin connector for the target system:

3M 3432-6002LCSC

#### A.6.2 S5U1C88816P2 specifications

#### S5U1C88816P2

Dimensions (mm):

 $154.35 \text{ (W)} \times 153 \text{ (D)} \times 18 \text{ (H)}$ 

#### I/O cable (60-pin/30-pin x 2, 1 cables)

S5U1C88816P2 connector (60-pin):

KEL 8830E-060-170L

Cable connector (60-pin):

KEL 8822E-060-171

Cable connector (30-pin):

Connector 3M 7930-6500SC × 2

 $\times 1$ 

 $\times 2$ 

 $\times 2$ 

185

Strain relief 3M 3448-7930  $\times 2$ 

Cable:

30-pin flat cable

Interface:

CMOS interface (3.3 V)

Length:

Approx. 40 cm

#### **Accessories**

30-pin connector for the target system: 3M 3440-6002LCSC

 $\times 4$ 

## **EPSON**

## **International Sales Operations**

#### AMERICA

#### EPSON ELECTRONICS AMERICA, INC.

#### - HEADQUARTERS -

150 River Oaks Parkway San Jose, CA 95134, U.S.A.

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#### Northeast

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#### Southeast

3010 Royal Blvd. South, Suite 170 Alpharetta, GA 30005, U.S.A.

Phone: +1-877-332-0020 Fax: +1-770-777-2637

#### **EUROPE**

#### **EPSON EUROPE ELECTRONICS GmbH**

#### - HEADQUARTERS -

Riesstrasse 15

80992 Munich, GERMANY

Phone: +49-89-14005-0 Fax: +49-89-14005-110

#### DÜSSELDORF BRANCH OFFICE

Altstadtstrasse 176

51379 Leverkusen, GERMANY

Phone: +49-2171-5045-0 Fax: +49-2171-5045-10

#### FRENCH BRANCH OFFICE

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### Scotland Design Center

Integration House, The Alba Campus Livingston West Lothian, EH54 7EG, SCOTLAND Phone: +44-1506-605040 Fax: +44-1506-605041

#### **ASIA**

#### EPSON (CHINA) CO., LTD.

23F, Beijing Silver Tower 2# North RD DongSanHuan ChaoYang District, Beijing, CHINA

Phone: +86-10-6410-6655 Fax: +86-10-6410-7320

#### SHANGHAI BRANCH

7F, High-Tech Bldg., 900, Yishan Road

Shanghai 200233, CHINA

Phone: +86-21-5423-5522 Fax: +86-21-5423-5512

#### EPSON HONG KONG LTD.

20/F, Harbour Centre, 25 Harbour Road

Wanchai, Hong Kong

Phone: +852-2585-4600 Fax: +852-2827-4346

Telex: 65542 EPSCO HX

#### **EPSON TAIWAN TECHNOLOGY & TRADING LTD.**

14F. No. 7. Song Ren Road

Taipei 110

Phone: +886-2-8786-6688 Fax: +886-2-8786-6677

#### **HSINCHU OFFICE**

No. 99, Jiangong Road Hsinchu City 300

Phone: +886-3-573-9900 Fax: +886-3-573-9169

#### EPSON SINGAPORE PTE., LTD.

401 Commonwealth Drive, #07-01 Haw Par Technocentre, SINGAPORE 149598 Phone: +65-6586-3100 Fax: +65-6472-4291

#### SEIKO EPSON CORPORATION KOREA OFFICE

50F, KLI 63 Bldg., 60 Yoido-dong

Youngdeungpo-Ku, Seoul, 150-763, KOREA

Phone: +82-2-784-6027 Fax: +82-2-767-3677

#### **GUMI OFFICE**

6F, Good Morning Securities Bldg., 56 Songjeong-Dong

Gumi-City, Seoul, 730-090, KOREA

Phone: +82-54-454-6027 Fax: +82-54-454-6093

#### SEIKO EPSON CORPORATION **ELECTRONIC DEVICES MARKETING & SALES DIV.**

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